

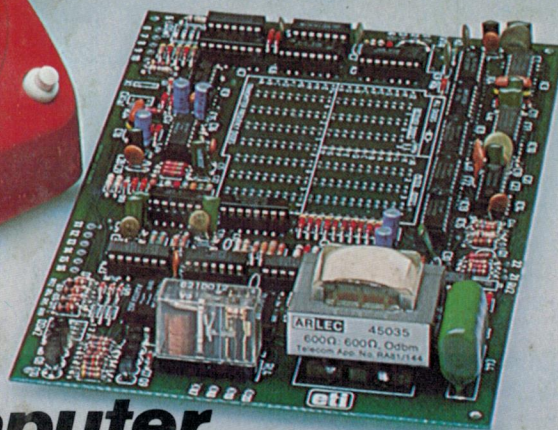
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Communication**

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loudspeaker protector***

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**Speech synthesiser
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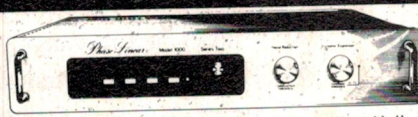
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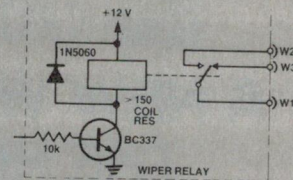
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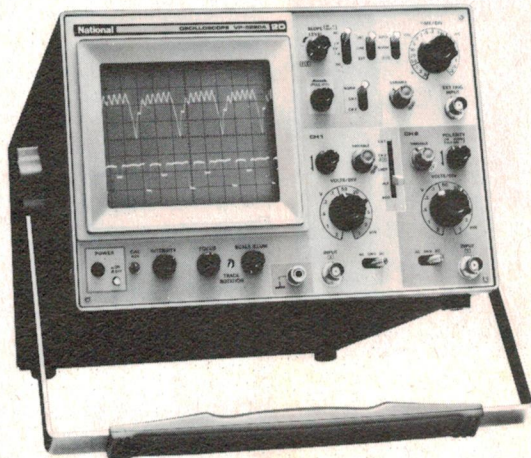
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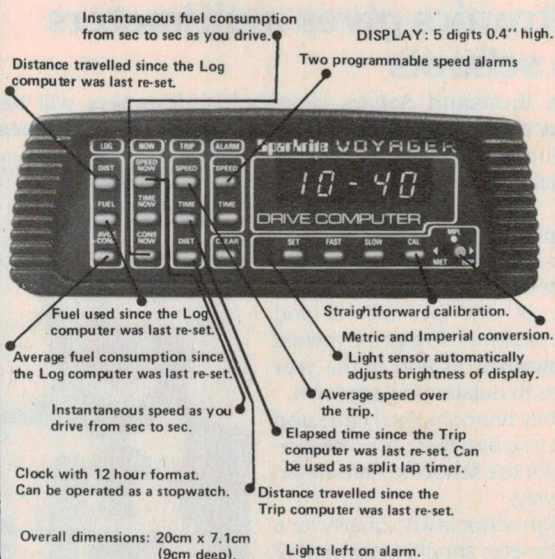
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Telecom technique gets international recognition

A new technique developed by Telecom Australia's Research Laboratories in Melbourne has received international recognition by being accepted by the International Telegraph and Telephone Consultative Committee (CCITT).

The technique determines a specification called the Crosstalk Noise Figure for a particular repeater and measures the performance of repeaters used in the digital transmission of telecommunications.

Digital transmission systems are being installed extensively by Telecom Australia and many overseas telephone administrations. These systems use pulse code modulation (PCM) for transmitting information such as the human voice.

They operate by taking very frequent samples of the signal level, encoding the level as a binary number and then transmitting this number as a series of very rapid on-or-off pulses.

Using digital transmission over existing cables allows up to

30 voice circuits to be carried on a pair of wires. These systems, however, require repeaters to re-transmit the signals about every two kilometres along the cable. One limitation on the number of systems and on the distance between repeaters is coupled interference or crosstalk between the PCM systems installed on different pairs of wires in the same cable.

The actual limits depend both on the crosstalk characteristics of the cable and on the sensitivity of the digital repeaters to this type of interference.

A specification called the 'Crosstalk Noise Figure' assesses the sensitivity of a repeater to crosstalk interference. It can be measured readily for each repeater and then used directly by Telecom network

designers to establish the maximum distance between repeaters and the number of PCM systems that can be installed on a given cable.

A technique to make this measurement has been invented by Dr Alan Gibbs of Telecom's Research Laboratories. Patents have been applied for. The practicability of the crosstalk noise figure concept, amply verified by other workers at the Laboratories, has been widely reported in technical publications around the world.

This work, which provided the basis of contributions to the CCITT by Telecom, led to the international acceptance of the crosstalk noise figure at a recent meeting in Geneva, by the appropriate CCITT working party.

Prior to adoption of the specification, engineers of the Bell Telephone Laboratories in the

United States had undertaken their own investigation of the crosstalk noise figure performance measure. During those investigations they loaned several digital repeaters to Telecom Australia as part of a joint program. Local industry has also been involved.

A prototype instrument for measuring the crosstalk noise figure is being developed by Jacobs Radio (Aust) Pty Ltd in Bayswater, a suburb of Melbourne. This work is being carried out under Telecom's Industrial Research and Development Contract Program which fosters indigenous expertise in telecommunications.

The CCITT is part of the International Telecommunication Union, a United Nations agency. It is responsible for the international standardisation of telephone and other telecommunications networks.

Altronics gives multimeters to schools

Five thousand dollars worth of multimeters will be given away to high schools who conduct electronics courses.

Altronics in Perth is making available 200 of their very popular Q1002 20 kohms/volt multimeters completely free to high schools which are running electronics courses. There is a limit of one per school and Altronics intends that the meters be awarded as end of the year prizes to outstanding students.

They are completely free and post free and there is no obligation for the school to Altronics in any way.

High schools who qualify for a multimeter should send in their request, on the school's letterhead, to Altronics, 105 Stirling St, Perth WA 6000.



Apple backs Telidon

Apple, one of the world's biggest manufacturers of personal computers, announced at a computer exhibition in the US that it has decided to back the Telidon viewdata system. Apple will be selling a US\$595 add-on circuit board for its computers that will allow them to work with Canada's Telidon viewdata system.

Apple decided this after considering Prestel, and its French rival, Teletel. This will be another setback for the British as they had hoped to establish Prestel as the world standard technology for viewdata terminals.

Telidon's key advantage over the two broadly similar European viewdata display methods is that it can offer far superior on-screen graphics. It also enjoys the support of the giant American Telephone & Telegraph.

Prestel and Teletel use an 'alpha-mosaic' method of presenting information on a viewdata page. This gives

crude, Lego-like graphics.

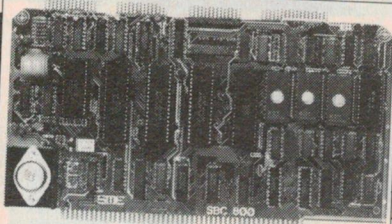
The Canadian system employs an 'alpha-geometric' technique which needs more sophisticated electronics in the terminal but can deliver properly curved lines.

Apple will promote Telidon as a cheap way to create computer graphics. A typical application would be where users compile their own personal 'databases' of graphic and textual information and swap the information with other Apple owners.

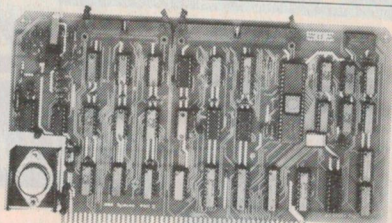
If the viewdata facility is successful in North America, Apple says that it will launch it in Europe.

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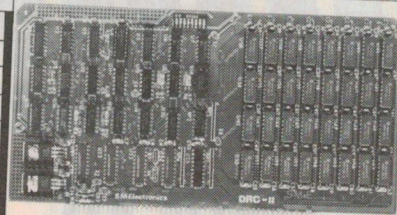
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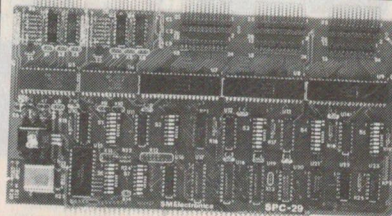
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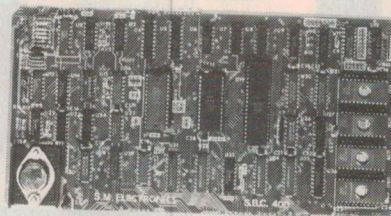
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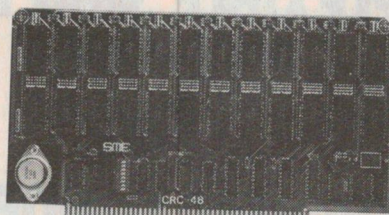
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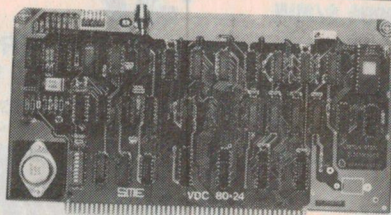
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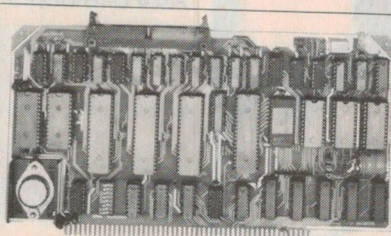
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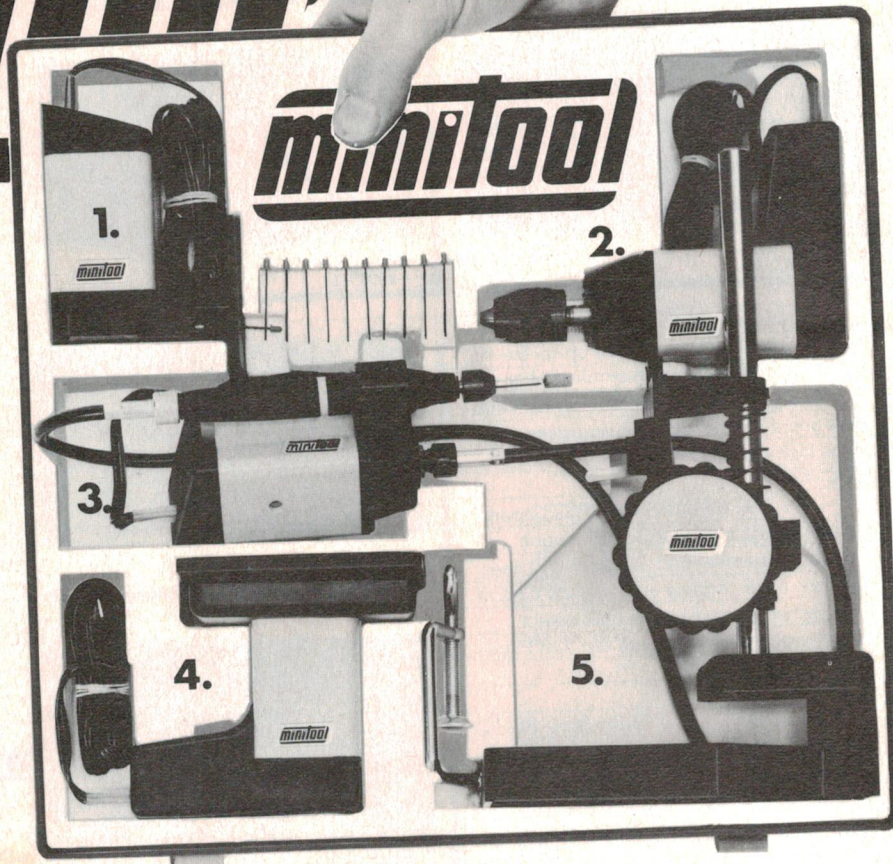
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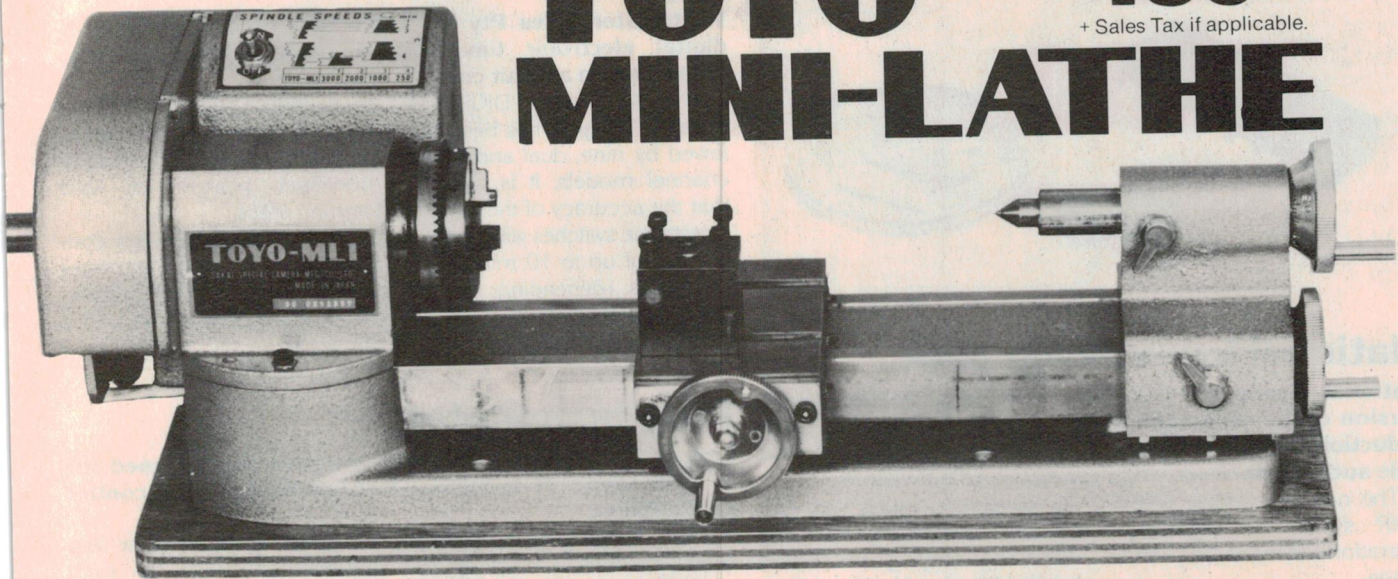
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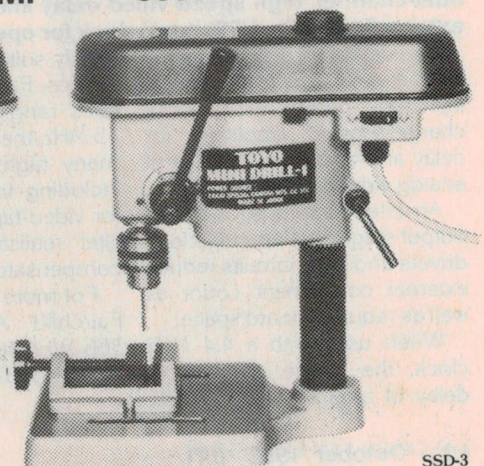
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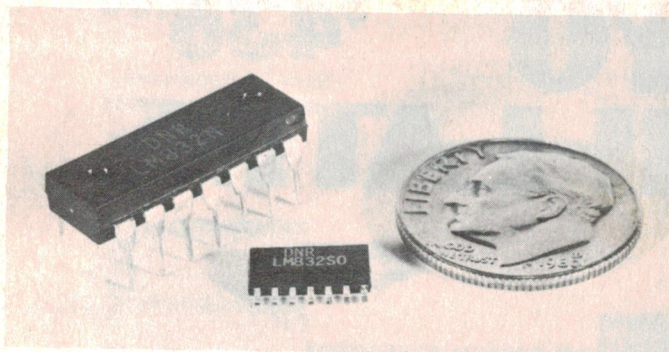
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MD-1 — 6 SPEED
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National low-voltage DNR

National Semiconductor has introduced a low-voltage version of the company's widely-used DNR (dynamic noise reduction) system for use in compact battery-operated portable audio equipment, video cameras and televisions.

The new low-voltage LM832 DNR system is capable of operation within the 1.5—9 volt range.

National's DNR system does not require signal encoding, and as a result is a universally-effective noise reduction system for all tape and FM broadcast signals.

No special tolerance devices and very minimal additional components are required for use of the LM832. For example, only twelve external components are required for the use of this device in stereo applications.

The DNR system utilizes a noise reduction method that serves to eliminate noise that may already be present in the source program.

DNR is a single chip dynamic-

ally variable low pass filter that reduces undesirable hiss and noise by varying the audio bandwidth as a function of the high frequency content of the input signal.

As the amplitude of the high frequency sound drops to where background noise could become audible, DNR reduces the bandwidth, thereby decreasing the audible noise level. Up to 14 dB (weighted) of noise reduction is possible when using the DNR system, National Semiconductor claim.

The LM832 is housed in a standard 14-pin dual-in-line package. Production quantities will be available this month, according to National Semiconductor.

Video delay line

Fairchild's CCD323 is an electrically variable, 283½-bit, dual-channel, high speed video delay line requiring only an externally supplied TTL-level clock for operation.

The device incorporates CCD analog shift registers, charge-injection ports and output charge-sensing amplifiers for delay and temporary storage of analog video signals.

An internal sample-and-hold output stage plus on-chip clock drivers and logic circuits reduce external component count as well as required board space.

When used with a 4.4 MHz clock, the device produces a delay of approximately 64 µs,

ideally suited for PAL TV applications, Fairchild say. With data rates ranging from 10 kHz to 15 MHz, the CCD323 is useful in many high-speed applications including time-base correction for video-tape recorders, comb filter realisations and dropout compensators.

For more information contact Fairchild Australia Pty Ltd, 366 Whitehorse Rd, Nunawading Vic. 3131. (03)877-5444.

Digital time switches

Wattmaster Sales Pty Ltd claim that their new range of digital electronic time switches will reduce electricity consumption and cut costs.

The 4 channel DIGI 127 released last year has been followed by nine, dual and single channel models. It is claimed that the accuracy of these electronic time switches will produce savings of up to 10 minutes in 10 hours (depending on the installation).

They can be used for accurate

switching control of interior lights, heating, cooling, flood-lights and controlled lighting of municipal playgrounds and sporting areas.

For further information contact Mr. J. Cronly, Wattmaster Sales Pty Ltd, 11 Rachel Close, Silverwater NSW 2141. (02) 648-1332.

Plug-in modules

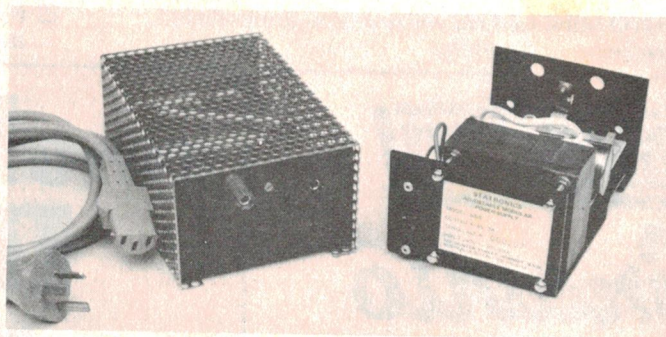
Logcon function and interface modules are designed for a wide range of industrial instrumentation and control applications.

The ac-powered product series includes a proportional switching amplifier, a tachometer amplifier and switch, a miniature four digit counter and a servo-control amplifier.

In the dc-powered function series there is a dual variable gate, a six input cascable AND/OR gate and a dual relay interface or drive module.

There is also an eight step sequence generator, an eight step to four function selector and an ac-dc power supply to drive these modules.

If you want to know anything more about these modules contact Solid-State Design and Development Pty Ltd, 10 Vista St, Bulleen Vic. 3105. (03) 850-6884.



Modular power supplies

These low cost, compact power supplies from Statronics are current limiting and feature an adjustable crowbar over-voltage protection.

There are four models in the range and, with the exception of the 53/2, they have a crowbar current rating of 5 A.

| Model | Current | Voltage |
|-------|----------|---------|
| 53/2 | 1 A dual | 12-15 V |
| 53/3 | 3 A | 12-15 V |
| 53/4 | 2.5 A | 24 V |
| 53/5 | 3 A | 5 V |

The line regulation is quoted as 0.5% and the load is regulated to 1%. The ripple and noise is quoted as 5 mV peak-to-peak.

For further information contact Statronics Pty Ltd, 103 Hunter St, Hornsby NSW 2077. (02)476-5714.

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DPM-200

LCD panelmeters BACK.....

We have been unable to keep up with the demand for these, that is why you have not seen them in our ads for the last few months.

DPM-200 - 3½ digit display with annunciators (pictured), 0.6" high, 200mV full scale. Each unit supplied with data sheet.

DPM-05 (Not illustrated), 3½ digit display with "plus", "minus" and "low batt". Annunciators with 0.5" readout. Both units sample at 3/second.

If you want to express any physical measurement in a bright easy to read display these are for you. They contain all analogue-to-digital electronics and LCD drive circuitry. Send SAE for more information.

good news

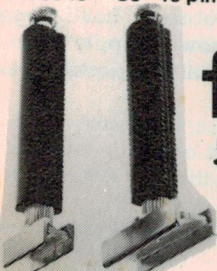
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- Ground strap can be connected.
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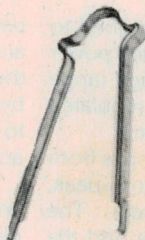


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\$2.95**

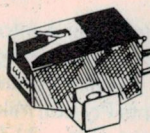
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Deceptively simple looking device. One piece metal construction, 8-40 pins
ET-840 ONLY \$2.95
IMPORTANT!!

Don't be conned into buying a non conductive inserter/extractor. The possible static damage to your MOS I.C.'s could cost you a fortune!!



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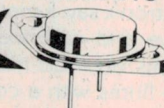


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BWD enter rental market

BWD Instruments of Mulgrave, Victoria have entered the equipment rental market with their comprehensive range of test equipment at what they claim are extremely competitive terms and prices.

BWD's products are widely used in industrial applications by military services, civil aviation and in education establishments ranging from primary schools to tertiary institutions.

For the fields of research, design, education and servicing BWD can provide a wide range of electronic test instruments comprising oscilloscopes up to 100 MHz (including storage and dual-trace models) sine wave

generators, signal, pulse and function generators, stabilised dc power supplies, electronic educational aids and equipment and accessories for many applications.

For further details and information on the product range available for rental, contact BWD Instruments Pty Ltd, Miles Street, Mulgrave Vic. 3170. (03)561-2888.

Tektronix FOTDR tester

The new OF150, introduced by Tektronix, has been designed to become the reference 'standard' for FOTDR's (fibre optic time domain reflectometer).

This instrument is rugged, portable and easy-to-use in the areas of field installation and maintenance of fibre optic links, Tektronix say. It provides quantitative and calibrated measurements on multi-mode 125 μm OD fibres with a core diameter of 50 μm .

Tektronix claim that the new OF150 is the first instrument which can make precise, calibrated loss and distance measurements while offering monolithic, rugged, and portable features. A chart recorder documents the presentation displayed on the CRT. There is also an LCD readout which provides the user with the capability of making calibrated, repeatable measurements.

Primary users will be tele-

phone companies, the military, and broad-band networks providing television, telephone, and two-way data communications to households.

The OF150 applies a pulse of radiant energy to the fibre under test via the optical output connector. As the pulse is travelling through the fibre, some energy is reflected back to the OF150. These reflections are processed and then displayed on the CRT, where distance and loss measurements can be made using the horizontal and vertical markers.

For more information about the OF150 contact Tektronix Australia Pty Ltd, 80 Waterloo Rd, North Ryde NSW 2113. (02)888-7066.

High-transparency drafting film

Known as Herculene Hi-Trans, a new high transparency polyester drafting film has been designed for use in overlay drafting.

The film is single matte but the special anti-static and anti-curl coating on the reverse side of the film will accept ink almost as well as the matt surface, the makers claim.

Two thicknesses are available, 0.08 mm (0.003 inch) and 0.1 mm (0.004 inch). Standard

size rolls up to 1.37 m (54 inch) wide will be available. Pre-cut sheets will be available to any size and can be offset-printed to order.

For samples, prices and literature, please contact Jasco Pty Ltd, P.O. Box 135, West Ryde NSW 2114. (02)807-1555.

Static protection for CMOS ICs

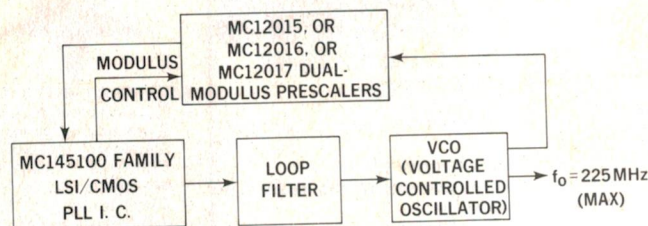
Static electricity can damage sensitive CMOS semiconductors, as we all know. This problem is aggravated as more and more circuits are crammed in a given size of silicon chip.

Amtext Electronics has available a range of conductive static protection products. The high density 'Statfree' foam is 6 mm thick and comes in two sizes: 300 x 600 mm and 635 x 1900 mm. This non-corrosive foam will not attack IC pins, the

makers claim.

The Statfree bags are made of nylon and come in sizes of 150 x 254, 203 x 305 and 254 x 457 mm.

Contact Amtext Electronics, P.O. Box 285, Chatswood NSW 2067. (02)411-1323.



Low power, dual-modulus prescalers

New high-speed prescalers from Motorola have toggle frequencies of 225 MHz at a typical power supply drain of 6 mA which makes these devices ideal for portable and mobile applications.

Designated MC12015P, MC12016P and MC12017P they provide dual-modulus division ratios of 32/33, 40/41 and 64/65 respectively.

The devices are designed to operate from unregulated power supplies over the voltage range of 5.5 to 9.5 Vdc or a regulated supply of 4.5 to 5.5 Vdc.

Signal-input level ranges from 200 to 800 mV peak-to-peak, according to Motorola. The modulus-control input and the prescaler output are TTL-compatible. This allows designers to achieve 'two-chip' frequency

synthesizers capable of over 200 MHz operation and requiring less than 10 mA of current drain.

The low current drain makes battery operation feasible and reduces heat generation in the equipment. System divide-by-N ratios of less than 1000 to greater than 65 000 are achievable.

These dual-modulus prescalers are available in small 8-pin plastic DIP packages from Motorola Semi Conductor Products, 250 Pacific Hwy, Crows Nest NSW 2065. (02)438-1955.

Telescope project underway

Work on the construction of the \$25 million Australia Telescope would begin before the end of the year, said the Minister for Science and Technology, Mr David Thomson.

The project received an initial grant of \$820 000 in CSIRO's allocation of the 1982/83 Federal Budget, and a commitment for funding over the next six years.

"The approval of the project is a shot in the arm for Australia's radio astronomers who faced the prospect of rapidly falling behind the rest of the world as their equipment became outdated," Mr Thompson said.

"But just as importantly, the telescope will mean a similar shot in the arm for Australian technology because the project is an all-Australian undertaking which will inject \$20 million into the technology sector."

Mr Thomson said antenna design studies and costing for the project had been undertaken as part of the CSIRO proposals for funding. "This means that detailed tender documents can now be prepared with a minimum of delay," he said.

The telescope consists of new antennas at Culgoora near Narrabri and at Siding Spring near Coonabarabran. These antennas will be linked by microwaves to the existing 64-metre CSIRO Radio telescope at Parkes, 350 kilometres away.

The combination of the antennas creates, in effect, one

huge, powerful radio telescope about 300 kilometres across. The telescope will provide high resolution radio images of the southern sky and ensure that Australia remains among the world leaders in radio astronomy.

Mr Thomson said without the instrument, Australian astronomers faced the bleak prospect of not being able to keep up with advances made overseas using superior instruments.

"As well, as the existing instruments aged, Australia would lose some of its most talented young radio-astronomers to overseas research groups," he said.

"But with the Australia Telescope, Australian radio astronomy has an instrument which will carry it into the 21st century."

"The construction of the telescope will provide a stimulus to a range of technological areas. These include antenna design and construction, low noise amplifiers, optical fibres and a range of electronics including very large scale integrated circuits."

The Australia Telescope would be operated as a National Facility by the CSIRO Division of Radiophysics from its Headquarters in Sydney.

PNP and NPN Darlingtontons

Following last year's introduction of the BDT20 p-n-p high-energy Darlington power transistor, Philips now announce the complementary n-p-n version, the BDT21. Both Darlingtontons are designed for driving inductive loads such as motors and relays.

Owing to the inclusion of a monolithic zener protection diode, the transistors can withstand very high energies up to a maximum of 100 mJ at an ambient temperature of 125°C, Philips claim.

The Darlingtontons have a V_{CEO} of 130 V and a total power dissipation

of 62.5 W at a mounting base temperature of 25°C. At 3 A, the h_{FE} of the BDT20 is given as 500 minimum, and that of the BDT21 750.

For further information contact Elcoma, 67 Mars Road, Lane Cove NSW 2066. (02) 427-0888.



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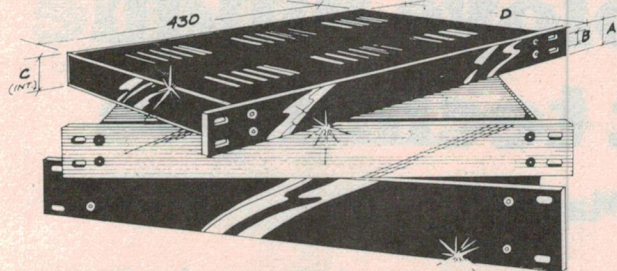
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| H 0403 | Natural | 132 | 89 | 126 | 49.50 | 45.00 |
| H 0411 | Black | 44 | 34 | 38 | 39.50 | 38.00 |
| H 0412 | Black | 88 | 57 | 82 | 45.00 | 42.50 |
| H 0413 | Black | 132 | 89 | 126 | 49.50 | 45.00 |

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Semi-conductors: 1-LSI, 4 - IC's, 7 - FET, 34 - Transistors, 42 - Diodes

Frequency range:

| | | | | | |
|-----|--------------|------|-----------|------|-------------|
| LW | 145-360 KHz | SW3 | 9-22 MHz | VHF3 | 88-108 MHz |
| MW | 525-1600 KHz | SW4 | 22-30 MHz | VHF4 | 108-136 MHz |
| SW1 | 1.6-3.8 MHz | VHF1 | 30-50 MHz | VHF5 | 144-176 MHz |
| SW2 | 3.8-9 MHz | VHF2 | 68-86 MHz | UHF | 430-470 MHz |

Power supply: AC 240V 50Hz. DC 12V (8 x "D" cells). Ext. DC 12V Car/Boat.

Speaker: 12.5cm Permanent Dynamic Speaker (3.2 ohm)

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3 x Telescopic Antenna for SW, VHF and UHF

Controls: Power ON-OFF switch — Digital display ON-OFF switch — Tape-Radio switch, Wide-Narrow band selector switch, AM band selector (LW/MW/SW1/SW2/SW3/SW4), VHF band selector (VHF1/VHF2/VHF3/VHF4/VHF5/UHF), Ant. Selector (Telescopic ANT./EXT. ANT.), Tuning control (direct gear drive),

volume control, Bass control, Treble control, Squelch control, BFO pitch control, RF gain control, Antenna adjustor control, Mode switch (USB-NOR-LSB/CW)

Terminals: Ext. Speaker/Headphone Jack, Tape IN-OUT jack, VHF/UHF ANT. connector (coaxial), SW EXT. ANT. terminal (Screw), Ext. battery jack

Meter: Tuning Meter

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Display: LW/MW/SW-1 KHz SW2-4/VHF1-5 MHz

Control: SW Calibrator Dimensions: 452mm (W) x 288mm (H) x 130mm (D)

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Please note that resellers may not have all the items advertised in stock, and as resellers have to bear the cost of freight, prices may be slightly higher than advertised. ALTRONICS reseller prices should however represent a considerable saving over our competitors' prices.

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Direct-connect computer modem employs unique new circuit technique

Trevor Marshall

This article describes the design and construction techniques of a direct-connect computer modem to facilitate communications between computers over cables, the telephone network or radio links. It employs a unique 'commutated filter' circuit technique for which a petty patent application has been filed. The way in which this circuit technique is employed overcomes many of the problems associated with conventional modem technology, permitting a very flexible design capable of a range of answer/originate operating modes and 'auto' operation under software control.

IN ORDER to transfer computer and other data of a digital (binary) nature over voice grade (analogue) telephone lines, radio links or cables it is current practice to convert the digital data to a frequency shift keyed (FSK) analogue signal. Data in such a format can be transmitted at up to 1200 bits per second (BPS) over a standard local or STD connection. Either of two frequencies are transmitted, by convention the lower is usually transmitted for a binary *mark*, the higher for a binary *space*. Devices to accomplish this task are called digital data modems. Devices to send analogue data via FSK are called analogue data modems.

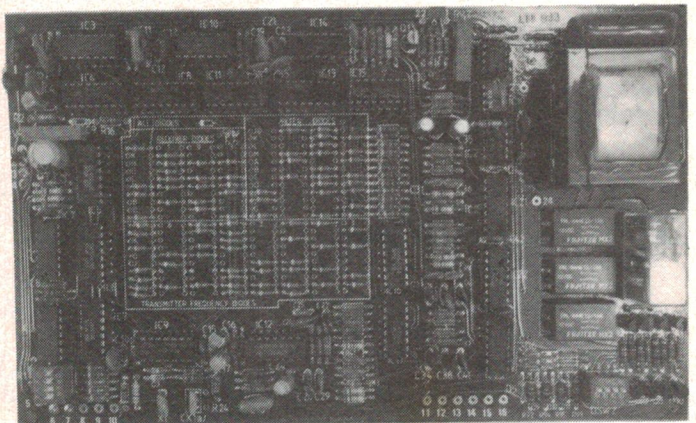
Many different channels for mark and space frequencies are currently used in Australia. Those used purely for data include a full duplex standard 980/1180 Hz and 1650/1850 Hz, a 1200 BPS simplex standard of 1300/2100 Hz and a 75 BPS channel at 390/450 Hz.

Prior Art

It is not common practice to manufacture equipment capable of performing in many (or all) these standard modes. This is due to several factors, but the cost and method of manufacture of the channel filters used to separate the wanted signals from random and impulsive noise or the backward data channel in the receiver has remained the primary consideration. The phase response of these filters is important, as they are being used to process a step frequency signal. Sharp cutoff filters cause a resonance phenomena, reducing the maximum data rate.

Linear-phase or approximately linear-phase filters do not have good selectivity in the frequency domain, and consequently more complex (higher order) filter networks are needed for optimal performance. The prior art has called for different filter network components for each channel to be used. These components, which are required to be manufactured to close tolerances (typically better than 5%), are then switched by electronic or manual means when the new channel is selected.

Alternatively, separate hybrid integrated circuit filter networks, each manufactured to high precision, have been used in a similar fashion.



The invention and the modem

The invention employed in this modem design relates to an improved method of manufacture for these channel filters and associated filters for the recovered data. Only one precision filter network is required to implement all the transfer functions for all current (or any new) mark/space frequency couples. In particular, the same filter can be programmed to the characteristics required for channels used internationally, which vary widely, country by country, from those used in Australia. In addition, many modes of transfer may be supported, such as telephone and radio analogue transmission channels, which require different FSK channel frequencies and data rates.

Further, the programming of the filter's characteristics can be performed adaptively, even synchronously within one period of the waveform being filtered, if necessary. The filter's transfer function with respect to data recovery may also be optimised adaptively during the transmission of a predetermined or random test sequence prior to the commencement of or during the transmission of the data.

Thus, not only is such a filter cheaper to manufacture than multiple discrete filters but its transfer function may be varied adaptively in order to improve data recovery from any given telephone line or other transmission media.

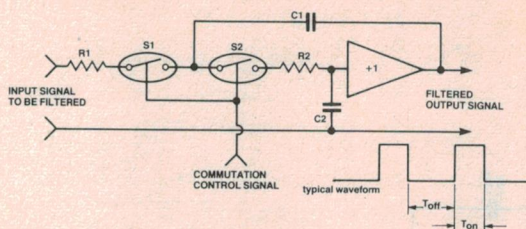


Figure 1. Typical circuit of a commutated filter. The commutation frequency is typically 50 times the signal frequency being filtered. S1 and S2 may be CMOS switches. e.g: 4016, 4066 or similar.

The filter uses active or passive filter networks (for example see Figure 1) in which the effective value of the network resistors (R1,R2) is made variable by varying the duty cycle of very fast electronic switches (S1,S2) in series with them. These electronic switches typically are commutated at a frequency greater than 20 kHz. The commutating frequency must be sufficiently large so that intermodulation components of significant amplitude are not generated in the FSK passband.

The transfer function of the circuit of Figure 1 (neglecting intermodulation component generation) is given by:

$$G(S) = \frac{S^2 C_1 C_2 R_1 R_2 \left(\frac{T_{on} + T_{off}}{T_{on}} \right) + S C_2 \left(\frac{T_{on} + T_{off}}{T_{on}} \right) (R_1 + R_2) + 1}{1}$$

It can be seen that the cutoff frequency of the filter is the same as if the filter had been made from resistors each of value:

$$R_{effective} = \frac{T_{on} + T_{off}}{T_{on}} \times R_{actual}$$

The filter networks can be highpass, lowpass or bandpass of any order (some examples are shown in Figure 2) and of widely varying topology.

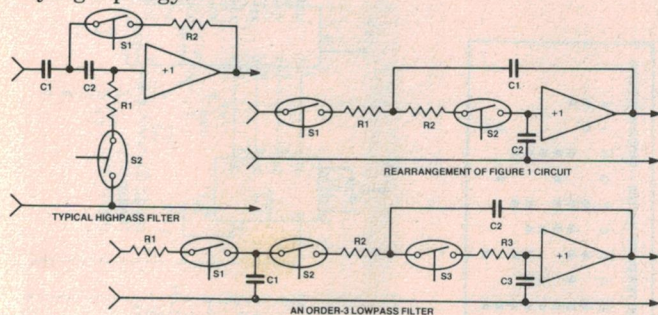


Figure 2. Examples of different implementations of the commutated filter.

It is possible to design filter networks in which the capacitors are commutated (switched capacitor filters), but they are harder (and more expensive) to make in discrete or hybrid IC form.

The capacitors in the lowpass networks may be designed to be of equal value, facilitating manufacture further, as only the mismatch between them, and not their absolute value, would be required to conform to the accuracy tolerance (typically 5%).

The recovered data is usually further processed to remove undesired frequency components from it. As the data rate may typically (but not essentially) vary from 75 BPS to 1200 BPS, a variable filter of the type disclosed by this invention, adaptive or not, can be used advantageously in such an application.

Design details

The CCITT Recommendation V.21 contains the specifications for a 200 baud modem for use on the switched public telephone network. Modems following this specification are the most commonly used variety currently in Australia.

Advancing technology has generally increased the reliable V.21 speed from 200 to 300 baud and this unit is capable of performing well at 600 baud (high band) and 450 baud (low band). The low band is 980 Hz (Binary 0, or space) to

1180 Hz FSK. The high band is 1650 Hz to 1850 Hz FSK.

Although this modem can achieve adequate data recovery at 600 baud on low band, the filter networks have been optimised for adjacent channel suppression, rather than high speed.

CCITT V.23 contains the specification for modems operating up to 1200 baud on the switched public network. The frequencies used are 1300 Hz and 2100 Hz. At a modulation rate of 1200 baud the sidebands extend down to about 300 Hz and up to about 2700 Hz, taking up the whole audio channel bandwidth. The only way to achieve faster data rates is to change to the PSK mode (usually with dedicated micro-processor control). This also places greater constraints on the line.

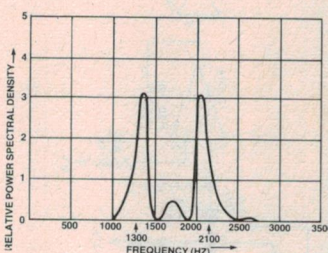


Figure 3. Relative power spectrum of a 600 baud FSK signal using 1300 Hz and 2100 Hz frequencies.

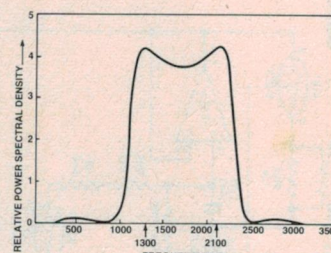


Figure 4. Relative power spectrum of a 1200 baud FSK signal using the same frequencies — modem receive filter requirements are much more stringent.

A backward channel 'for error control' has been defined as 75 baud, 390 Hz/450 Hz. Although the sidebands of the forward and backward channels overlap, provided adequate balancing of the line hybrid is achieved at both ends the two channels may be used simultaneously.

A 1200/75 baud modem would seem to be the most efficient means of transferring data primarily in one direction, which is the primary mode of data transfer currently used by hobbyists.

Both 1200/75 baud and 300/300 baud transmission modes are used with acoustically coupled equipment.

Problems with conventional technology

No modems currently available implement all the above modes due to the variety of filter requirements, amongst other things. Lowpass filters with orders of at least four are required at 500 Hz, 1200 Hz, 1950 Hz and 2500 Hz, and similarly complex highpass filters are required at 400 Hz, 900 Hz and 1550 Hz.

Conventional modem designs use a phase-locked loop to decode the incoming FSK signal back into the original binary data. A phase-locked loop uses a feedback signal from a phase comparator to vary the frequency of an internal voltage-controlled oscillator until it matches that of the input signal. Normally, a simple RC filter is employed in the feedback loop. It is the design of the feedback filter that ultimately determines the maximum data rate that can be achieved with such a decoder.

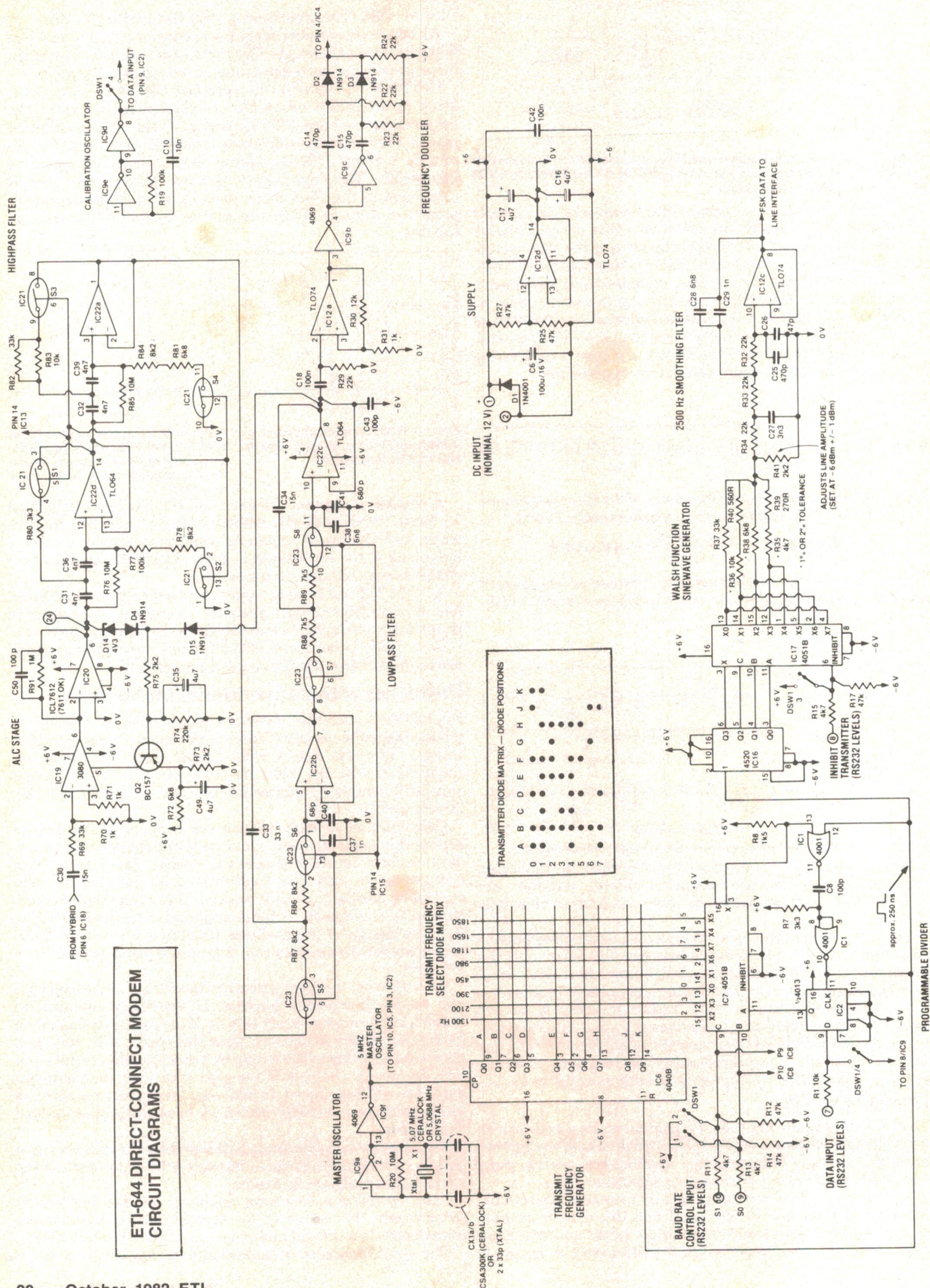
As with most feedback circuitry, the design of the loop gain and feedback transfer function is extremely complex, so optimal performance rarely results. If the gain is too high for a particular feedback network then the circuit will 'ring' and may even oscillate at the wrong frequency whenever the input FSK frequency changes. If the gain is too low the maximum data rate will be reduced. In addition, a frequency counter and much patience is usually necessary to set up a phase-locked loop decoder.

This design has eliminated this problem by employing a simple commutated variable frequency filter for each of the lowpass and highpass functions.

Also, by using a frequency doubling detector this circuit circumvents the necessity for critical design, making the components less critical and ensuring optimal performance.

The input frequency is doubled and used to trigger a mono-

ETI-644 DIRECT-CONNECT MODEM CIRCUIT DIAGRAMS



stable. A crystal-derived reference frequency is used to trigger an identical monostable. The filtered ('dc') outputs of these are then compared to determine whether the input frequency is higher than the reference or vice versa.

The only adjustment necessary is to compensate for imbalance in the two monostables. There is no feedback network and the design of the filters can follow conventional procedures.

These approaches have the additional advantage of eliminating the need for critical adjustments to ensure optimum performance. If phase-locked loop discriminators were used, each would require bandwidth and centre frequency adjustment with a CRO and frequency counter — a total of eight critical adjustments. In addition, the eight transmitter frequencies would have to be adjusted individually.

This design uses frequencies derived (phase-coherently) from a crystal reference. The output sinusoid is then generated digitally using the order-8 Walsh function synthesis (best approximation to a sinusoid).

Circuit operation

There are three major sections in the modem: the transmitter, the receiver and the line interface. In addition, a 600 baud (nominal) reference oscillator is provided on-board for setting up purposes.

This is not the place, nor is there the room, for discourses on how data multiplexers, Walsh function generators and linear-phase filters work, so this explanation assumes at least a passing knowledge of circuit techniques mentioned, the overall operation being explained with reference to the circuit diagram.

The transmitter

The master oscillator is a parallel oscillator involving IC9a, one gate from a 4069 hex inverter, the output being buffered by IC9f. Either a 5.0688 MHz quartz crystal or a Murata 5.07 MHz 'Ceralock' ceramic resonator can be used. If a crystal is employed, two 33p NPO ceramic capacitors are required, whereas if a Ceralock is used, the matching loading capacitor pack (CSC300K) is required. Provision has been made on the pc board to accommodate either alternative. The 5.07 MHz master oscillator output, which serves both the transmitter and receiver, is from pin 12 of IC9.

The master oscillator drives a 4040B 12-stage counter (IC6) used as a programmable divider. This counter is reset whenever a terminal count, set by a diode matrix, is reached. There are eight sets of diodes, one for each output frequency. A 4051 8-to-1 analogue multiplexer (IC7) connects the appropriate diodes to the input of the reset monostable in IC6 (pin 11). This multiplexer is controlled by the data input (board terminal 7) via a latch, IC2, and a 2-bit channel select code (labelled S0 and S1) applied at terminals 9 and 10 on the board. S0 and S1 may be 'set' by two switches from DSW1, obviating external control. Table 1 gives the selection logic for S0 and S1.

The reset pulses to IC6 are at 16 times the required output frequency. They are fed to the clock input of one counter of a 4520 dual 4-bit up-counter (IC16, pin 1). This is used to derive polarity and amplitude information for the transmitter sinusoid signal synthesis. The outputs of the 4520 control a 4051 data selector (multiplexer, IC17), the outputs of which generate a stepped sinusoid. This is the Walsh function generator.

A linear-phase filter with a cutoff of 2500 Hz follows. This involves IC12c (one op-amp from a TL074) and surrounding components. This acts as a smoothing filter to remove the steps from the waveform and any out-of-passband components which may be present.

Transmitter output is from pin 8 of IC12c and goes to the line interface circuitry.

The receiver

The centre channel frequencies are derived from the master oscillator via a 4020B counter (IC5) and half of a 4052 dual 4-to-1 multiplexer (IC8), controlled by the 2-bit channel select signals (S0 and S1, to pins 9 and 10 of IC8). The same reset-on-terminal-count scheme as used in the transmitter is implemented here.

The input signal from the line interface is buffered by an amplifier (IC18) and passed to an automatic level control (ALC) stage involving IC19, a 3080 transconductance amplifier, the output of which is buffered by an Intersil ICL7612 (or 7611) low power, high input impedance op-amp capable of rail-to-rail output drive. The transconductance control signal for the 3080 is derived from two feedback paths. The first feedback path is from the output of the receiver filters, from pin 8 of IC22c, via D15. This provides base current to Q2, the collector current of which controls the current through pin 5 of the 3080, the transconductance control pin, thus varying the gain. This feedback path operates for low-level signals up to a certain threshold where the output of IC20 exceeds the zener voltage of D14 and the second feedback path comes into operation, further increasing the collector current of Q2, decreasing the gain of the 3080, IC19. This feedback path acts to prevent clipping on the output of IC20.

The ALC loop adapts the received signal strength such that a signal level of about 1.5 V peak-to-peak results on pin 8 of IC22c.

The receiver filters comprise two 0.5 dB Chebychev commutated filters, one highpass involving IC22 a and d, and one lowpass involving IC22 b and c. These remove unwanted signal components. A comparator with hysteresis, IC12a, squares up the signal for the frequency doubling circuitry. This involves two inverters from IC9 (b and c). Two signals are derived by these gates, 180° out of phase and they drive a two-diode multiplier (i.e. a full-wave rectifier) the output of which goes to the data channel input of the frequency comparator (to pin 4, IC4a).

The commutated filters in the receiver are controlled by variable pulse width signals derived digitally from three 4-bit adders, IC11-IC13-IC15 (all 4008). A scan signal at 2.5 MHz, derived from the master oscillator by IC2, a 4013, is supplied to the B inputs of these adders from a 4520 dual 4-bit synchronous counter (IC16). The adder A inputs are programmed via a diode matrix. The 2-bit channel control signal (S0, S1) is decoded by IC8, the Y outputs driving the diode matrix to the 4-bit adders. If, say, the A inputs of an adder were programmed with 1110 then the carry output would be inactive for scan counts of 0 or 1, but would be active (true) for the rest of the count cycle. Thus, the output pulse width can be preprogrammed digitally via the diode matrix and optimised without the necessity for many high stability components.

The frequency comparator is implemented with a 4528 dual monostable flip-flop, IC4, two commutated filters and a voltage level comparator.

The twice-frequency pulses from the receiver trigger monostable IC4a. The output of this is cleaned up by an order-3 commutated lowpass filter involving IC14a, one op-amp from an LM324, to remove components of the resultant signal other than the data. The reference pulses are processed identically by IC4b and IC14d, to ensure matching over the entire range of input frequencies. IC14b provides the comparator and output is at RS232 levels via terminal 6 on the board. Adjustment of the output symmetry is provided by varying monostable IC4b by means of RV1. Note that, for proper matching of the data and reference channel filters, switches S9/S12, S10/S13 and S11/S14 must each be in the same physical package.

A 'carrier received' signal is provided via IC14 and Q1. The latter drives a LED which lights whenever a tone above 350 Hz is present at the input of the modem. Terminal 5 on the board

provides a 'carrier received' signal output for external use.

Line interface

The line interface couples the modem to the line and provides line switching, answer and dialling facilities.

Signals are coupled to and from the line via an Arlec 600 ohm to 600 ohm line isolating transformer, type 45035 (Telecom approval No. RA81/144). The isolated winding of T1 (pins 6 and 10) is in one 'leg' of a bridge hybrid which separates transmit and receive signals. This consists of R68 and the isolated winding of T1 on one side, and R79/R108 on the other side. The transmit signal is applied to the 'top' of the bridge at the junction of R68 and R79. The received signal is taken from the nodes of the bridge at the junctions of R68/T1 and R79/R108 via a TL071 op-amp, IC18. Sidetone level is adjusted by varying the negative feedback via RV2 which may be a front panel control if desired, to cope with week-to-week variations in line balance.

A full bridge network is used to balance lines of uncertain impedance, particularly those exhibiting capacitive or inductive reactance. A capacitor may be added to compensate for either case. Usually, lines will be found to be capacitive, in which case C46 should be added. The exact value should be determined experimentally. A value of 10n is a good starting point. For inductive lines, C45 should be added. Again, start with a value of 10n. One or other of these capacitors should be added and the value found which gives the deepest sidetone null when adjusting RV2. This is best done using an audio monitor (see later) or an ac voltmeter connected to the output of IC18 (terminal 23) and *must be performed while the modem is on*.

Line switching is provided by a DPDT relay, RL1. This connects the line to a standard telephone, or whatever appliance is installed in cases where a telephone line is not used (e.g. a radio transceiver), whenever the modem or computer is de-powered. (Note: Telecom approval is required before attaching the modem to Telecom lines.)

Relay RL4 is normally closed and relay RL3 is normally open, except when dialling. When dialling under computer control RL3 is first closed, then RL4 is pulsed with the appropriate sequence. RL3 is opened after dialling is completed.

Relay RL2 is used to 'answer' a call. It is normally open, but when a ring is detected by the ring detector circuit, RL2 closes to 'answer' the call.

The ring detector circuit consists of a rectifier and voltage comparator. Diode D13 rectifies the ring pulses coming via T1, and charges C24. When the charge on C24 exceeds 3 V, the output of IC12b goes towards the positive supply rail (high), providing a 'ring detect' signal at terminal 11 on the board. If so desired, C24 may be omitted, in which case the 'ring detect' output will toggle high and low when ringing pulses are received.

Adjustments

A 600 baud (nominal) reference oscillator is provided on-board for setting up purposes. This is a simple two-inverter oscillator using two gates from IC9 (e and d). This connects to the transmitter data input, at pin 9 of IC2, via DSW1/4.

In setting up the modem, the transmitter and receiver are both programmed for 300 baud operation (high channel, 1750 Hz) and DSW1/4 turned on. A dc voltmeter is connected between terminal 3 (data output) and terminal 4 (which provides a reference of about -0.5 V). Trimpot RV1 is then adjusted until the dc voltmeter reads close to zero. If a CRO is available, then the data output should be adjusted such that the output signal has 50-50 symmetry.

The sidetone level adjust trimpot, RV2, is set whilst the modem is 'on line' so that the level of any signal present at the output of IC20 (terminal 24) due to the transmitted signal is

nullled. This can best be done by connecting a dc voltmeter across C35 (across the emitter resistor of the ALC control transistor, Q2). Adjust RV2 such that this voltage is minimised. This procedure should be done in conjunction with the line balance.

The transmitter line level is best set at around -12 dBm, from experience. This can be effected by varying the value of R41. If you wish, a multiposition switch can be used to provide line levels of -6, -12 and -18 dBm to cope with differing line conditions.

Computer control

It is expected that this modem will be controlled by a computer. Hence, modem control algorithms need to be developed for effective use of the modem. There are two modes of operation: answer and originate. (Note: Telecom approval is required before attaching the modem to Telecom lines.)

Answer mode

It is expected that the computer will energise RL1 and RL4, and de-energise RL2, RL3 and the transmitter and wait for an output from the ring detector. When a valid ring has been obtained RL2 will be closed to answer the call.

Each baud rate will be selected in turn until a valid carriage return signal has been obtained. The transmitter will then be energised and another two valid carriage returns sought. If they are not obtained within a time-out period it will be assumed that the transmitter has 'blanketted' the receiver signal and the remote user will be notified to try another line (at the current baud rate) and then disconnected.

If all is OK the communication session can begin.

Originate mode

The computer will energise RL1, RL4 and RL3. RL4 will then be pulsed in the proscribed 2:1, 100 msec nominal manner until the number to be dialled has been completed. RL2 will then be energised, RL3 de-energised and the transmitter enabled.

Monitoring

Two monitors are recommended: an audio monitor and a signal meter. Both are quite simple to implement, but have not been included on the printed circuit board.

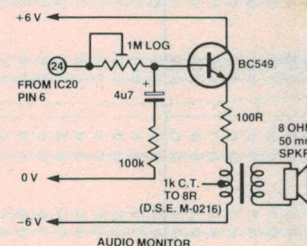


Figure 5. Suggested circuit for an audio monitor.

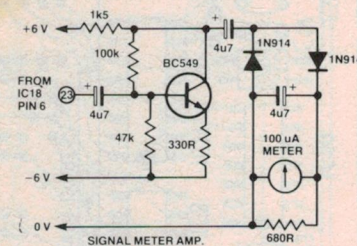


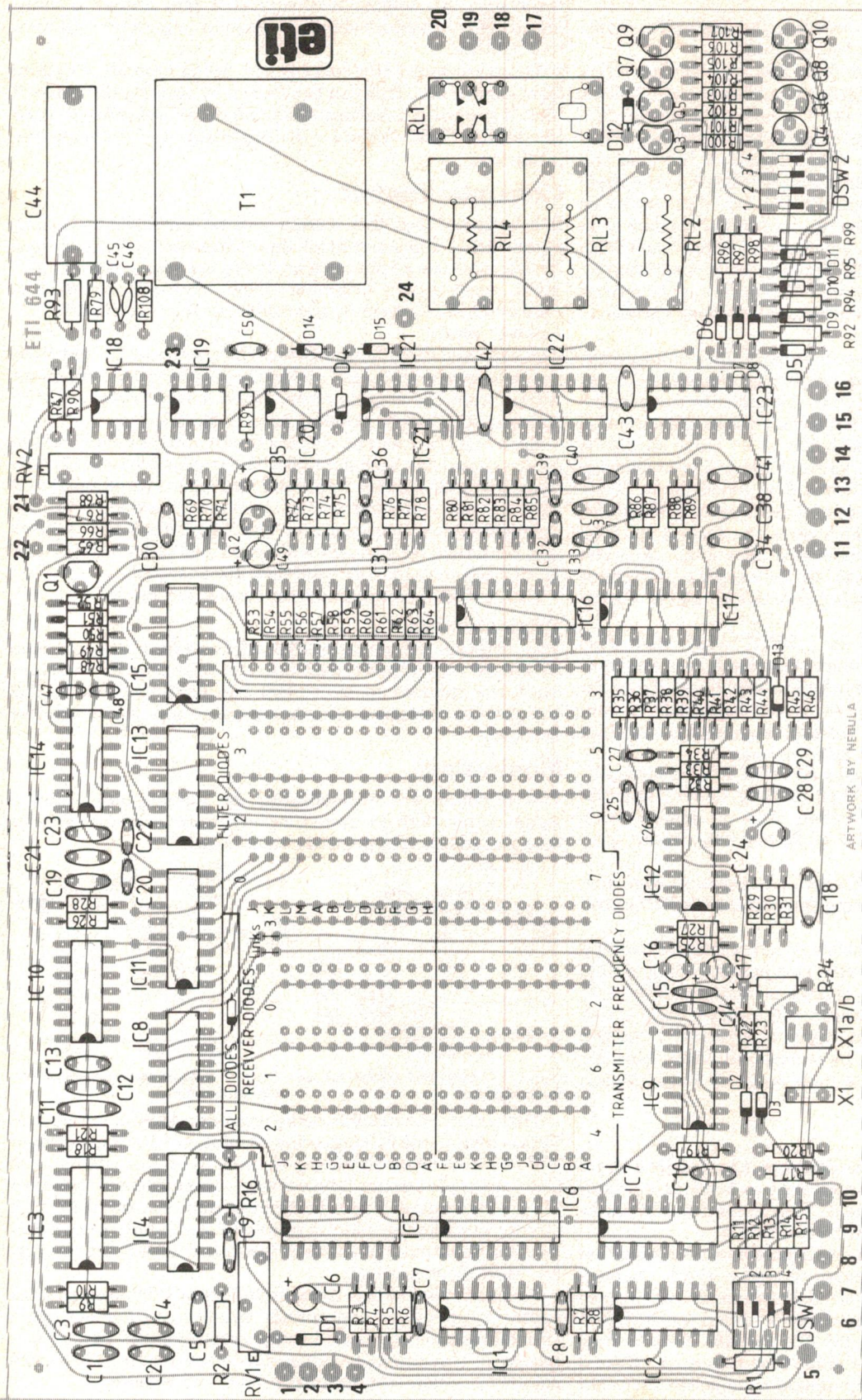
Figure 6. Suggested circuit of a signal strength meter.

The circuit of a suitable audio monitor is shown in Figure 5. This is invaluable in monitoring the status of dialling software and to assist when balancing the line hybrid (selecting C45/C46). A simple emitter follower takes output from terminal 24 (ALC stage output), gain being adjustable via the 1M pot. A 50 mm diameter speaker is coupled via a small 1k-to-8 ohm 'transistor output' transformer (such as the Dick Smith type, No. M-0216, or similar).

A signal meter circuit is shown in Figure 6. This comprises a simple ac amplifier and voltage doubler rectifier driving a 100 microamp meter. Input comes from terminal 23 (audio to receiver). This circuit is very useful for setting the line balance pot. (RV2). A low cost miniature meter can be used.

Construction

As the circuitry is complex, construction on a double-sided plated-through hole pc board is recommended. In fact, the final ▶



ETI-644 DIRECT-CONNECT MODEM COMPONENT OVERLAY

Note: only the top-side tracks are shown for the sake of clarity.

| S1 | S0 | TRANSMIT CHANNEL | RECEIVE CHANNEL |
|----|----|------------------|---------------------------|
| 0 | 0 | 75 BAUD | 1200 BAUD |
| 0 | 1 | 1200 BAUD | 75 BAUD |
| 1 | 0 | 300(600) | 300 BAUD 'ANSWER' MODE |
| 1 | 1 | 300 | 300(600) 'ORIGINATE' MODE |

TABLE 1.
BAUD RATE
CONTROL LOGIC

- DSW1/1 S0 > BAUD RATE
- DSW1/2 S1 > SELECT (ON=HI)
- DSW1/3 INHIBIT TRANSMITTER (ON=INHIBIT)
- DSW1/4 CALIBRATION OSC. (ON=CALIBRATE)
- DSW2/1 OPERATE RL1 (MODEM ON LINE)
- DSW2/2 OPERATE RL2
- DSW2/3 OPERATE RL3
- DSW2/4 OPERATE RL4

TABLE 3. DIP SWITCH FUNCTIONS

Resistors all 1/2 W, 5% unless noted

| | | |
|-----|---------|-----------------------|
| R1 | 10k | 1k |
| R2 | 270k | 6k8 |
| R3 | 12k | 2k2 |
| R4 | 10k | 220k |
| R5 | 10k | 2k2 |
| R6 | 10k | 10M |
| R7 | 4k7 | 100k |
| R8 | 3k3 | 8k2 |
| R9 | 10k | 4k7 |
| R10 | 10k | 3k3 |
| R11 | 4k7 | 6k8 |
| R12 | 47k | 3k3 |
| R13 | 47k | 3k |
| R14 | 47k | 3k |
| R15 | 47k | 10M |
| R16 | 220k | 8k2 |
| R17 | 47k | 8k2 |
| R18 | 10k | 7k5 |
| R19 | 100k | 12k |
| R20 | 10M | 1M |
| R21 | 10k | 10k |
| R22 | 22k | 560R |
| R23 | 22k | 10k |
| R24 | 22k | 10k |
| R25 | 47k | 560R or link |
| R26 | 10k | 560R or link |
| R27 | 47k | 560R or link |
| R28 | 10k | 10k |
| R29 | 22k | 10k |
| R30 | 12k | 10k |
| R31 | 1k | 10k |
| R32 | 22k | 10k |
| R33 | 22k | 4k7 |
| R34 | 22k | 4k7 |
| R35 | 4k7, 2% | 4k7 |
| R36 | 10k, 2% | 100k 10-turn trimpot. |
| R37 | 33k, 2% | 50k 10-turn trimpot. |
| R38 | 6k8, 2% | (or rotary pot.) |
| R39 | 270R | |
| R40 | 560R | |
| R41 | 2k2 | |
| R42 | 2k2 | |
| R43 | 220k | |
| R44 | 220k | |
| R45 | 10k | |
| R46 | 10k | |
| R47 | 47k | |
| R48 | 68k | |
| R49 | 68k | |
| R50 | 2k2 | |
| R51 | 10k | |
| R52 | 3k3 | |
| R53 | 10k | |
| R54 | 10k | |
| R55 | 270R | |
| R56 | 2k2 | |
| R57 | 47k | |
| R58 | 560R | |
| R59 | 33k | |
| R60 | 1k | |

Capacitors

| | | |
|-----|-----------------------|-----------------------|
| C1 | 33n greencap | 33n greencap |
| C2 | 15n greencap | 15n greencap |
| C3 | 3n3 ceramic | 3n3 ceramic |
| C4 | 1n ceramic | 1n ceramic |
| C5 | 1n5 styro or mica | 1n5 styro or mica |
| C6 | 100u/16 V RB electro. | 100u/16 V RB electro. |
| C7 | 100p ceramic | 100p ceramic |
| C8 | 100p ceramic | 100p ceramic |
| C9 | 1n5 styro or mica | 1n5 styro or mica |
| C10 | 10n ceramic | 10n ceramic |
| C11 | 100n greencap | 100n greencap |
| C12 | 4n7 greencap | 4n7 greencap |
| C13 | 33n greencap | 33n greencap |
| C14 | 470p ceramic | 470p ceramic |
| C15 | 470p ceramic | 470p ceramic |
| C16 | 4u7/16 V RB electro. | 4u7/16 V RB electro. |
| C17 | 4u7/16 V RB electro. | 4u7/16 V RB electro. |
| C18 | 100n greencap | 100n greencap |
| C19 | 22n ceramic | 22n ceramic |
| C20 | 10n ceramic | 10n ceramic |
| C21 | 33p ceramic | 33p ceramic |
| C22 | 10n ceramic | 10n ceramic |

| | | |
|--------|---|---|
| C23 | 6n8 greencap | 6n8 greencap |
| C24 | 4u7/16 V RB electro. | 4u7/16 V RB electro. |
| C25 | 470p ceramic | 470p ceramic |
| C26 | 47p ceramic | 47p ceramic |
| C27 | 3n3 greencap | 3n3 greencap |
| C28 | 6n8 greencap | 6n8 greencap |
| C29 | 1n ceramic | 1n ceramic |
| C30 | 15n greencap | 15n greencap |
| C31 | 4n7 greencap | 4n7 greencap |
| C32 | 4n7 greencap | 4n7 greencap |
| C33 | 33n greencap | 33n greencap |
| C34 | 15n greencap | 15n greencap |
| C35 | 4u7/16 V RB electro. | 4u7/16 V RB electro. |
| C36 | 4n7 greencap | 4n7 greencap |
| C37 | 1n ceramic | 1n ceramic |
| C38 | 6n8 greencap | 6n8 greencap |
| C39 | 4n7 greencap | 4n7 greencap |
| C40 | 68p ceramic | 68p ceramic |
| C41 | 680p ceramic | 680p ceramic |
| C42 | 100n ceramic (blue chip) | 100n ceramic (blue chip) |
| C43 | 100p ceramic | 100p ceramic |
| C44 | 2u2 greencap or PETP type, 100 V | 2u2 greencap or PETP type, 100 V |
| C45 | 10n greencap (see text) | 10n greencap (see text) |
| C46 | 10n greencap (see text) | 10n greencap (see text) |
| C47 | 100n ceramic (blue chip) | 100n ceramic (blue chip) |
| C48 | 100n ceramic (blue chip) | 100n ceramic (blue chip) |
| C49 | 4u7/16 V RB electro. | 4u7/16 V RB electro. |
| C50 | 100p ceramic | 100p ceramic |
| CX1a/b | 2 x 33p NPO ceramic if xtal used, or CSC300K Ceralock loading capacitor (see text). | 2 x 33p NPO ceramic if xtal used, or CSC300K Ceralock loading capacitor (see text). |

Semiconductors

| | | |
|--|---|---|
| D1 | 1N4001 | 1N4001 |
| D2 to D13, D15 | 1N914 | 1N914 |
| D14 | 4V3, 400 mW zener | 4V3, 400 mW zener |
| (note, 1N914 diodes will be needed for programming the diode matrices) | | |
| IC1 | 4001 | 4001 |
| IC2 | 4013 | 4013 |
| IC3, 10, 21, 23 | CD4016CN or MM5616AN | CD4016CN or MM5616AN |
| IC4 | 4528B | 4528B |
| IC5 | 4020B | 4020B |
| IC6 | 4040B | 4040B |
| IC7, IC17 | 4051B (not Motorola type) | 4051B (not Motorola type) |
| IC8 | 4069 | 4069 |
| IC9 | 4088B | 4088B |
| IC11, 13, 15 | TL074, uA774 etc. | TL074, uA774 etc. |
| IC12 | LM324 | LM324 |
| IC14 | 4520 | 4520 |
| IC16 | TL071, uA771 etc. | TL071, uA771 etc. |
| IC18 | CA3080, LM3080 | CA3080, LM3080 |
| IC19 | ICL7612 or 7611 | ICL7612 or 7611 |
| IC20 | TL064 (preferred, otherwise use TL074, uA774) | TL064 (preferred, otherwise use TL074, uA774) |
| IC22 | TIL220 orange LED | TIL220 orange LED |
| LED1 | BC549 | BC549 |
| Q1 | BC157 | BC157 |
| Q2 | BC559 | BC559 |
| Q3, 5, 7, 9 | | |

Miscellaneous

| | |
|------------|---|
| DSW1, DSW2 | 4-way SPST DIP switches |
| RL2, 3, 4 | Fujitsu FRL644E05/1AK (5 V) or FRL644E12/1AK (12 V) micro reed relays, Telecom approval |
| RA81/130 | Fujitsu FRL621D012, 12 V DPDT pc mount relay, Arlec 600/600 ohm isolating transformer 45035, Telecom approval |
| RL1 | RA81/144 |
| T1 | 5.0688 MHz HC18/U crystal or CSA5.07MT Ceralock resonator. |
| X1 | |

ETI-644 pc board; 12 Vdc plugpack, 200 mA or greater rating or similarly rated power supply.

Price estimate

\$160 — \$170

(excludes power supply)

TABLE 2. TERMINAL DESIGNATIONS

| TERMINAL | USE |
|----------|---|
| 1 | +VE INPUT, 12 V SUPPLY |
| 2 | -VE INPUT, 12 V SUPPLY |
| 3 | TO METER +VE OUTPUT |
| 4 | TO METER -VE OUTPUT |
| 5 | 'CARRIER RECEIVE' OUTPUT |
| 6 | DATA OUTPUT (RS232 LEVELS) |
| 7 | DATA INPUT (RS232 LEVELS) |
| 8 | INHIBIT TRANSMITTER (HI=INHIBIT) |
| 9 | S0 BAUD RATE CONTROL |
| 10 | S1 INPUT (RS232 LEVELS) |
| 11 | 'RING DETECT' OUTPUT |
| 12 | 0 V (BOARD COMMON) |
| 13 | DRIVE RL4 (HI TO OPERATE) |
| 14 | DRIVE RL3 (HI TO OPERATE) |
| 15 | DRIVE RL2 (HI TO OPERATE) |
| 16 | DRIVE RL1 (HI TO OPERATE) |
| 17 | TO APPLIANCE NORMALLY CONNECTED TO LINE |
| 18 | TO APPLIANCE NORMALLY CONNECTED TO LINE |
| 19 | LINE INPUT |
| 20 | LINE INPUT |
| 21 | LED 1 ANODE |
| 22 | LED 1 CATHODE |
| 23 | RECEIVED AUDIO (TO SIGNAL STRENGTH MONITOR) |
| 24 | AUDIO MONITOR OUTPUT |

(Note: Telecom approval required before attaching modem to Telecom lines.)

prototype boards employed this sort of construction with the addition of solder masking, tinned tracks and a silk-screened component overlay. The board measures 223 x 136 mm and mounting holes are located near each of the four corners.

As some 23 ICs, 50 capacitors and over 100 resistors plus quite a number of other components are used, the component overlay has an orthogonal numbering scheme so that components are readily identified and located with the aim of considerably reducing errors in assembly. Component numbering commences at the top left hand corner of the board. Numbers are allocated in ascending order from top to bottom, left to right across the board. Revisions of earlier prototypes, necessitated from experience in actual use, have required minor changes to be made and there are several exceptions to this component numbering convention. Components out of order are as follows:

| | |
|-----------------------------|------------------------------|
| R47 right of RV2 | C49 left of Q2 |
| R79 below R93 | C50 right of IC19, 20 |
| R108 below R79 | D13 between R44,45 |
| C45,46 right of IC18 | D14 right of IC20 |
| C47,48 right of IC14 | D15 right of IC21 |

The relays are numbered to correspond with the switch numbers on DSW2. Twenty four terminals are provided for data input, data output, programming control, monitoring and supply. These are located around the edge of the board,

numbered clockwise from the left hand side, with the exception of terminals 23 and 24. Terminal 23 is located to the right of IC19, 24 to the right of IC21.

The three diode matrices are located adjacent to the relevant circuitry with the numbered columns running vertically and the lettered rows running horizontally. All diodes inserted should have their cathode (banded) ends facing toward the left. Details on programming the matrices are given later.

Prototypes were constructed using sockets for the ICs to facilitate debugging and performance checking. They aren't essential but can be handy.

Actual assembly is fairly straightforward. A temperature-controlled soldering iron with a narrow wedge tip is recommended for best results with soldering. If you're using IC sockets, these should be assembled and soldered in place first. All the resistors and capacitors can follow, otherwise, they come first. Watch orientation on the electrolytic capacitors. The component overlay indicates the positive lead with a small '+'. Note that links may be needed in lieu of resistors for R96, R97 and R98 — see relay mounting details later.

Leave out C45 and C46, these won't be necessary until actually setting up the modem. You'll only need one or the other anyway. Alternatively, as most lines seem to exhibit a capacitive reactance (from experience), you could take a punt and put in a 10n at C46.

Leave C44 till last. Note that this should be a metallised

MODEMS AND THE LAW

At present, the Australian Telecommunications Commission ("Telecom"), by virtue of the Telecommunications Act 1975 ("The Act"), has control of what may be connected to the public telephone network. It is an offence under the Act to attach any apparatus other than an approved device or an appliance rented from Telecom to a telephone line. There is nothing wrong with constructing and using the modem described here provided you do not attach it to Telecom lines without approval from Telecom. It could be used by radio amateurs for computer communications by radio, for example, or the modem could be used over an internal intercom or other cable system. Here is what the Act has to say about the subject:

"Section 3 (1)

"telecommunications installation" means —

- (a) a line; or
- (b) any equipment, apparatus, structure, tower, antenna, tunnel, man hole, pit or pole used, or intended for use, in connexion with a telecommunications service;

"telecommunications service" means —

- (a) a service for transmitting, by means of electric or electro-magnetic energy —
 - (i) sounds, including speech and music;
 - (ii) visual images;
 - (iii) signals for the communication, whether as between persons and persons, things and things or persons and things, of any matter otherwise than in the form of sounds or visual images; or
 - (iv) signals for the actuation or control of machinery or apparatus; or

- (b) a service for receiving any such sounds, images or signals that have been transmitted by means of electric or electro-magnetic energy;

"telecommunications system" means a system controlled by the Commission in connexion with the provision of a telecommunications service

Section (11)

- (1) The Commission may, from time to time, make, with the approval of the Minister determinations fixing or varying —
 - (a) The rental payable in respect of standard telephone services provided by the Commission;
 - (b) The charges for telephone calls made within Australia, other than charges for special services provided by the Commission in connexion with those calls; and
 - (c)
- (2) The Commission may, from time to time, make determinations fixing or varying rentals and charges, other than rentals and charges referred to in Sub-Section (1), for Telecommunications services and other services that the Commission provides under this Act.

Section 13

(1) The Commission —

- (a) may authorise a person to erect, maintain or operate a telecommunications installation other than an installation for the purpose of transmitting or receiving messages by means of wireless telegraphy; and
- (b) may authorise the attachment of a line, equipment or apparatus, including equipment or apparatus for the purpose of transmitting or receiving messages by means of wireless telegraphy, to a telecommunications system.

(2) The Commission may specify, in an authorisation issued under Sub-Section (1) the period in respect of which, and the terms and conditions subject to which, the authorisation is to operate.

Section 87

(1)

(2) A person shall not, by means of an apparatus or device connected to a telecommunications installation belonging to the Commission or used in connexion with a telecommunications system —

- (a) defraud the Commission of any rental, fee or charge properly payable for the use of a telecommunications service; or
- (b) cause the Commission to provide a telecommunications service to some other person without payment by that other person of the appropriate rental, fee or charge.

Penalty: Imprisonment for five years.

Section 90

A person shall not wilfully damage, deface, interfere with, remove or destroy a telecommunications installation or any part thereof belonging to the Commission. Penalty: Imprisonment for two years.

Section 94

- (1) Subject to Sub-Sections (2) and (3) a person other than the Commission
 - (a) Shall not erect, maintain or operate a telecommunications installation within Australia; or
 - (b) Attach a line, equipment or apparatus to a telecommunications system

Penalty: Imprisonment of two years.

(2)

(3) Sub-Section (1) does not apply to the attachment of a line, equipment or apparatus to a telecommunications system to the extent that the attachment is authorised by the Commission under Section 13 and the terms and conditions subject to which it is so authorised are complied with.

(4) The Commission may authorise a person to take possession of, remove or destroy the whole or any part of a telecommunications installation established, maintained or operated in contravention of this section and may enter on premises at any reasonable hour of the day for that purpose."

Those wishing to seek approval for the use of this modem on the public telephone network should contact the Telecom head office in the capital city of their state. A fee is payable for type approval testing.

mylar ('greencap') or metallised polyester (PETP) type rated at 100 V or 250 V with a 'self-healing' breakdown characteristic.

Special mention must be made of CX1a/b. If a quartz crystal is used, then two 33p NPO ceramic capacitors are required. These mount to the right of X1, parallel to the long axis of the board — you'll see two component pads to the right of the box marked 'CX1a/b'. If you are using a Ceralock resonator (made by Murata), then its accompanying loading capacitor pack — CSA300K — mounts inside the box marked 'CX1a/b'.

All the diodes and transistors may be assembled next, but take care with diode orientation. The two ten-turn trimpots, RV1 and RV2, may be mounted next, followed by X1 (quartz crystal or Ceralock resonator) and the two 4-way DIP switches (DSW1 and DSW2). Watch which way around you mount these last two components.

The three micro reed relays may now be mounted. Note that their connections are marked on the case and the pc board component overlay shows the same markings to indicate correct orientation. These relays are available with coils for either 5 V or 12 V operation. If you have the 5 V types, then resistors R96, R97 and R98 will be necessary (560R each), otherwise links should be inserted in these resistor positions on the board.

With the micro reed relays in place, you can mount the double-pole changeover relay and the line isolating transformer, T1, last of all.

Note that there are two links required. These are located in the receiver diode matrix area, at the top of column 3.

Having completed this phase of construction, check the board thoroughly, *both sides*. Correct any errors you may find and see that all soldered joints are properly made and that there are no solder bridges between adjacent pads. A solder-masked board generally obviates the latter problem.

If all is OK, plug in IC12 *only*, set all switches in DSW1 & 2 open, and apply power. Check the power supply rails with a multimeter for correct operation. Operate each of the switches in DSW2 in turn and see that the corresponding relay operates: DSW2/1 operates RL1, DSW2/2 operates RL2 etc. Trace and correct any faults.

Power down and then insert all of the ICs, taking care to orient them correctly — all 'vertical' ICs have pin 1 facing the top of the board, all 'horizontal' ICs have pin 1 facing the left hand side. Use the usual handling precautions with the CMOS ICs. The modem can then be tried out by carrying out the operations listed under the heading 'Adjustments'.

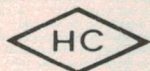
PRINTED CIRCUIT BOARDS

Readers wishing to construct this modem can purchase printed circuit boards by mail order direct from ETI. We have retained copyright on the pc board design and boards have been manufactured for us. Note that a number of electronics retailers are stocking complete kits for this project — see the Shoparound page in this issue.

If you wish to purchase a board only, they cost \$50 each (post free). Send a cheque or money order for the number of boards you require to:

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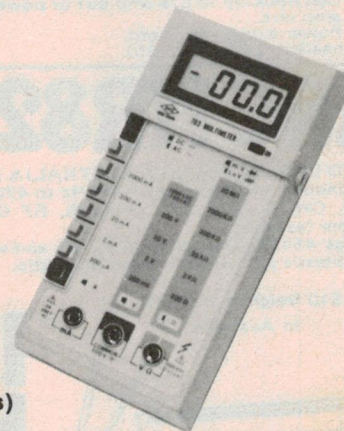
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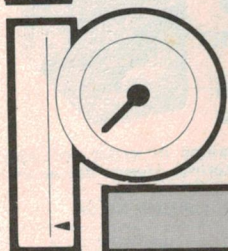
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Ref: Electronics Australia June-August 1982

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NB. The photo shows the prototype which was finished in white. The production units are only available in black. Freight anywhere in Australia only \$10.00.

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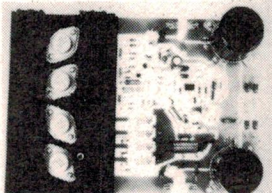
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***** HE114 also required for complete operation.

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* * NEW KIT * *

Ref: Electronics Australia October 1982.

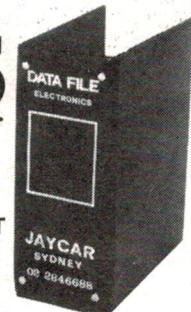
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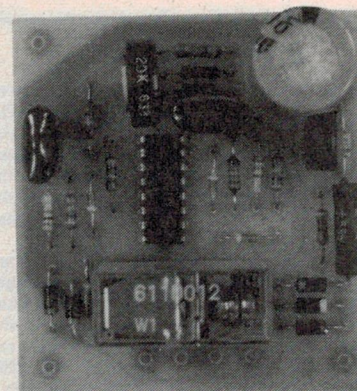
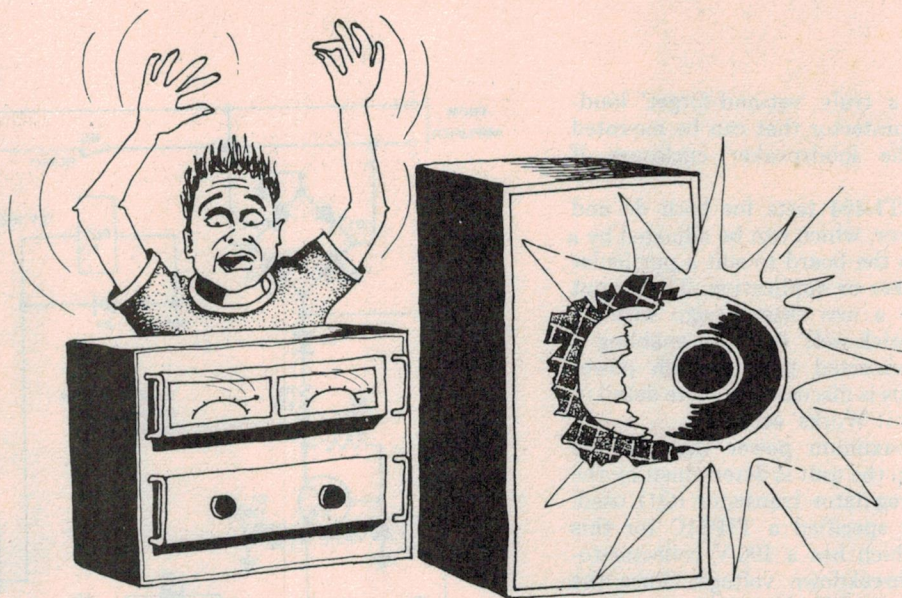
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Signal powered loudspeaker protector

David Tilbrook



This unit affords both dc and over-power protection of loudspeakers or loudspeaker systems rated at up to 1500 watts! The unit requires no power supply and has no discernible audible effect on sound quality making it suitable for both hi-fi and sound reinforcement applications.

THE ETI-455 loudspeaker protector has proved to be a very popular project. It was published in March 1980 and since then we have had numerous phone calls from readers with stories of how the unit had saved their loudspeakers from almost certain disaster. Usually the power amplifier had gone faulty and applied the full dc supply rail to the loudspeaker terminals. Without the loudspeaker protector in circuit the result would be at least an open circuit bass driver and probably worse. The protector prevents this by monitoring the loudspeaker lines for the presence of dc, opening a set of relay contacts if this occurs, disconnecting the loudspeaker from the faulty amplifier.

The ETI-455 works well but requires its own power supply, either batteries or a small regulated mains supply. Another disadvantage results from the type of filter used to distinguish between dc and

the audio signal. This was a conventional passive filter set to around 10 Hz. The problem is that it is still possible with very large amplifiers to trigger the protector on low frequency audio content. So the circuit, although perfectly satisfactory for its quoted maximum power of around 100 watts or so, is unsuitable for very high powered amplifiers.

We decided to overcome these limitations in this new loudspeaker protector, the ETI-494. Since the old one was published we have had numerous requests for a circuit that could be mounted inside a loudspeaker enclosure. These requests have come largely from the sound reinforcement industry although the unit would obviously be applicable to all loudspeakers. The protector would not be able to be powered from a mains supply since it is not always desirable or even possible to

run mains to the loudspeakers. This is particularly true in a sound reinforcement or public address system. Similarly, batteries are unsuitable since access would have to be provided to facilitate testing and changing them when required. In addition, when we published the ETI-499 MOSFET P.A. Module (March 1982), we promised to follow up with a loudspeaker protector. This is it. The solution, used in this project, is to power the unit from the audio signal itself.

This is done in this case by placing a fullwave rectifier across the speaker lines and charging a 1000u capacitor through a 47 ohm resistor. The worst possible load presented to the speaker line is therefore 47 ohms and this is only while charging the capacitor and for signal voltages in excess of 12 V. This ensures that the unit has no discernible effect on audio quality but makes ►

Project 494

possible a truly 'set-and-forget' loud-speaker protector that can be mounted inside the loudspeaker enclosure if desired.

The ETI-494 tests for both dc and over-power, which can be adjusted by a preset on the board to suit a particular loudspeaker or application. The circuit also uses a new filter design with an almost 'brick wall' response enabling it to be connected to very high power amps. This is discussed in more detail in the 'How it Works' section.

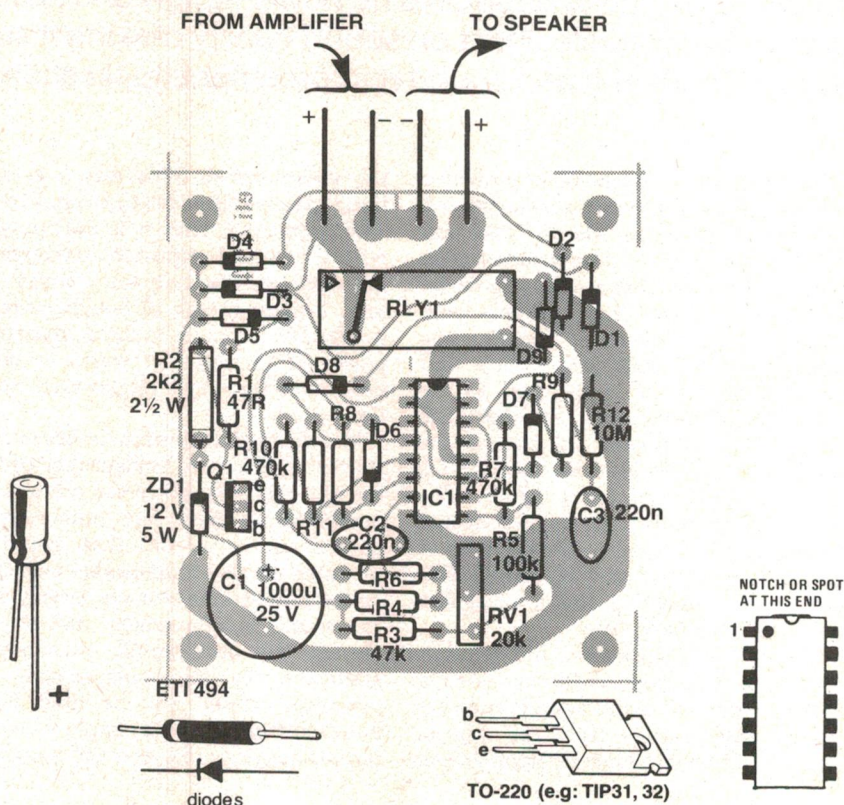
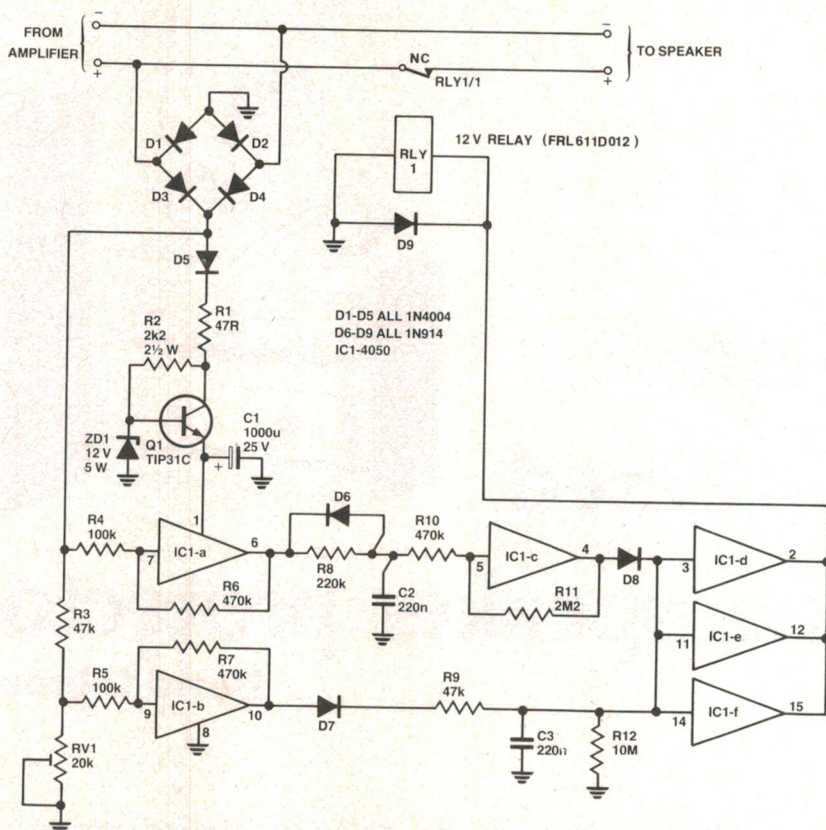
The maximum power that can be applied to the unit is determined by the type of regulator transistor (Q1) used. We have specified a TIP31C for this device which has a 100 V collector-to-emitter breakdown voltage. Since the emitter is at 12 V, the maximum voltage that can be applied to the unit is 112 V. This is equivalent to an amp capable of

HOW IT WORKS — ETI 494

The signal from the power amp is rectified by the fullwave rectifier formed by D1-D4. The output of this is fed through a 12 V regulator circuit formed by Q1 and its associated resistors and zener diode, and charges the electrolytic capacitor, C1. The output of the rectifier is also fed to the input of the dc sense and over-power detection circuitry.

IC1 gates a and c form the dc filter. Resistors R4 and R6 form a Schmitt trigger with a small deadband. When the signal goes above the trigger voltage the output of the trigger swings hard to the positive supply rail of the IC, charging C2 through the 220k resistor, R8. Resistors R10 and R11 with gate c form a second Schmitt trigger monitoring the voltage across C2. If the voltage across C2 reaches the trigger voltage of this second Schmitt, gates d, e and f are activated, pulling in the relay contacts and disconnecting the loudspeaker. It takes about 100 ms to charge C2 through R8, and on normal audio content the output of gate 'a' will be driven low before this occurs, discharging C2 rapidly through D6. Only signals which do not have a zero crossing for longer than 100 ms will trigger the protector.

The over-power protector consists simply of a voltage divider feeding a third Schmitt trigger. Whenever the signal voltage exceeds the trigger voltage the output of gate 'b' is driven high and C3 starts to charge. If this condition persists for long enough the output gates are turned on and the relay pulls in. Note that both the dc and over-power sense circuits charge C3 when activated. The circuits are decoupled from this capacitor by diodes so that, once charged, C3 can only be discharged by the parallel resistor R12 (the effect of the input impedance of the gates is negligible). Since it takes about one second to discharge this capacitor, the relay will hold in for this time. The protector therefore reconnects the loudspeaker approximately one second after the fault condition has been removed.



PARTS LIST — ETI-494

Resistors

| | |
|-----------|------------------|
| R1 | 47R |
| R2 | 2k2, 2½ W |
| R3, R9 | 47k |
| R4, R5 | 100k |
| R6, 7, 10 | 470k |
| R8 | 220k |
| R11 | 2M2 |
| R12 | 10M |
| RV1 | 20k min. trimpot |

Capacitors

| | |
|--------|------------------------|
| C1 | 1000u/25 V RB electro. |
| C2, C3 | 220n greencap |

Semiconductors

| | |
|-------|-----------------|
| D1-D5 | 1N4004, EM404 |
| D6-D9 | 1N914, 1N4148 |
| IC1 | 4050 |
| Q1 | TIP31C |
| ZD1 | 12 V, 5 W zener |

Miscellaneous

ETI-494 pc board; RL1 — Fujitsu FRL611D012, 12 volt SPDT 10 A contacts or similar relay (pc mount type).

Price Estimate
\$20-\$24

supplying approximately 784 watts into an 8 ohm load or 1568 watts into a 4 ohm load. If the amplifier to be used is capable of powers greater than these the regulator transistor should be substituted for a device with a higher V_{ce0} rating. The relay pulls around 40 mA when operated, so power dissipation in the regulator transistor will be around 4 watts when dropping 100 volts. Although this is not a particularly high

dissipation it is high enough to lie outside the SOAR rating of many high voltage transistors, so be careful when choosing an alternate regulator transistor.

Construction

Construction is straightforward since all of the components are mounted on the pc board. The usual precautions should be taken to ensure that all polarised components have been mounted with the correct orientation. The IC used is a CMOS type and is therefore static sensitive. Solder this last and preferably using an earthed soldering iron. It is a wise precaution to discharge yourself before handling the device by first touching an earthed metal appliance. For a more detailed description of precautions when handling CMOS refer to our article 'Electrostatic discharge — nemesis of electronic systems' in the June edition, 1981.

It is a wise precaution to space the 2.5 W resistor, R2, off the pc board slightly. In the case of a high powered loudspeaker going faulty with dc this component will get quite hot and spacing improves ventilation around the component and prevents the possibility of charring the pc board. If you can't obtain a 2.5 watt type (e.g. Philips PR52), then a 5 W type may be substituted.

Before mounting the unit check operation by connecting around 20 V dc across the speaker input terminals on the pc board. The relay should cut in after about one tenth of a second. If the

protector passes this test connect the speaker wiring. If the preset is turned fully down (turn it anticlockwise when viewing the board with the components on top and the relay to the right) the relay will cut in when the power exceeds around 20 watts for an extended period. The protector allows transients to the full supply rail to pass but will prevent a continuous 20 W from being applied to the loudspeaker. To increase this, turn the preset clockwise until the desired response is achieved.

Performance

We tested the loudspeaker protector for its effect on audio performance as well as its reliability. A variety of power amps were used to ensure that the load represented by the protector would not affect audio performance. Even a very low power amplifier, with a comparatively small damping factor (high output impedance) could drive the unit with no degradation to the sound quality. During every test the protector worked well and cut in at the correct time to prevent damage to the loudspeakers.

NOTE. Some amplifiers are unstable when driven into an open circuit. This is particularly true of valve power amplifiers some of which destroy themselves the moment the speaker is disconnected. Loudspeaker protectors are however, not usually required for use with valve power amps since the possibility of dc on the speaker lines is remote, but over-power protection may be required.

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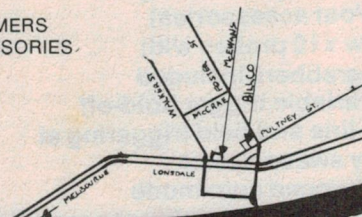
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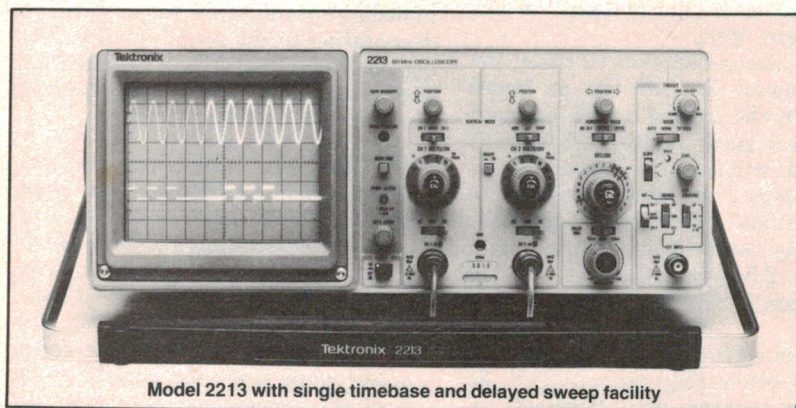
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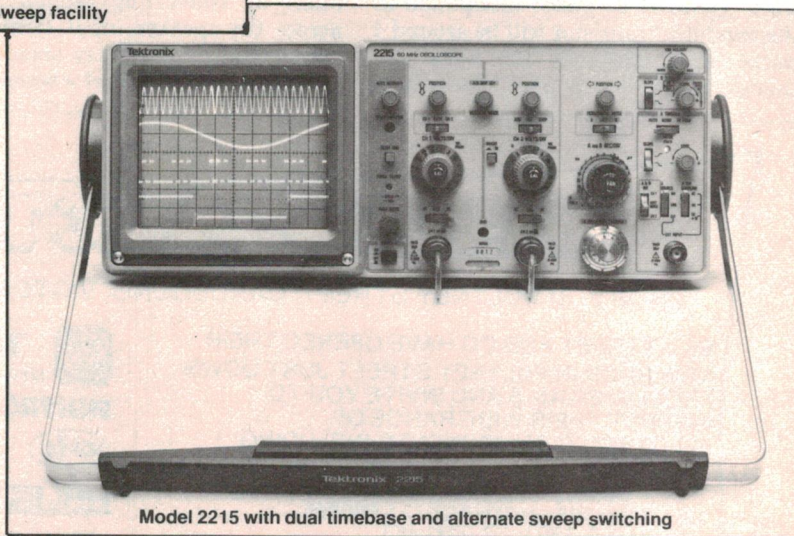


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FEATURES See review, pages 15 to 18, July issue.

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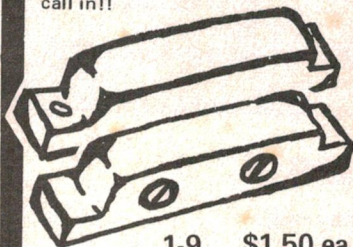
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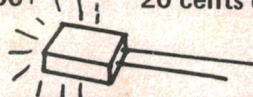


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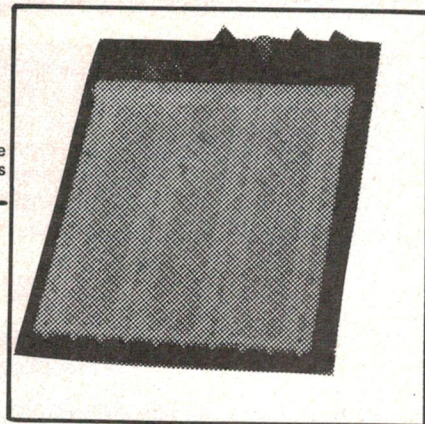
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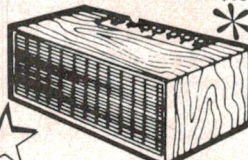
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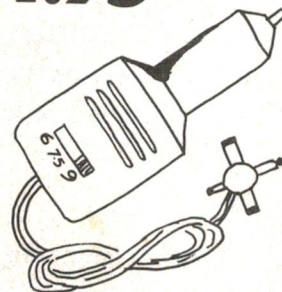
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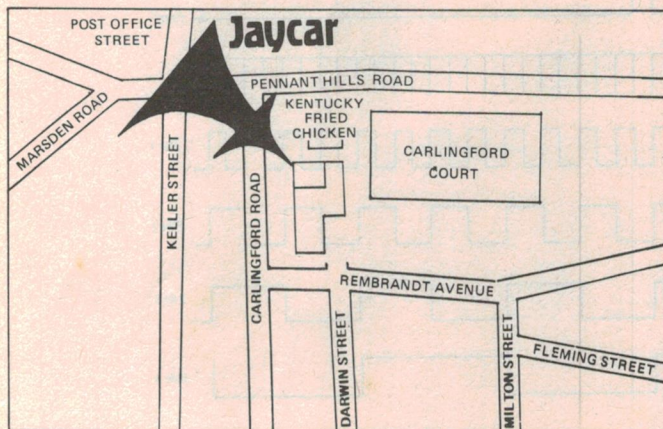
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An ordinary NICKEL PLATED lead set like this costs around \$4.75. The gold versions normally are around \$9.95. You can have one of these for \$3.95 and that's better than Nickel any day! Worth it for the 4 x RCA's alone!

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JAYCAR

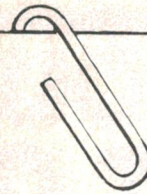
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Lab Notes



Super timer — from μs to days

Timing long periods has always been a problem because of the high leakage characteristics of the timing capacitor. This is no longer true! The XR-2240 IC is a programmable timer capable of producing ultra-long time delays without sacrificing accuracy.

Barry Davis

THIS IC can generate time delays from microseconds up to five days, and with a little ingenuity can generate a delay of a couple of years! A functional block diagram of the IC is shown in Figure 1.

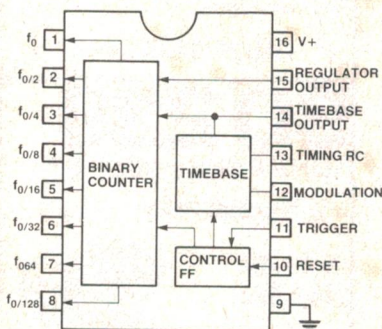


Figure 1. Functional block diagram of the XR-2240.

The circuit consists of an internal timebase generator, a programmable 8-bit counter and a control flip-flop. The time delay at the output is set by an external CR network and can be any period from $1 \cdot \text{CR}$ to $255 \cdot \text{CR}$. Herein lies the secret. The CR timebase generator can be set to give a very accurate short period, and binary multiples of this short period are then programmed and taken from the output. Each output is capable of sinking approximately 5 mA of load current.

The features of the IC are:

1. Timing from microseconds to days.

2. Programmable delays: $1 \cdot \text{CR}$ to $255 \cdot \text{CR}$
3. Wide supply voltage range: 4 V to 15 V
4. TTL compatible inputs and outputs
5. High accuracy: 0.5%
6. Excellent temperature stability
7. Period $T = C \times R$

Circuit operation

The timing cycle is initiated by a positive-going pulse on pin 11. This trigger pulse performs three functions:

1. Activates the timebase generator
2. Enables the counter
3. Sets all counter outputs to the low state

The timebase generator produces timing pulses with a period, T , equal to $1 \cdot \text{CR}$. These clock pulses are counted by the binary counter inside the IC and the timing period is complete when a positive-going pulse is applied to pin 10 (i.e. the circuit is reset). In most applications one or more of the output terminals are connected back to the reset input. The circuit will commence timing when the trigger pulse is applied, and automatically reset on the completion of the timing period.

Remember, the outputs are normally high and are set to low when timing is initiated, returning again to the high level on completion of the timing period.

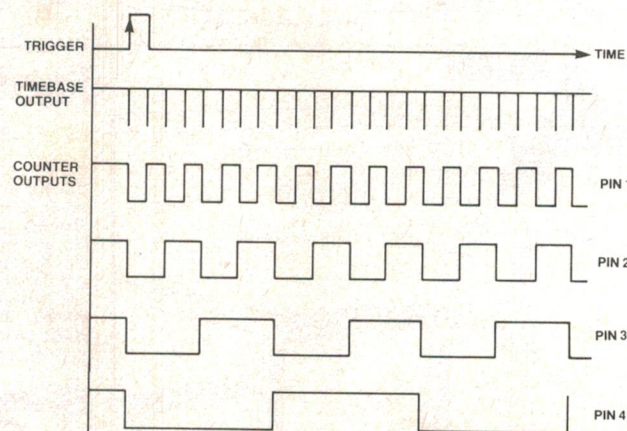


Figure 2. Output waveforms and timing diagram.

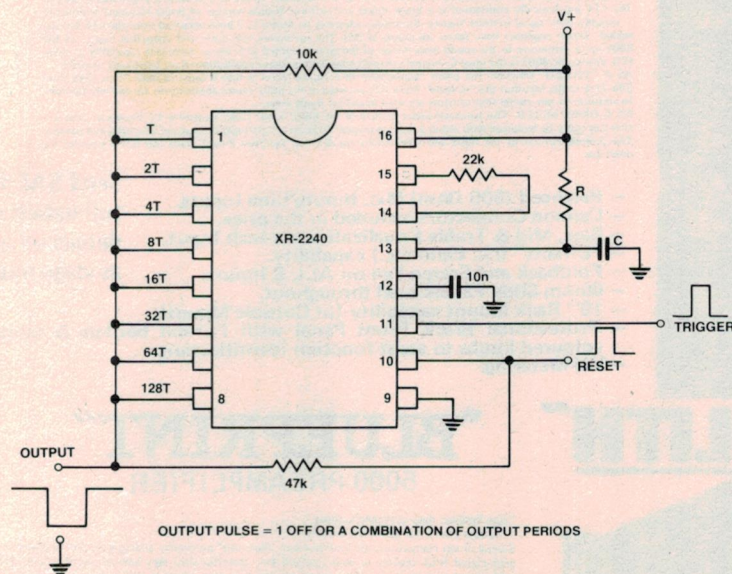


Figure 3. Connections for a practical circuit.

Circuit construction

The binary outputs, pins 1 to 8, are open collector and can be connected together to a common pull-up resistor. The output of the timer will be low as long as any one output is low. In this manner the time delays associated with each output can be added by simply connecting them together to a common bus. The outputs can be used individually or wired together.

For example, the output at pin 4 is $8 \times CR = 8T$. If pins 4 and 3 are connected together the output will become $12 \times CR = 12T$.

Figure 3 shows the actual connections for a practical circuit. When the power is applied, with no trigger or reset inputs, the circuit sets up to the initial state of all outputs high. Once triggered, the circuit is totally immune to any additional trigger inputs until the timing period is completed, or a reset pulse is applied.

Choice of timing components

Once a signal timing period, T , is established, the output can be determined by 'wiring-in' periods of T following a binary progression. However, the procedure may have to be reversed when a certain accurate output period is required.

For example, if a timing period of 6 hours 30 seconds is required, firstly convert the time to seconds:

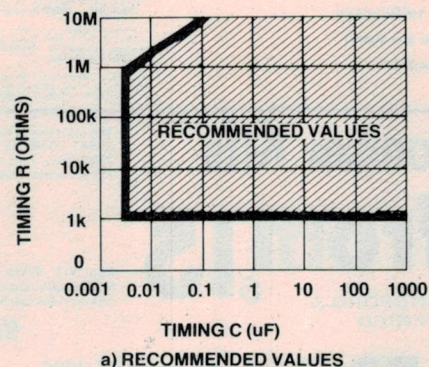
$$= 6 \times 60 \times 60 + 30 \\ = 21630 \text{ seconds.}$$

The maximum number of timing periods available with one IC is $(1+2+4+8+16+32+64+128)/T = 255T$. Therefore the period of T can be calculated:

$$T = \frac{21630}{255} \\ = 84.82 \text{ seconds}$$

With a low-loss capacitor (such as tantalum) as one timing component, R can be calculated. If $C = 100 \mu\text{F}$:

$$T = CR$$



Therefore:

$$R = \frac{T}{C} \\ = \frac{84.82}{100 \times 10^{-6}} \\ = 848.24 \text{ k}\Omega$$

This can be tailored precisely for very accurate timing with a resistive network or potentiometer, or simply rounded off to 850k.

Figure 4 shows two graphs which will assist you in choosing:

1. The recommended range of timing component values.
2. The time period, (T) up to 100 seconds, to be expected from combinations of C and R values.

An example of output periods to be expected using a $100 \mu\text{F}$ capacitor (tantalum) and 1M resistor as the timing components is shown in Table 1.

| $T = CR = 100 \mu\text{F} \times 1 \text{M} = 100 \text{ seconds}$ | |
|--|------------------------|
| T | Period of Output |
| 100 | secs = 1.7 min |
| 2T | 200 secs = 3.3 min |
| 4T | 400 secs = 6.7 min |
| 8T | 800 secs = 13.3 min |
| 16T | 1600 secs = 26.7 min |
| 32T | 3200 secs = 53.3 min |
| 64T | 6400 secs = 1.8 hours |
| 128T | 12800 secs = 3.6 hours |
| 255T | 25500 secs = 7.1 hours |

Table 1. Example of accurate time available using the XR-2240.

The type of circuit operation discussed to this point has been monostable i.e: the output goes low on triggering, stays low for the timing period and returns to a high level. It will not time again until it is retriggered. An XR-2240 can also be used in a free-running or astable mode.

Astable operation

To operate in this mode the reset line to pin 10 is disconnected from the output. ▶

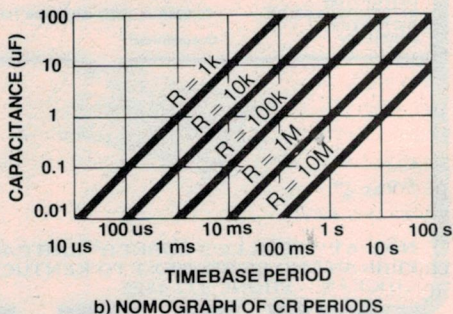
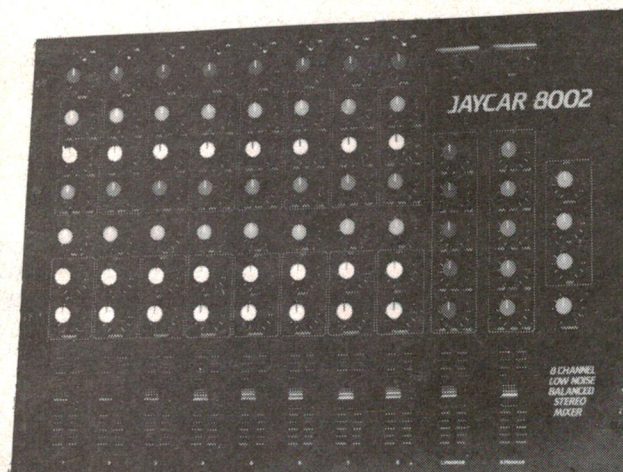


Figure 4. Graphs to assist in the choice of values of C and R .

JAYCAR QUALITY MEANS PERFORMANCE RESULTS

FROM \$495



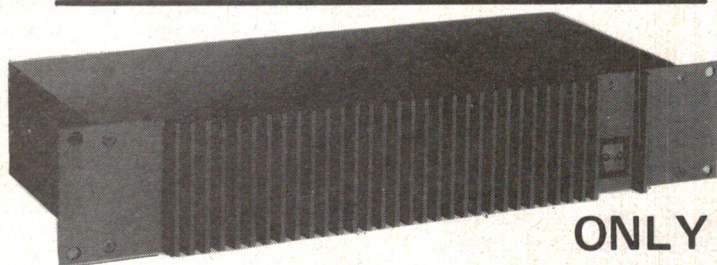
8 CHANNEL MIXER KIT

The Jaycar 8002 Mixer was originally conceived to be the successor to the very popular ET1414 Master Mixer. The 414 was basically configured as a 'stage' mixer and suffered from a number of severe technical limitations - notably poor signal-to-noise figures. Enormous advances in Audio IC's have occurred since the 414 was designed. Jaycar engineers have taken advantage of this. The incredibly low noise and distortion figures of the 8002 are a testimony to the sound basic design of the mixer coupled with the performance capability of these IC's. Whilst the 8002 is the ideal 8 channel compact stage mixer, other applications have been kept in mind. AS A "STUDIO" MIXER. The prime requirement of a studio mixer is that it must be quiet - i.e. have good S/N. Due to the fact that the "miracle" 5534 IC's are used in the 8002 studio applications are entirely feasible. In addition to this, metal film resistors are used in critical signal areas. AS A DISCO MIXER. The balanced input feature of the 8002 is not really necessary for disco use. This section can easily be bypassed with either a moving magnet (Dynamic Cartridge) preamp, or a moving coil preamp. The sensible format of the 8002 and tremendous equalization facilities should make this mixer popular for disco use.

- Balanced (600 Ohm) Mic. Inputs/Line Inputs.
- Cannon Connectors included in the price.
- Bass, Mid & Treble Equalization on each Input.
- "Effects" (i.e. Echo etc.) capability.
- Foldback and Stereo Pan on ALL 8 Inputs.
- 60mm Slide Faders used throughout.
- 19" Rack Mount capability (or Console Mount).
- Professional Black Front Panel with Format borders & multi-coloured knobs to assist function identification.
- VU Metering.

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Write in (SAE) for a new glossy leaflet on both amps.

SPECIFICATIONS

POWER OUTPUT FREQUENCY RESPONSE

Around 100W RMS into 8 ohms
8Hz to 20kHz, +0 - 0.4dB

2.8Hz to 65kHz, +0 - 3dB
Note: these figures are determined solely by passive filters

INPUT SENSITIVITY

HUM NOISE

2nd HARMONIC DISTORTION

3rd HARMONIC DISTORTION

TOTAL HARMONIC DISTORTION

INTERMODULATION DISTORTION

STABILITY

1V RMS for 100W output
- 100dB below full output (flat)
- 116dB below full output (flat, 20kHz bandwidth)
<0.001% at 1kHz (0.0007% on prototypes) at 100W output using a *56V supply rated at 4A continuous
<0.003% at 10kHz and 100W
<0.0003% for all frequencies less than 10kHz and all powers below clipping
Determined by 2nd harmonic distortion (see above)

<0.003% at 100W (50Hz and 7kHz mixed 4:1)

Unconditional

"BLUEPRINT"

5000 PREAMPLIFIER

"One Swallow does not make a spring"

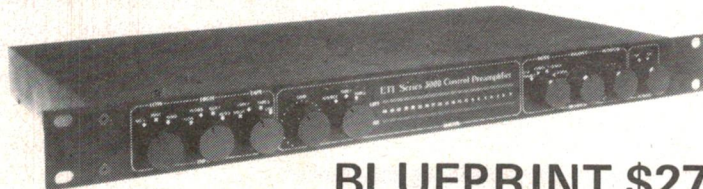
- Neither does a few gold RCA sockets!

Several of our competitors are imitating our "Blueprint" preamp by adding a few bits and pieces, notably gold plated RCA sockets to their standard kits. Unfortunately they have missed the point. We supply gold plated sockets in our "Blueprint" preamp but only where it makes sense to do this, i.e. on the inputs - NOT the outputs. 16 gold sockets are provided by us. This, however, does not make a "Blueprint". THIS DOES:

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- Factory pre-tinned PCB's to reduce chances of dry or noisy solder joints.
- Quality LEDs, polished finish, multicoloured display.
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- Special rear panel.
- Special low noise selection LM394H NOT CH device in M.C. preamp.
- Thermalloy (U.S. made) heatsink on 7805 regulator.
- English Lorlin selector switches.
- Apart from the 16 gold RCA's we throw in a pair of gold plated line RCA plugs - worth \$5.
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BLUEPRINT \$275

SPECIFICATIONS

Frequency Response

High level input: 15Hz-130kHz, +0, -1dB
Low level input: conforms to RIAA equalisation
-0.2dB (see detail on Phono spec.)
1kHz: 0.003% on all inputs (limit of resolution on measuring equipment due to noise limitation)
High level input, master full, with respect to 300mV input signal at full output (1.2V):
92dB flat; 100dB A-weighted

Distortion

MM input, master full, with respect to full output (1.2V) at 5mV input, 500 ohm source resistance connected: 86dB flat
92dB A-weighted

S/N noise

MC input, master full, with respect to full output (1.2V) and 200mV input signal:
71dB flat 75dB A-weighted

ETI-478MM Moving Magnet input stage

Gain

0.001%, 1kHz, 10mV RMS input

Frequency Response

20dB with respect to 5mV RMS input signal, i.e. 135mV RMS

Total Harmonic Distortion

122mV 'A' input shorted, 216mV flat, input shorted

Noise

1mV 5mV 10mV
73dB 87dB 93dB
A-weighted 78dB 92dB 95dB

S/N ratio

24

ETI-478MC Moving coil input stage

Gain

0.003%, 1kHz, 30mV input

Frequency Response

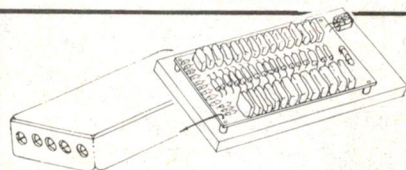
Total equivalent input noise: 83mV flat, input shorted

42mV 'A' input shorted

56mV flat, after RIAA Eq. input shorted

34mV 'A', after RIAA Eq. input shorted

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Lab Notes

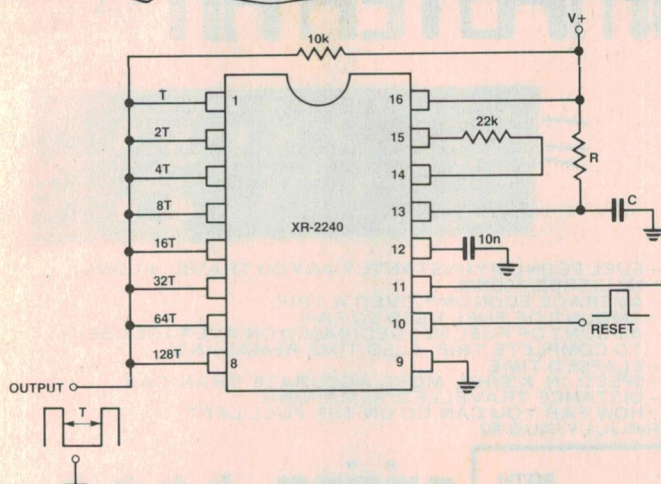


Figure 5. Astable operation under control of external trigger and reset controls.

Figure 5 shows an astable circuit under the control of the external trigger and reset signals. It will start timing when an external trigger pulse is applied, and will not stop until a reset pulse is applied.

Alternatively, the circuit can be made truly free running. The circuit in Figure 6 self-triggers automatically when the power is switched on, and continues to operate in its free running mode indefinitely.

When the timer is used in this mode, each counter output can be used individually as synchronised oscillators, or they can be connected together to provide complex pulse patterns.

Ultra-long delays

In some applications delays of four days may be required. This is particularly useful in electronic farming for controlling the rate of supplementary feeding. The timing components required can be calculated thus:

$$\begin{aligned} 4 \text{ days} &= 96 \text{ hours} \\ &= 5760 \text{ minutes} \\ &= 345600 \text{ secs.} \end{aligned}$$

Maximum number of T combinations = 255

Therefore

$$\begin{aligned} T &= \frac{345600}{255} \\ &= 1355.3 \text{ secs} \\ &= 22.6 \text{ minutes} \end{aligned}$$

Incidentally, 20 minutes is about the longest time recommended for 1.C.R as anything beyond this suffers from leakage problems.

$$T = CR$$

if $C = 500 \mu\text{F}$ (low leakage)

$$R = \frac{T}{C}$$

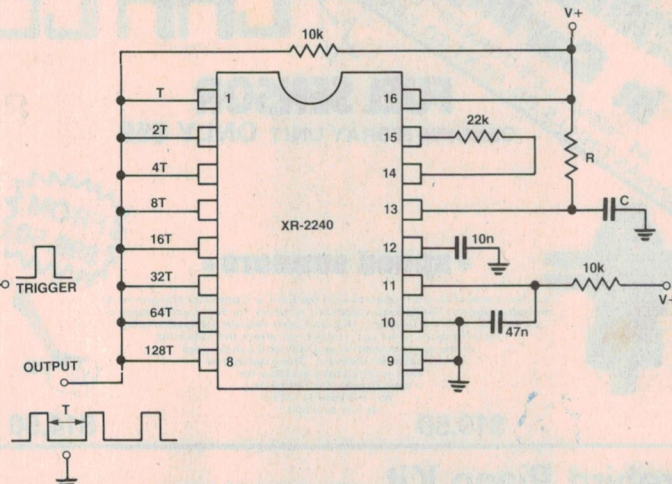


Figure 6. Free-running circuit.

$$\begin{aligned} &= \frac{1355.3}{500 \times 10^{-6}} \\ &= 2M7 \end{aligned}$$

Two XR-2240 ICs can be cascaded to generate extremely long time delays. When used in this format the reset and trigger terminals of the ICs are tied together and the timebase of unit 2 disabled as shown in Figure 7.

The output is normally high. When a positive-going trigger pulse is applied the output goes low and stays in the low state for $(256)^2 = 65536$ periods of the timebase oscillator. Therefore the total timing period of two cascaded units can be from $256.CR$ to $65536.CR$. The output is available in 256 discrete steps by selectively connecting one or a combination of the outputs from unit 2 to the output bus.

With $T = 20$ minutes an example of an ultra-long delay can be given.

$$CR = T = 20 \text{ mins.}$$

$$\begin{aligned} 65536T &= 1310720 \text{ min} \\ &= 21845 \text{ hours} \\ &= 910 \text{ days} \\ &= 2.5 \text{ years!} \end{aligned}$$

This article highlights the use of an XR-2240 as a precision timer. Other application suggestions are:

1. Sequential timing
2. Binary pattern generation
3. Frequency synthesis
4. Pulse counting or summing
5. A/D conversion
6. Digital sample and hold

Further information on the IC can be obtained from Exar Integrated Systems or their agents (Total Electronics).

This article was made possible by the courtesy of Exar Integrated Systems. Data was taken from their publication 'XR-2240/2340 Programmable Timer Counter'.

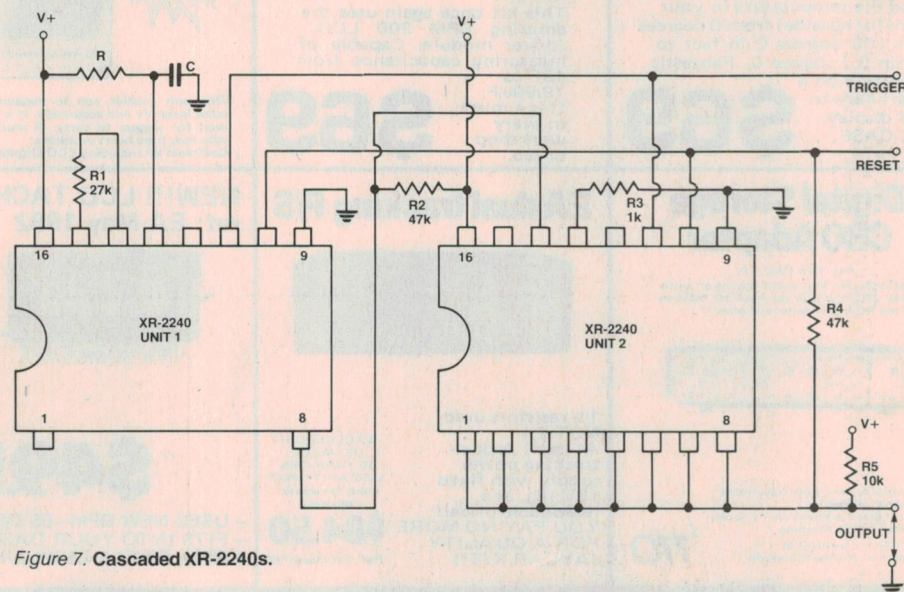


Figure 7. Cascaded XR-2240s.

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HEART RATE MONITOR

Ref: EA 7/82

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Ref: ETI SEPTEMBER 82
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EA dual tracking P/S



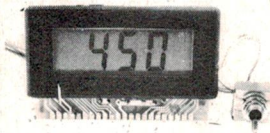
1% resistors used
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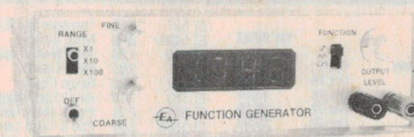
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Ref: EA April 1982

Function Generator



Unit pictured with EA Panel

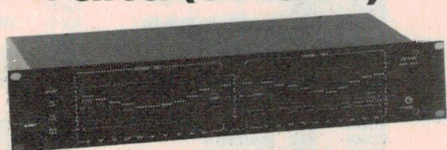
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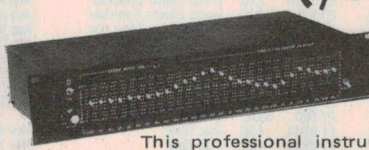
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This professional instrument is used for total control of acoustic problems. Many hundreds of these are in use by live P.A. operators, Disco installations, recording studios and even the audiophile who demands the ultimate in control.

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syntom

As used by Warren Cann of 'Ultravox'

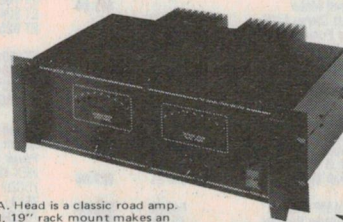


Only \$36.50

Original design from the UK magazine "Electronics and Music Maker" April 1981. Self-contained unit produces a variety of fixed and falling pitch effects. Trigger by tapping the unit itself or by striking a drum to which the unit is attached. The Jaycar "SYNTOM" comes complete with high quality pre-drilled moulded all ABS box 152 x 80 x 47mm with professional silk-screened front panel.

FEATURES: Decay from less than 0.1 second to several seconds, pitch control, sweep control and volume on/off.

GOOD-BYE 3002

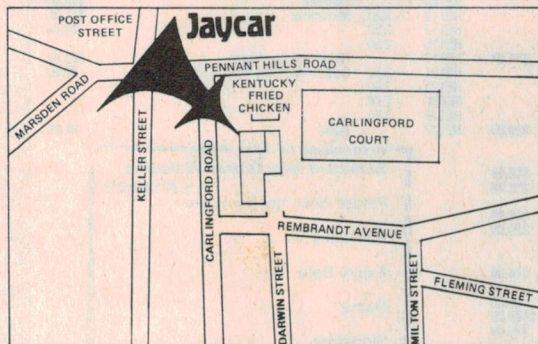


This 2x300W RMS P.A. Head is a classic road amp. Ruggedly constructed, 19" rack mount makes an ideal main P.A. or foldback unit. Great for Disco use as well.

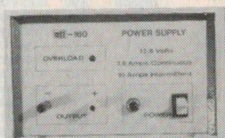
We are discontinuing this amp because it is becoming too expensive to make. The metalwork costs alone now account for well over 50% of the unit. Because of this we have reluctantly decided to discontinue the unit. You can grab one now while they last for only \$399

\$489 \$399

Send SAE for full spec. sheet.



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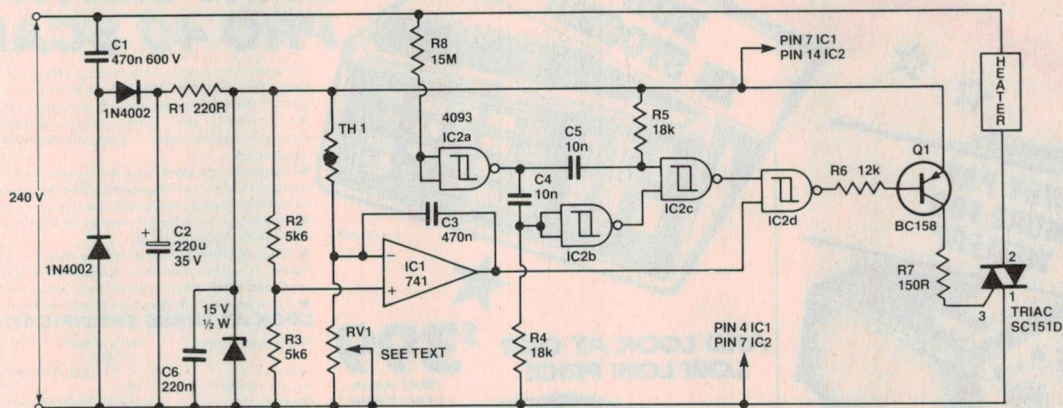
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Ideas for Experimenters

These pages are intended primarily as a source of ideas. As far as reasonably possible all material has been checked for feasibility, component availability etc, but the circuits have not necessarily been built and tested in our laboratory. Because of the nature of the information in this section we cannot enter into any correspondence about any of the circuits, nor can we produce constructional details.



Electronic thermostat

This circuit, designed by **Steve Gagen of North Balwyn Victoria**, has been used for several months to control the temperature in an incubation room. According to Steve it has performed well, achieving drift-free temperature regulation with an accuracy of $\pm 0.5^\circ\text{C}$.

The low voltage supply to the ICs is taken directly from the mains, via capacitor C1. The thermistor (T) should be of the bead type and, if necessary, may

be sited at some distance from the rest of the circuit. RV1 should be chosen so that its mid-range resistance is approximately equal to the resistance of the thermistor at the desired temperature.

The difference signal between the thermistor in the RV1 network and the voltage divider R2-R3 is amplified by IC1 and used to gate the output of the pulse generating circuit formed by IC2. Capacitor C3 prevents the amplification of any extraneous ac.

When the output from IC1 is high, 90 mA pulses of about 200 μs length are applied to terminal 3 of the triac at the

beginning of each mains half cycle, turning it on.

The circuit tends to cycle on and off every minute or so and the triac avoids the problem of burnt contacts which a relay would experience in these circumstances. Since the heat control is non-proportionating, the circuit is suitable for use with fan heaters.

Care should be used as the entire circuit is at mains potential, and the triac should be mounted on its heatsink using a mica insulator. A heatsink is essential as when controlling a 2.4 kW heater the triac dissipates about 15 W.

1 kW dummy load for testing audio power amplifiers

Philip Allison of Summer Hill NSW has worked out a cheap, simple method for testing audio power amplifiers.

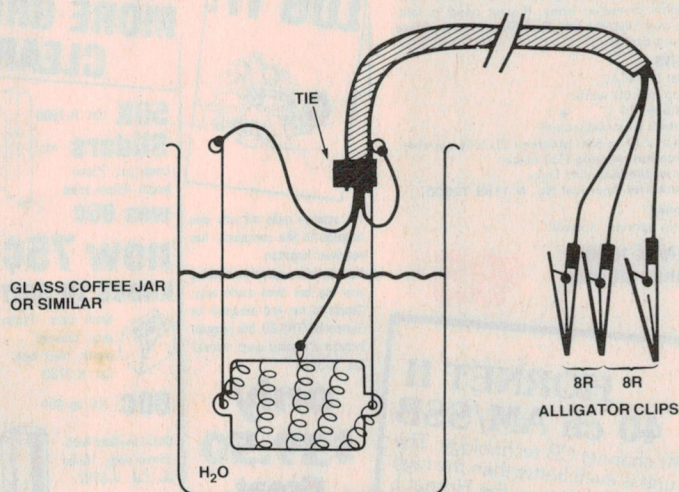
You'll need a 1.6 kW electric jug replacement element, complete with its supporting arms, which you can buy from a hardware store.

The coil of Ni-Chrome wire on these elements has a resistance of about 36 ohms which needs to be reduced to 16 ohms for our purpose. To do this, first remove the coil and cut it at 16 ohms. Then stretch this length so that it equals the original length and carefully rewind it on the ceramic former.

Find the centre of this coil (8 ohms) and make a small twist. Using a length of three core mains flex attach the green wire to the twist and the blue and brown

wires to the ends of the brass rods as shown in the diagram. Fit alligator clips or plugs to the other ends of this lead.

When immersed in water this unit



will comfortably dissipate 500 watts per 8 ohm side or 1000 watts with a 4 ohm load (blue and brown linked) or with a 16 ohm load (using blue and brown only).

REAP THE BENEFIT

DICK'S



The NEW Dick Smith Temperature Controlled Soldering Station

At last — A professional quality temperature controlled iron at a budget price! Yet it has even more features than those irons you'd pay \$\$\$ more for! Features like... an exclusive temperature meter to show you exactly what's happening... fully controllable temperature from around 200°C to 500°C... self contained iron stand, tip cleaning sponge and a bit well. And more... it's fully approved by the Electricity Authority! Easy to use, lightweight — so that operators don't suffer from the fatigue problems of earlier irons. If your need is fast, efficient soldering, you need the new Dick Smith Temperature Controlled Soldering Station.

Specifications

Operating voltage: 240V AC
Iron Wattage: Nominal 50 watts
Iron Element Voltage: 24 V
Temperature Range: 200 to 500 (degrees C), fully variable
Temperature Indication: Moving Coil Meter
Tips Available: Fine, Medium and Thick
Electricity Authority Approval No. N 1188 T2000

Accessories supplied

Iron stand, cleaning sponge, tip well

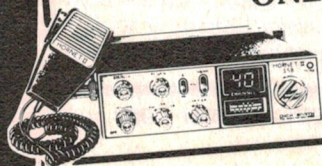
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Cat D-1710
DOC APP
NO. 249A006



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Don't be fooled — the ONLY fuel flow sensor to give optimum results with the car computer is the fully imported Moray unit. We are sole authorised Australian agents for this device. Buy the best and get the incredible accuracy you deserve!

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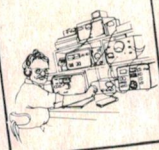
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SPECIAL OFFER

Buy both sensors and get the speed sensor for \$3.00! YES! The exclusive Moray Fuel Flow Sensor plus the Driveline Sensor at the bargain price of just \$72.50! This makes the complete Dick Smith Car Computer only \$166.00 — with nothing else to buy! GREAT VALUE!!!

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★★★★★

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a professional sized computer you build yourself!
COMPLETE WITH INBUILT POWER SUPPLY



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DICK SMITH

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Not just simple sticks: full space-age guiding devices with 48 keys as well. You get incredible control capability — far more than possible with other games. And slide-in overlays for controllers: you can see exactly what to do! When the controllers are placed in the console, they form a standard typewriter-style 'QWERTY' keyboard — that's versatility!

Fantastic game Cartridges

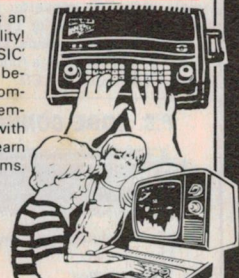
21st century games — like you've never played before. And every cartridge has many games — up to 32 in each! You'll never be satisfied with anything less after playing the WIZZARD. Every cartridge has a built-in demonstration program to show you how to play the game.



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'BASIC' CARTRIDGE
Available soon — watch our ads for details!



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Cat. Y-1600

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(Midscale 200 Ω)

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100K

Spectrol model 534 1/4" shaft
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Ideal for use with flat
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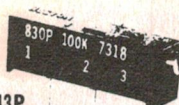
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20 TURN CERMET TRIM POT



SPECTROL 43P ACTUAL SIZE

STOCK RESISTANCE VALUES
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2K, 5K, 10K, 20K, 50K, 100K, 200K,
500K, 1M, 2M
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10-99 \$1.30
100 \$1.20
Values may be mixed.

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19-key pad in-
cludes 1-10 keys
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Ideal for dream
project

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| DE-9C | 9 PIN COVER | 2.20 | 2.10 | 1.90 |
| DA-15P | 15 PIN MALE | 4.50 | 4.20 | 3.90 |
| DA-15S | 15 PIN F/MALE | 5.10 | 4.90 | 4.70 |
| DA-15C | 15 PIN COVER | 2.30 | 2.10 | 2.00 |
| DB-25P | 25 PIN MALE | 5.90 | 5.60 | 5.10 |
| DB-25S | 25 PIN F/MALE | 6.90 | 6.60 | 6.10 |
| DB-25C | 1 pc. Grey Hood | 2.40 | 2.20 | 2.00 |
| DB-25C2B | 2 pc. Black Hood | 2.80 | 2.70 | 2.50 |
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| DC-37S | 37 PIN F/MALE | 10.90 | 9.90 | 9.10 |
| DC-37C | 37 PIN COVER | 4.90 | 4.50 | 4.10 |
| DH-S | Hardware set (2 Pairs) | 2.10 | 1.90 | 1.80 |



cermet single TURN TRIM POT

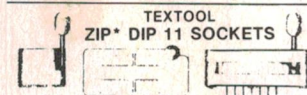
Spectrol model 63P
ACTUAL SIZE

STOCK VALUES
10R, 20R, 50R, 100R, 200R, 500R, 1K,
2K, 5K, 10K, 20K, 50K, 100K, 200K,
500K, 1M, 2M
1-9 \$1.00
10-99 0.90
100 0.80
Values may be mixed.

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| SD7 | 7 | 2.40 |
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| IN5408 | 35c | 30c |
| IN4007 | 12c | 11c |

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| 3.00 | 2.90 | 2.50 | 2.00 | 2.00 | 1.50 | |
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| 5.80 | 5.40 | 4.90 | 3.80 | 2.90 | 2.70 | |
| HS4 - 225mm | | | | | | |
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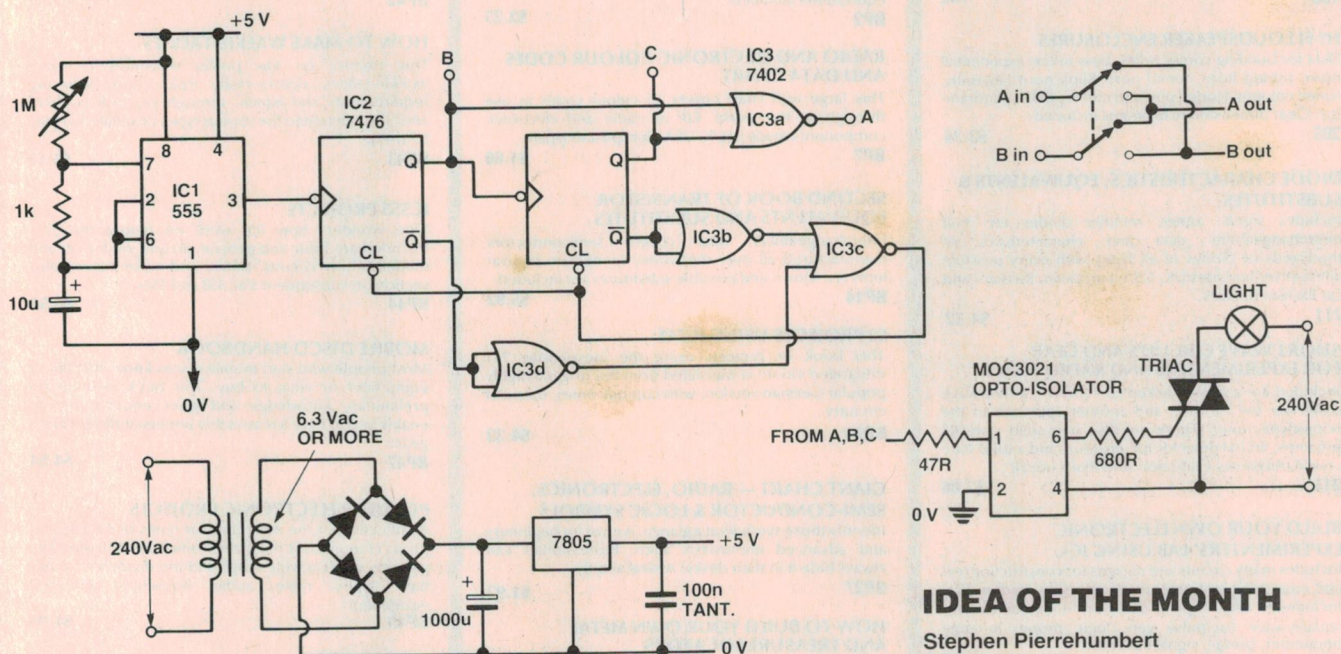
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Ideas for Experimenters



Three-channel light chaser

This simple idea comprises a three-channel light chaser incorporating a 'reversing' switch so that you can change the direction of the 'chase'.

A 555 is arranged as a variable astable multivibrator (IC1). Its output drives one flip-flop from a dual JK flip-flop (IC2). The Q output of this flip-flop drives the second flip-flop and a group of NOR gates (IC3) such that three outputs are produced, going high successively. The three outputs then drive opto-

isolators which trigger triacs which drive the lamps.

A simple power supply circuit provides supply to IC1, IC2 and IC3. The DPDT switch reverses the A and B drives to reverse the chase sequence. The 1M variable pot varies the speed of the chase.

'IDEA OF THE MONTH' CONTEST

Scope Laboratories, who manufacture and distribute soldering irons and accessory tools, have offered to sponsor a contest with a prize to be given away every month for the best item submitted for publication in the 'Ideas for Experimenters' column — one of the most consistently popular features in ETI. Each month we will be giving away a Scope Panavise pc board holder, model 333 — as described in News Digest, p.8, October '81 issue. Selections will be made at the sole discretion of the editorial staff of ETI Magazine. Apart from the prize, worth about \$70, each winner will be paid \$10 for the item published. You must submit original ideas of circuits which have not previously been published. You may send as many entries as you wish.

RULES

This contest is open to all persons normally resident in Australia with the exception of members of the staff of Scope Laboratories, Murray Publishing, Offset Alpine, Australian Consolidated Press and/or associated companies.

Closing date for each issue is the last day of the month. Entries received within seven days of that date will be accepted if postmarked prior to and including the date of the last day of the month.

The winning entry will be judged by the Editor of ETI, whose decision will be final. No correspondence can be entered into regarding the decision.



Winner will be advised by telegram the same day the result is declared. The name of the winner, together with the winning idea, will be published in the next possible issue of ETI.

Contestants must enter their names and address where indicated on each entry form. Photostats or clearly written copies will be accepted but if sending copies you must cut out and include with each entry the month and page number from the bottom of the page of the contest. In other words you can send in multiple entries but you will need extra copies of the magazine so that you send an original page number with each entry.

This contest is invalid in states where local laws prohibit entries.

Entrants must sign the declaration on the coupon that they have read the above rules and agree to abide by their conditions.

COUPON

"I agree to the above terms and grant Electronics Today International all rights to publish my idea in ETI Magazine or other publications produced by them. I declare that the attached idea is my own original material, that it has not previously been published and that its publication does not violate any other copyright".

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Title of idea
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Address
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Included are a crystal locked HF receiver, a shortwave converter for 2-6 MHz and another approach to the homodyne tuner. There are also two solid state RF preamps, an FM detector for amateurs and a solid state crystal frequency calibrator. And that's not all

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Introduction to digital ICs (mainly TTL 7400). Besides simple projects, includes logic test set to identify and test digital ICs. Also includes digital counter-timer.

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Full practical constructional details of receivers with performance equal to commercial units. Also 'add-on' circuits of Q meter, S meter, noise limiter etc.

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Shows equivalents and pin connections of a selection of popular linear ICs, with details of families, functions, country of origin and manufacture. Includes devices from Analog Devices, Advance Micro Devices, Fairchild, Harris, ITT, Motorola, Philips, RCA, Raytheon, Signetics, Sescocem, SGS-ATES, Siemens, AEG-Telefunken, Teledyne, Texas Instruments.

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One wonders how life went on before the 555! Included are basic and general circuits, motor car and model railway circuits, alarms and noise makers plus section on subsequent 556, 558 and 559s.

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Most people who start mobile discos know little about equipment or what to buy. This book assumes no preliminary knowledge and gives enough info to enable you to have a reasonable understanding of disco gear.

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A collection of the most popular types of circuits and projects to interest most electronics constructors. The projects cover a wide range and are divided into four basic types: radio, audio, household and test equipment.

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LM 3900 IC PROJECTS

Unlike conventional op-amps, the LM 3900 can be used for all the usual applications as well as many new ones. It's one of the most versatile, freely obtainable and inexpensive devices around. This book provides the groundwork for simple and advanced uses — it's much more than a collection of projects. Very thoroughly recommended.

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LONG DISTANCE TV RECEPTION (TV-DX)

Written by UK authority, the book includes many units and devices made by active enthusiasts. A practical and authoritative intro to this unusual aspect of electronics.

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YOUR ELECTRONIC CALCULATOR AND YOUR MONEY

Starts with a basic revision of percentages and decimals, then deals with mortgages, cars, insurance, fuel, shopping, tax etc. There's a section on investment and the last section deals with the calculator in a small business.

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RADIO STATIONS GUIDE

This is an aid for all those who have a radio receiver. Shows the station site, country, frequency and/or wavelength, as well as the Effective Radiation Power (ERP) of the transmitter and in some cases, the station's call sign as well.

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Besides including both simple and more sophisticated burglar alarm circuits using light, infrared and ultrasonics, this book also gives circuits for gas and smoke detectors, flood alarms, fire alarms, doorphones, baby alarms, etc.

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BEGINNER'S GUIDE TO DIGITAL ELECTRONICS

Covers all essential areas including number systems, codes, constructional and sequential logic, analog/digital/analog conversion.

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ELEMENTS OF ELECTRONICS

This series provides an inexpensive intro to modern electronics. Although written for readers with no more than basic arithmetic skills, maths is not avoided — all the maths is taught as the reader progresses.

The course concentrates on the understanding of concepts central to electronics, rather than continually digressing over the whole field. Once the fundamentals are learned the workings of most other things are soon revealed. The author anticipates where difficulties lie and guides the reader through them.

BOOK 1 (BP62): All fundamental theory necessary to full understanding of simple electronic circuits and components.

BOOK 2 (BP63): Alternating current theory.

BOOK 3 (BP64): Semiconductor technology leading to transistors and ICs.

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Used properly, this chart should enable the reader to trace most common faults quickly. Across the top of the chart are four rectangles containing brief descriptions of these faults: sound weak but undistorted; set dead; sound low and distorted; background noises. Selecting the appropriate fault, the reader simply follows the arrows and carries out the suggested checks in sequence until the fault is cleared.

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REMOTE CONTROL PROJECTS

Covers radio, infra-red, visible light, ultrasonic controls. Full explanations are provided so that the reader can adapt the projects for domestic and industrial as well as model use.

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Provides constructors with practical circuits for the less complex music equipment including fuzz box, waa-waa pedal, sustain unit, reverb and phaser, tremolo generator etc. Text covers guitar effects, general effects, sound generators, accessories.

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This book gives a number of power supply designs, including simple unregulated types, fixed voltage regulated types and variable voltage stabilised designs. The designs are all low voltage types for semiconductor circuits.

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How to build typical computer circuits using discrete logic. This book is useful intro to devices such as adders and stores as well as a general source book of logic circuits.

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RADIO CONTROL FOR BEGINNERS

How complete systems work with constructional details of solid state transmitters and receivers. Also included — antennas, field strength meter, crystal controlled superhet, electro-mechanical controls. Ideal for beginners. Section dealing with licensing etc. not applicable to Australia.

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ELECTRONIC SYNTHESISER PROJECTS

For the electronic music enthusiast, an invaluable reference. This book is full of circuits and information on how to build analogue delay lines, sequencers, VCOs, envelope shapers, etc. etc. The author takes a clear and logical approach to the subject that should enable the average enthusiast to understand and build up what appears to be a quite complex instrument.

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DIGITAL IC PROJECTS

Companion to No. 225 Practical Introduction to Digital ICs and BP61 Beginner's Guide to Digital Electronics. The projects included in this book range from simple to more advanced projects — some board layouts and wiring diagrams are included. The more ambitious projects have been designed to be built and tested section by section to help the constructor avoid or correct any faults that may occur.

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INTERNATIONAL TRANSISTOR EQUIVALENTS GUIDE

Companion to BP1 and BP14 equivalents books, but contains a huge amount of information on modern transistors produced by over 100 manufacturers. Wherever possible, equivalents are subdivided into European, American and Japanese types. Also shown are the material type, polarity, manufacturer and indication of use or application.

BP85 \$9.76

SIMPLE LED CIRCUITS — BOOK 2

Sequel to BP42. Further light-emitting diode circuits. If you liked BP42 you'll love this one. If you don't know either it's well worth buying both!

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HOW TO USE OP-AMPS

Design notes and applications on many topics including basic theory, amplifiers, power supplies, audio circuits, oscillators, filters, computers and control engineering. It's written around the 741 IC but includes design notes for most of the common op-amps.

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EASY ELECTRONICS-CRYSTAL SET CONSTRUCTION

For those who wish to participate in the intricacies of electronics more through practical construction than by theoretical study. The circuits are based on those from earlier publications but have been modified to use modern inexpensive components and home wound coils.

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ELECTRONIC CIRCUITS FOR MODEL RAILWAYS

Constructional details of a simple model train controller, a controller with simulated inertia, a high-power controller, an electronic steam whistle and a 'chuff generator'. Signal systems and train lighting and RF suppression also covered.

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MINI-MATRIX BOARD PROJECTS

This book provides a selection of 20 useful circuits which can all be built on a mini-matrix board which is just 24 holes by 10 copper strips in size. Simple and easy for those with not much experience in electronics.

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AN INTRODUCTION TO VIDEO

This book is written in layman's language and is for anyone who is thinking about buying or renting or who has just bought or rented a video recorder and wants to get the best out of the machine.

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THE 6809 COMPANION

It is not a beginners introduction to microprocessors in general but a discussion of the features of the 6809 and a reference work for the 6809 programmer in particular.

BP102 \$6.56

AERIAL PROJECTS

Practical aerial designs including active, loop and ferrite which are relatively simple and inexpensive to build. The complex theory and mathematics of aerial design have been avoided.

BP105 \$6.56

THE ART OF PROGRAMMING THE 1K ZX81

This book explains how to use the features of the ZX81 including its random number generator, graphics and timer. PEEK and POKE are explained and you should learn enough to develop programs of your own.

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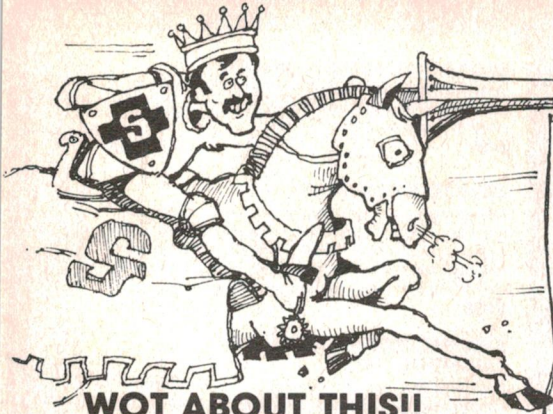
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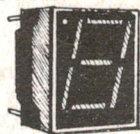
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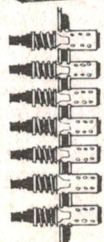
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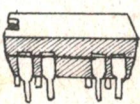
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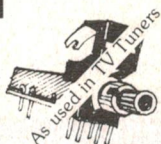
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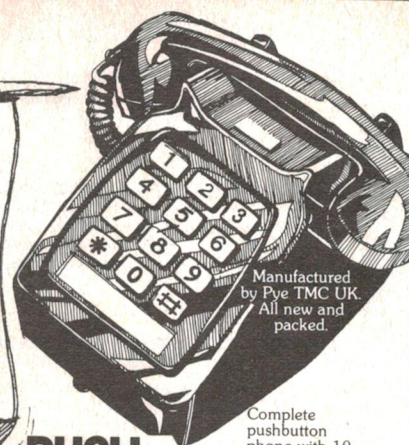
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☆ ETI-647 SPEECH SYNTHESISER ☆ (Sept. issue)

☆ Sydney readers unable to find a supplier for this project will be happy to know that kits can be obtained direct from ETI's offices. Cost is \$250. You'll find us at 4th Floor, 15 Boundary St, Rushcutters Bay. Southern state readers are directed to Rod Irving Electronics. ☆

THIS PAGE is to assist readers in the continual search for components, kits and printed circuit boards for ETI projects. If you are looking for a particular component or project — check with our advertisers if it is not mentioned here.

ETI-644 direct-connect modem

This project can be obtained in complete kit form from Jaycar in Sydney, Altronics in Perth and Microtrix and Rod Irving Electronics in Melbourne. As advised in the article, pc boards are obtainable direct from ETI, for those with a shelf full of parts able to supply most themselves. Boards cost \$50 each, post free, from:

ETI-644 pc board
ETI Magazine
154 Clarence St
Sydney NSW 2000

Most of the parts are fairly stock-standard items, even the 2% (or 1%) resistors, as are many of the ICs

employed. A number of items you will notice, however, are not common stock items, but we've hunted around for sources. Firstly, there seem to be only two sources currently for the Arlec 45035 600:600 ohm isolating transformer — Jaycar in Sydney (who have them in stock) and Kalextronic (who will order them for you) of 101 Burgundy St, Heidelberg Vic. 3084, (03)458-2976 (they're also in Melton). The Fujitsu micro reed relays and DPDT relay used in the line interface are distributed by IRH and plenty of stocks are held so you can order them through your favourite supplier. Same goes for the 5.07 MHz Ceralock resonator and CSC300K loading capacitor pack. If you'd rather use a quartz crystal, 5.0688 MHz crystals are available from Ellistronics in Melbourne and Applied Technology in Sydney.

As for the semiconductors, most are readily available over the counter, with the exceptions of the TL064 and ICL7612 or 7611. For TL064s try Rod Irving

Electronics in Melbourne or Jaycar in Sydney. The Intersil 7612 or 7611 is available from R&D Electronics; in Melbourne they're located at 257 Burwood Hwy, Burwood 3125; in Sydney at 133 Alexander St, Crows Nest 2065. Note that these ICs are not cheap.

ETI-686 EPROM programmer

Designed to team with the ETI-685 2650 S100 Single Board Computer (December '81 ETI), but suitable for any computer with three programmable 8-bit I/O ports, this programmer is simple to build and operate and low in cost. The designer, Ron Koenig, has retained copyright on the pc board design and will be supplying pc boards, wholesale and retail. If you want to arrange your own parts and buy a pc board from him, write to R&S Koenig Computer Products, P.O. Box 363, Hornsby NSW 2077. He also has software for the programmer available on either disk or tape.

Most of the components for the project are widely available, but not all electronics suppliers carry things like ZIF sockets or SIP resistors. Most of the components should be obtainable from Applied Technology in Sydney or Data Parts in Shepparton, Victoria. Rod Irving Electronics has indicated they'll be carrying a kit. If you strike trouble obtaining the BY257 bridge rectifier, there are a number of equivalents in similar packages with in-line pins. Tandy stores stock a range, catalogue numbers 276-1146, 276-1171 and 276-1173. They cost around \$2 or so.

ETI-494 loudspeaker protector

Save yourself the anguish of replacing expensive burnt-out speaker drive units — build yourself this signal-powered loudspeaker protector. Better still, build several and mount one inside each speaker!

Components for this project are all off-the-shelf items and you should have no trouble assembling the parts. Printed circuit boards will be available from suppliers listed in this column previously.

Kit and component suppliers have generally indicated they'll be stocking this project as a kit, or at least have the pc boards available, plus the components in stock.



Here is an extraordinarily useful tool for the electronics hobbyist and/or home handyman. It's a thermoplastic glue gun. For those who've never used or heard of thermoplastic glue before, it's almost a 'universal' glue, 'bonding' readily to most natural and man-made materials. Thermoplastic adhesives are used by model-makers and display artists, glaziers, mirror-makers, electrical equipment manufacturers, bookbinders, furniture-makers etc, etc. This glue gun has only recently been released in Australia and, having used one for the past month, it's a wonder how we got along without it! (We left those niggardly jobs for a rainy day!). The gun is German-made, distributed here under the Homelec name. Two varieties of glue are available to cope with a wide variety of material. It comes in 'sticks' that insert at the rear of the barrel. Hot glue comes out of the tip. Available from hardware stores. Further details from Homelec Products, 1073 Victoria Rd, West Ryde NSW 2114.

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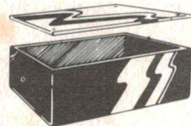
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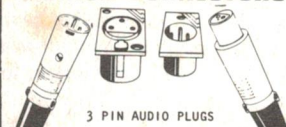
12⁰⁰

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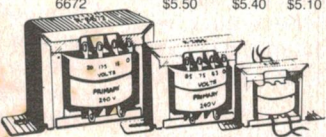
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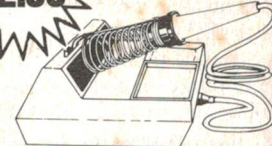
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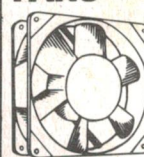
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| IN4004 | .07c | .06c | .05c | .04c |
| IN4005 | .09c | .08c | .07c | .06c |
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8062A

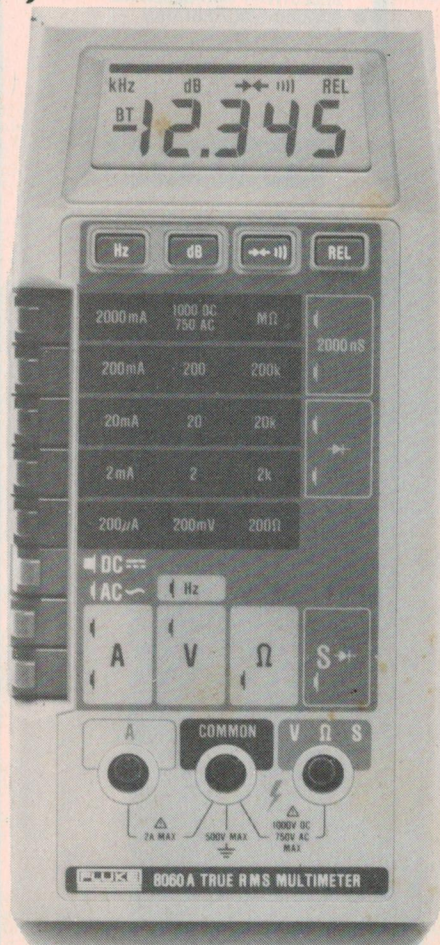
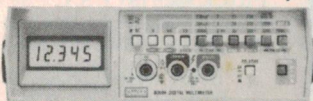
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TELEPHONE

Hi, I'm Col Beeforth from Parameters and I'd like to say something about our Thurlby multimeter.



Another one? They're a dime a dozen these days you say. Digital meters may be, but quality design and precision are not so commonly available. Thurlby have not 'cobbled up' this meter around the application notes of an integrated analogue to digital converter chip, it has been designed from first principles. The 1503 uses a dual slope A/D converter implemented with selected FET input op-amps and CMOS bilateral switches. Every aspect of the A/D converter was designed not simply accepted along with an LSI chip. The result speaks for itself. Reliable auto-zero, superb linearity, accuracy well beyond normal requirements and real repeatability of measurement. The design philosophy sounds fine but what practical effect does it have on DMM intended for real world usage?

A neat enclosure and compact size help to make the 1503 well suited for both bench and portable jobs. It is not supplied with rechargeable batteries, however the life of dry cells will be long as the current consumption is typically less than 22 mA. Provision is made for external DC power and any 9 V DC plug pack would be readily pressed into service. The front panel is neat without being gaudy. Selection of the current range is unusual, pressing volts and ohms selectors simultaneously. Not familiar, but quite functional. Like other features of the Thurlby meter, it is not obvious, but if you read the manual and use the meter it works just fine. For example, the 1503 can provide a 3200 mV range with an input impedance of more than 1000 MΩ. This is achieved with a little unorthodox manipulation of the selector buttons and is very useful around instrument amplifier inputs or tracking down leakage in high impedance circuits.

Auto-ranging is a useful feature of many DMM's, however it would be wasted on the 1503. Due to the large full scale count of 32000 the resolution on the 320 V FS range is 10 mV. For low level circuits the 32 V FS range allows a 1 mV resolution. Range changing is an infrequent operation. Diode testing is often a problem with DMM's. On the 3200 mV range, the 1503 uses a test current of 1 mA and reads the junction forward voltage drop directly in millivolts. With a little caution, like remembering to turn the power off first, a lot of semiconductor testing can be performed in circuit. This is typical of the 1503. Its advantages are not readily obvious reading the sales brochure. The convenience of the instrument only becomes obvious when it is put to work.

One question remains. Do I know what I'm talking about? I have owned and used a 1503 in daily service work for the last six months. When it comes to range selection I am worse than most. Connect the meter first, then select the range. To date, I have blown the current overload fuse three times. I also observed the claim in the specifications about the ohms range being protected to 370 V peak. Masochism is not a strong point in my personality, however curiosity is. I guess you could say the devil made me do it. With 320K ohms selected and a wicked gleam in the eye the probes were plugged into a live 240 V mains outlet. My expectations of a pyrotechnic display were replaced with respect when the instrument continued to function perfectly. Likewise, I have accidentally measured 2 KV DC when the manufacturer expressed the wish not to have more than 1200 V applied to the instrument. The fact that the 1503 is still operating perfectly, despite all efforts to do it a mischief, only confirms my opinion that it is remarkably reliable, well designed, and a very usable digital multimeter.

Don't take my word for it, you can check them out through any Parameters location, they're right throughout Australia.



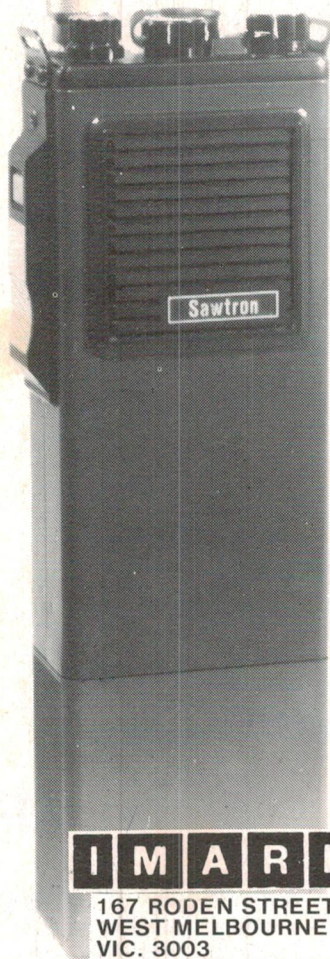
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Mechanically rugged the SX-200 uses high quality double-side Epoxy-Glass printed circuit boards throughout. Some of its other outstanding features include 3 MODE SQUELCH circuitry which allows the lockout of spurious and carrier only signals, extremely low spurious count, AM and FM detection on all bands, FINE TUNING control for off channel stations, 240 VAC on 12 Volt DC operation, Accurate QUARTZ CLOCK, Squelch operated OUTPUT for switching a tape recorder etc, 16 Memory channels, MEMORY BACKUP which lasts up to two years, high SENSITIVITY and SIGNAL-TO-NOISE ratio on all bands, CRYSTAL FILTER for excellent SELECTIVITY and easy servicing due to component layout as well as a 90 day warranty.

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1296 MHz Hawaii-California, one way

The KH6HME beacon in Hawaii was heard on the US mainland on 6 August by Chip Angle, N6CA, and several other California amateurs. The path is nearly 4000 km long (2475 miles).

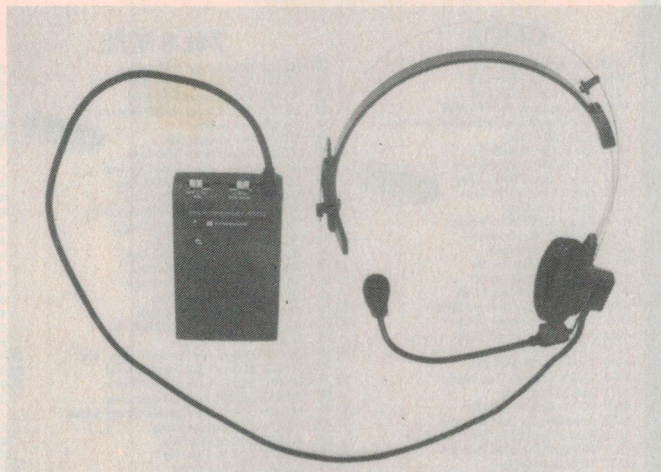
The beacon runs 50 watts into a quad set of 25 element loop yagi antennas. It was received in the Los Angeles area by N6CA using a single 44 element loop yagi.

It was not possible to try for a two-way contact since Paul Lieb, KH6HME (who operates the beacon) was visiting the mainland at the time.

An attempt at the two-way record was planned during August but no results are to hand at time of going to press. Equipment used at both ends of the path was developed by N6CA.

Current claimants to the world 1296 MHz two-way distance record are Dick Norman VK2BDN and Brian Ryall ZL1AVZ who had a contact over a distance of 2134 km on 9 February last (reported in ETI, April issue, p.8). I say claimant because the Federal body of the Wireless Institute of Australia hasn't decided yet who really has the record, nor have any counter claims been published.

Dick Norman is pressing ahead with efforts to span the Tasman sea to New Zealand on 2300 MHz. (US details from Westlink Report.)



'Hands free' personal communications

GFS Electronic Imports recently announced the release of a new personal mobile VHF FM transceiver, the C-900 'Talkman'.

Designed to provide its user with two way communications over distances up to one kilometer, the Talkman is said to be extremely simple to operate particularly because it makes use of a light weight headset.

This, coupled with its compact size, light weight and voice-operated transmitter make it suitable for hundreds of different communications applications, especially where 'hands free' operation is required,

according to the importers.

The Talkman is approved by D.O.C. and operates on the 55 MHz band. The only two controls include a volume-level switch and a VOX sensitivity switch.

Weighing only 250 grams the Talkman will conveniently clip onto a belt or into a pocket.

For further details, contact GFS Electronic Imports, 15 McKeon Road, Mitcham Vic. 3132. (03) 873-3939.

UHF Yagi antennas from Scalar

Scalar has released a new range of Y400 series UHF yagi antennas designed for use in the 400 — 520 MHz band with quoted gains ranging from 3 dB to 14 dB.

They are manufactured from high grade seamless aluminium tubing and feature a 4% bandwidth at a VSWR of less than 1.5:1, 1.3:1 at centre frequency, according to the specifications. Special heavy duty models featuring stainless steel construction are also available.

A cable tail-to-N-type female termination is provided allowing easy weatherproofing.

The Y415PT is a special 'RF control' model designed for applications where a tightly controlled beam pattern is required and meets

Department of Communications Draft Specification RB234C.

The Y415PT is a 15 element design having a multi-element reflector. Sidelobe levels at any angle greater than 55° from the centre of the main lobe are quoted to be at least 17 dB below forward gain. The Y415PT can be supplied either as an end-mount or with a centre-mount elbow.

Enquiries to Scalar office in Melbourne, Sydney, Brisbane or Perth.

Tx/Rx multicouplers

Vicom International now distributes the range of multicouplers and duplexers manufactured by Tx/Rx Systems Inc of USA.

Specific filters include bandpass cavity, series notch, varinotch and T-Pass, all of which can be cascaded to achieve an arithmetic sum of individual attenuation.

Included in the range is a series of expandable receiver multicouplers of modular design capable of handling 4 — 64 channels. Gain is adjustable for optimum receiver performance with no degradation in 1M specification it is claimed. A

typical figure is 20 dB isolation between channels and a 3rd-order intercept point of +46 dBm is said to be achieved.

Models of multicouplers are available to cover both receiver and transmitter functions up to 1000 MHz.

Further details and pricing on the extensive range is available from Vicom International, Melbourne and Sydney. (03)62-6931; (02)437-2766.

New rig and antenna from Yaesu

The world-famous Japanese amateur gear manufacturer Yaesu has released a new transceiver, the FT-102, and an active antenna, the FRA-7700.

The FT-102 transceiver covers 160 - 10 m, including the three new ('WARC') bands, with a quoted 240 W input and reportedly 'impressive' receiver specs.

The rig is an all-mode unit, CW/SSB in the basic rig, AM/FM being available as an option. (Seems to foreshadow a VHF/UHF transverter?).

Accessories include an external synthesised VFO (FV-102DM) and an external speaker with audio filter (— that's new).

Interestingly, Yaesu claim the rig's noise blanker is highly effective

against the Russian Woodpecker and other impulse noise.

The FRA-7700 active antenna is designed for use with Yaesu's popular deluxe general coverage receiver, the FRG-7700. The unit uses a 1.2 metre tall whip antenna coupled to a low distortion MOSFET preamp.

If you can't — or don't want to, erect an outdoor antenna, then this allows you to pull in the signals — even with the whip mounted indoors. It comes with gain and peaking controls and you can turn the preamp on or off.

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| 4008 | 4-Bit Full Adder | |
| 4009 | Hex Buffer (Inverting) | |
| 4010 | Hex Buffer (Non-Inverting) | |
| 4011 | Quad 2 Input NAND Gate | |
| 4012 | Dual 4 Input NAND Gate | |
| 4013 | Dual D Flip Flop | |
| 4014 | 8-Stage Static Shift Register | |
| 4015 | Dual 4-Bit Static Register | |
| 4016 | Quad Analog Switch | |
| 4017 | Decade Counter/Divider | |
| 4018 | Resettable Divide-by-N | |
| 4019 | Quad AND OR Select Gate | |
| 4020 | 14-Stage Binary Counter/Divider | |
| 4022 | Divide-by-8 Counter/Divider | |
| 4023 | Triple 3-Input NAND Gate | |
| 4024 | 7 Stage Binary Counter/Divider | |
| 4025 | Triple 3-Input NOR Gate | |
| 4026 | Decade Counter-7-Seg Output | |
| 4027 | Dual JK Flip Flop | |
| 4028 | BCD-to-Decimal Decoder | |
| 4029 | Resettable Up-Down Counter | |
| 4030 | Quad 2 Input EXCLUSIVE OR Gate | |
| 4031 | 64 Stage Static Shift Register | |
| 4032 | Triple Serial Adder | |
| 4034 | 8 Stage TRI STATE Universal BUS Register | |
| 4035 | 4 Bit Shift Register | |
| 4038 | Triple Serial Adder (Negative Logic) | |
| 4040 | 12 Stage Binary Counter | |
| 4042 | Quad Clocked D Latch | |
| 4046 | Micropower Phase-Lock Loop | |
| 4047 | Low Power Mono/Astable Multi | |
| 4049 | Hex Inverting Buffer | |
| 4050 | Hex Buffer (Non-Inverting) | |
| 4051 | 8 Channel Analog Multiplexer | |
| 4053 | Triple 2 Channel Analog Multiplexer | |
| 4055 | Quad Analog Switch | |
| 4067 | 1-of-16 Analog Switch | |
| 4068 | 8 Input NAND Gate | |
| 4069 | Hex Inverter | |
| 4070 | Quad 2 Input EXCLUSIVE OR Gate | |
| 4071 | Quad 2 Input OR Gate | |
| 4073 | Triple 3 Input AND Gate | |
| 4075 | Triple 3 Input OR Gate | |
| 4076 | Quad D Flip Flop | |
| 4077 | Quad EXCLUSIVE NOR Gate | |
| 4078 | 8-Input NOR Gate | |
| 4081 | 2-Input AND Gate | |
| 4093 | Quad 2-Input NAND Schmitt Trigger | |
| 4411 | Decade Counter 7 Seg Output | |
| 4500 | 4501 Triple Gate | |
| 4502 | Strobed Hex Inverter | |
| 4503 | Hex 3 Stage Buffer (Non Inverting) | |
| 4508 | Dual 4-Bit Latch | |
| 4511 | BCD to 7 Seg Decoder/Driver | |
| 4512 | 8 Channel Data Selector | |
| 4514 | 4 to 16 Line Decoder/4 Bit Latched | |
| 4517 | Dual 64 Bit Static Shift Register | |
| 4518 | BCD Up Counter | |
| 4520 | Dual Binary Up Counter | |
| 4526 | Programmable 4-Bit Binary Counter | |
| 4528 | Dual Mono Multi | |
| 4538 | Dual Mono/Precision Multi | |
| 4543 | BCD to 7 Seg Latch/Decoder/Driver/LCD | |
| 4553 | 3 Digit BCD Counter | |
| 4557 | 1 to 64 Bit Variable-Length Shift Register | |
| 4581 | 4 Bit | |
| 4582 | Look Ahead Carry Block | |
| 4584 | Hex Schmitt Trigger | |
| 74C02 | Quad 2 Input NOR Gate | |
| 74C04 | Hex Inverter | |
| 74C14/40106 | Hex Schmitt Trigger | |
| 74C922 | 16 Keyboard Encoder | |

SUPPORT

| | | |
|--------|---------------------------------|---------|
| 8255 | PPI | \$9.95 |
| 5303 | UART | \$12.95 |
| 1771 | Floppy Disk Controller | \$19.95 |
| 1488 | RS232 Quad Line Driver | \$1.50 |
| 1489 | RS232 Quad Receiver | \$1.50 |
| 8131 | 6-Bit Comparator (Negative Out) | \$4.95 |
| 81LS97 | Octal Buffer | \$2.35 |
| 81LS98 | Octal Buffer (Inverting) | \$2.35 |

74LS TTL

| | | |
|---------|--|--|
| 40 | 74LS00 Quad 2 Input NAND Gate | |
| 35 | 74LS01 Quad 2 Input NOR Gate | |
| 48 | 74LS02 Quad 2 Input NOR Gate | |
| 40 | 74LS03 Quad 2 Input NAND Gate (oc) | |
| 38 | 74LS04 Hex Inverter | |
| 95 | 74LS05 Hex Inverter (oc) | |
| \$1.10 | 74LS08 Quad 2 Input AND Gate | |
| 90 | 74LS09 Quad 2 Input AND Gate (oc) | |
| 35 | 74LS10 Triple 3 Input NAND Gate | |
| 20 | 74LS11 Triple 3 Input AND Gate | |
| 90 | 74LS12 Triple 3 Input NAND Gate (oc) | |
| \$1.00 | 74LS13 Dual NAND Schmitt Trigger | |
| 95 | 74LS14 Hex Schmitt Trigger | |
| \$1.20 | 74LS20 Dual 4 Input AND Gate | |
| 50 | 74LS27 Triple 3 Input NOR Gate | |
| 75 | 74LS30 8 Input NAND Gate | |
| \$1.45 | 74LS32 Quad 2 Input OR Gate | |
| \$1.25 | 74LS33 Quad 2 Input NOR Gate (oc) | |
| 28 | 74LS42 1 of 10 Decoder | |
| \$1.18 | 74LS47 BCD to 7 Seg Decoder/Driver (oc) | |
| 35 | 74LS48 BCD to 7 Seg Decoder/Driver | |
| \$2.20 | 74LS49 BCD to 7 Seg Decoder/Driver | |
| \$1.18 | 74LS73 Dual JK Flip Flop with Clear | |
| \$1.25 | 74LS74 Dual D Flip Flop with Preset & Clear | |
| \$1.18 | 74LS75 4 Bit Bistable Latch | |
| \$2.20 | 74LS76 Dual JK Flip Flop | |
| \$1.95 | 74LS83 4 Bit Full Adder | |
| \$2.95 | 74LS85 4 Bit Magnitude Comparator | |
| 50 | 74LS86 Quad 2 Input EXCLUSIVE OR Gate | |
| \$1.95 | 74LS90 Decade Counter | |
| \$1.65 | 74LS92 Divide by 12 Counter | |
| 70 | 74LS93 4 Bit Binary Counter | |
| \$1.55 | 74LS95 4 Bit Shift Register | |
| \$1.75 | 74LS107 Dual JK Flip Flop | |
| 75 | 74LS109 Dual JK Edge-Triggered Flip Flop | |
| 60 | 74LS112 Dual JK Edge-Triggered Flip Flop | |
| 95 | 74LS122 Retrigger Monostable with Clear | |
| \$1.75 | 74LS123 Quad Retrigger Mono with Clear | |
| 50 | 74LS125 Quad BUS Buffer Tri-State (lo enable) | |
| \$2.95 | 74LS126 Quad BUS Buffer Tri-State (hi enable) | |
| 48 | 74LS139 Dual 4 Line Decoder/Multiplexer | |
| 70 | 74LS147 10 Line dec to 4 Line Priority Encoder | |
| 48 | 74LS148 8 Line to 3 Line Priority Encoder | |
| 48 | 74LS151 1 of 8 Selector/Multiplexer | |
| 65 | 74LS153 Dual 4 Line to 1 Line Selector/Multiplexer | |
| 60 | 74LS155 Dual 1 of 4 Decoder | |
| \$1.25 | 74LS156 Dual 1 of 4 Decoder (oc) | |
| 65 | 74LS157 Quad 2 Input Multiplexer (Non-Inverting) | |
| 60 | 74LS160 BCD Decade Counter Asynch. Reset | |
| 80 | 74LS161 4-Bit Binary Counter Synchron. Reset | |
| \$17.75 | 74LS162 BCD Decade Counter Synchron. Reset | |
| \$2.20 | 74LS163 4-Bit Binary Counter Synchron. Reset | |
| \$2.50 | 74LS164 8-Bit Shift Register | |
| \$1.95 | 74LS165 Parallel Load 8-Bit Shift Register | |
| \$1.95 | 74LS166 8-Bit PISO Shift Register | |
| 95 | 74LS173 4-Bit D Type Register | |
| \$5.00 | 74LS175 Quad D Type Flip Flop | |
| \$1.50 | 74LS190 Up/Down Decade Counter | |
| \$1.25 | 74LS191 Up/Down Binary Counter | |
| \$2.50 | 74LS192 Up/Down Decade Counter Dual Clock | |
| \$1.95 | 74LS193 Up/Down Binary Counter Dual Clock | |
| \$1.50 | 74LS194 4-Bit Left-Right Shift Register | |
| \$1.50 | 74LS221 Dual Monostable Multivibrator | |
| \$1.50 | 74LS240 Octal Tri-State Driver | |
| \$1.25 | 74LS241 Octal Tri-State Driver | |
| \$2.65 | 74LS244 Octal Buffer/Driver | |
| \$1.95 | 74LS245 Octal Bus Transceiver | |
| \$6.50 | 74LS257 Quad 2 Input Multiplexer Tri-State | |
| \$3.50 | 74LS258 Quad 2 Input Multiplexer Tri-State | |
| \$3.50 | 74LS259 8 Bit Addressable Latch | |
| \$2.50 | 74LS266 Quad EXCLUSIVE NOR Gate (oc) | |
| \$1.65 | 74LS273 Octal D Type Flip Flop | |
| 50 | 74LS365 Hex Buffer with Common Enable | |
| 55 | 74LS366 Hex Inverter with Common Enable | |
| \$1.50 | 74LS367 Hex Buffer 4 Bit and 2 Bit | |
| \$9.50 | 74LS373 Octal Transparent Latch | |
| | 74LS374 Octal D Type Flip Flop | |
| | 74LS393 Dual LS93 | |

SPEECH SYNTHESISER

| | | |
|---|--|---------|
| DT1050 | National Digitaltalker Processor & ROM Set | \$95.00 |
| DT1057 | National Digitaltalker Second ROM Set | \$69.00 |
| Send SAE for more details. NOTE new low price. Data sheets included in chip sets. | | |

LINEAR DEVICES

| | | |
|--------|---------|--------|
| 35 | ZN414 | \$2.50 |
| 40 | LM381N | \$2.45 |
| 40 | LM382 | \$2.45 |
| 60 | LM384CH | \$4.75 |
| 40 | NE555 | 98 |
| 60 | LM555 | \$1.45 |
| 40 | LM556 | \$3.50 |
| 50 | LM567 | \$3.50 |
| 40 | LM570 | \$7.50 |
| 40 | LM571 | \$5.50 |
| 40 | LM572 | \$2.95 |
| 60 | LM741 | \$3.50 |
| 60 | LM747 | \$3.50 |
| 85 | LM1496 | \$1.95 |
| 35 | LM3046 | \$1.25 |
| 60 | LM3080 | \$1.95 |
| 40 | LM311 | \$1.80 |
| 40 | LM313 | \$1.80 |
| 45 | LM314 | \$1.45 |
| 95 | LM334 | \$2.50 |
| 95 | LM339 | \$1.95 |
| \$2.95 | LF374N | \$2.50 |
| \$1.95 | LM348 | \$2.50 |
| 95 | OM350 | \$9.50 |
| 75 | LM361 | \$3.75 |
| 65 | LM380 | \$1.48 |

DISCRETE DEVICES

| | | |
|--------|---------------|---------|
| \$1.25 | VN88AF | \$3.00 |
| 50 | BCY71 | 95 |
| 75 | JDT9203 | \$1.95 |
| \$1.25 | 40411 | \$5.00 |
| 85 | MJ15003 | \$4.75 |
| 75 | MJ15004 | \$4.75 |
| 75 | 2S49(P) | \$6.50 |
| 75 | 2B50 | \$8.50 |
| 60 | BDV648 | \$5.95 |
| 75 | BUX80 | \$9.50 |
| \$1.10 | ZK134(N) | \$6.50 |
| 55 | BU105 | \$12.50 |
| 75 | MPSA05 | 95 |
| 92 | TP131A&B | 90 |
| \$1.45 | TIP32A | 90 |
| \$2.95 | TIP2955 | \$1.20 |
| \$3.95 | TIP3055 | \$1.20 |
| 75 | TIP3055 | \$1.20 |
| 75 | MPF102 | 95 |
| \$1.00 | MPF106/2N5459 | 85 |
| \$1.00 | MPF106/2N5485 | 78 |
| 79 | MEF131 | \$1.98 |
| 95 | BF200 | 75 |
| 95 | BF469 | \$1.65 |
| \$1.20 | BF470 | \$1.98 |
| \$1.20 | MJE340 | \$1.98 |
| 60 | MJE350 | \$1.98 |
| \$2.95 | ZTX301 | \$4.75 |
| 95 | BC337 | 75 |
| \$1.65 | BC546 | \$1.65 |
| \$1.25 | BF469 | \$1.65 |
| \$1.15 | BF470 | \$1.65 |

SPECIAL FUNCTION

| | | |
|---------|------------|--------------------------|
| \$9.95 | SAB0600 | Door Chime |
| \$9.50 | AY-1-5050 | |
| \$17.50 | TEA-1022 | PAL Colour Encoder |
| \$29.50 | EXAR 2206 | Function Generator |
| \$5.95 | MOC-3020 | TRIAC Opto Coupler |
| \$2.50 | MM 5837 | Noise Generator |
| \$3.90 | TA 7205P | Audio Amp |
| \$4.50 | 7216A | Frequency Display Driver |
| \$47.50 | SN 76488NF | Sound Effects Generator |
| \$5.95 | AY-1-0212 | Top Octave Synthesiser |

Z-80A/6800

| | | |
|---------|------------------|---------|
| \$8.50 | Z-80A CPU 4MHz | \$8.50 |
| \$8.50 | Z-80A PIO 4MHz | \$8.50 |
| \$8.50 | Z-80A CTC 4MHz | \$8.50 |
| \$20.95 | Z-80A SIO/0 4MHz | \$20.95 |
| \$20.95 | Z-80A SIO/1 4MHz | \$20.95 |
| \$20.95 | Z-80A SIO/2 4MHz | \$20.95 |

ZENER DIODES 1 WATT

From 3.3V to 33V

35

MEMORIES

| | | |
|--------|---------------------------|---------|
| 2102 | 1K x 1 RAM | \$1.95 |
| 2114-3 | 1K x 4 RAM 350ns | \$2.45 |
| 4116 | 16K x 1 Dynamic RAM 250ns | \$2.50 |
| 6116 | 2K x 8 CMOS RAM | \$2.50 |
| 2708 | 1K x 8 EPROM | \$7.95 |
| 2716 | 2K x 8 EPROM (Single +5V) | \$6.50 |
| 2532 | 4K x 8 EPROM (Single +5V) | \$14.95 |
| 2564 | 8K x 8 EPROM (Single +5V) | \$35.00 |

VOLTAGE REGULATORS

| | | |
|---------|-------------------------------|---------|
| 7805 | 5V+ Regulator TO-220/amp | \$1.05 |
| 7805 | 5V- Regulator TO-220/amp | \$1.65 |
| 7812 | 12V+ Regulator TO-220/amp | \$1.05 |
| 7812 | 12V- Regulator TO-220/amp | \$1.65 |
| 7815 | 15V+ Regulator TO-220/amp | \$1.05 |
| 7815 | 15V- Regulator TO-220/amp | \$1.65 |
| 78L12 | 12V+ Regulator 100mA | \$1.00 |
| 78L15 | 15V+ Regulator 100mA | \$1.00 |
| 78L15 | 15V- Regulator 100mA | \$1.00 |
| LM323 | 3amp 5V TO-3 | \$7.50 |
| LM309 | 1.5amp 5V TO-3 | \$1.50 |
| LM317T | 2.30V adjustable 1 amp TO-220 | \$2.80 |
| LM317KC | 2.30V adjustable 3 amp TO-3 | \$6.50 |
| LM337K | 3 amp variable 2.30V TO3 | \$9.50 |
| LM396K | 10 amp variable 1.25-15V TO3 | \$19.50 |

LED's

| | | |
|------------|--------------------------|----|
| 5mm Red | 15 3mm Yellow | 28 |
| 5mm Green | 26 Rectangular Red | 40 |
| 5mm Yellow | 30 Rectangular Green | 40 |
| 5mm Orange | 28 Rectangular Yellow | 40 |
| 3mm Red | 18 Flashing Red | 49 |
| 3mm Green | 28 Flashing Red 10+ each | 45 |

BRIDGE RECTIFIERS

| | |
|---------|--------|
| W04 | 68 |
| P04 | \$2.95 |
| MDA3504 | \$4.50 |

OPTO

| | | |
|--------|----------------|--------|
| COY89 | Infr Red LED | \$1.25 |
| FND500 | | \$1.35 |
| BPW50 | Infr Red Diode | \$2.95 |
| FND507 | | \$1.75 |
| LT327 | 7 seg Display | \$1.45 |
| LT547 | | \$2.95 |
| LT303 | 7 seg Display | \$1.45 |

DIODES

| | | |
|--------|---------------|--------|
| IN4002 | 8BA102 | 25 |
| IN4006 | 20 BA114 | 25 |
| IN5404 | 35 BA215/219 | 25 |
| IN914 | 5 8R100 | 95 |
| BAW62 | 29 BYX21L/200 | \$1.90 |
| | 5082-2800 | \$2.75 |

IC SOCKETS - SOLDER

| | | | | | |
|--------|----|--------|----|--------|----|
| 8 Pin | 25 | 16 Pin | 35 | 24 Pin | 50 |
| 14 Pin | 30 | 20 Pin | 38 | 28 Pin | 60 |
| 16 Pin | 35 | 22 Pin | 40 | 40 Pin | 70 |

QUALITY WIRE WRAP IC SOCKETS

| | | | |
|--------|--------|--------|--------|
| 14 Pin | 85 | 22 Pin | \$1.40 |
| 16 Pin | \$1.05 | 24 Pin | \$1.75 |
| 18 Pin | \$1.35 | 28 Pin | \$2.25 |
| 20 Pin | \$1.45 | 40 Pin | \$2.65 |

DIL PLUGS AND COVERS SOLDER TYPE

| | | | |
|--------------|----|------------------|--------|
| 14 Pin Plug | 85 | 16 Pin Cover | 25 |
| 14 Pin Cover | 25 | 24 Pin Plug Only | \$2.50 |
| 16 Pin Plug | 95 | | |

CRYSTALS

| | | |
|---------------------|-----------------------------|--------|
| Case Style: HC-33/U | 4.9152MHz | \$6.50 |
| 1.000MHz | 5.000MHz | \$5.50 |
| 1.8432MHz | 8.867238MHz | \$5.50 |
| 2.000MHz | 10.000MHz Parallel Resonant | \$9.50 |
| 3.000MHz | | \$5.50 |
| Case Style: HC-18/U | 12.000MHz | \$5.50 |
| 4.000MHz | 16.000MHz | \$5.50 |
| 4.43MHz | 20.000MHz | \$7.50 |

MINIMUM MAIL ORDER \$5.00

CARLINGFORD PHONE 872 4422

NUMBER 1 COMPONENTS

**NEW STORE
opening October**
in **Carlingford**
AT "NEVILLES CORNER" CNR CARLINGFORD &
PENNANT HILLS RD. NEXT TO KENTUCKY FRIED CHICKEN

JAYCAR PASSIVE COMPONENTS

We regret to advise that we have had to pass on drastically increased costs due to the devaluation of the Australian dollar and the savings sales tax increase.

Most of the prices on this page had not changed in 18 months. In some cases our cost has increased 50% whereas our increases average about 10%. We haven't got the heart to pass on to you the costs that were passed on to us!

We are very proud of our range of components at Jaycar. Our quality is first class - so remember that if someone offers you a component that may APPEAR cheaper. Jaycar components are 100% prime spec and guaranteed. They are the same parts that were used in TV stations, telephone exchanges and other industrial equipment.

NOTE! We have saved space in some places by saying that components conform to "E12". This means that we have 12 values in each "decade" i.e. 10, 12, 15, 18, 22, 27, 33, 39, 47, 56, 68, 82 and back to 10 in other words, 22pF, 220pF, 2200pF and so on.

POTENTIOMETERS

POTENTIOMETERS CARBON ROTARY

Jaycar rotary pots did not go metric. We keep 1/2" plain shaft types with 3/8th bush mount. Most have a flat machined on the shaft.

| LINEAR | linear | SINGLE GANG LINEAR | LOG |
|----------------------------|--------|--------------------|------|
| 500 ohms | | | 1k |
| 1k | | | 5k |
| 5k | | | 10k |
| 10k | | | 25k |
| 25k | | | 50k |
| 50k | | | 100k |
| 100k | | | 250k |
| | | | 500k |
| 500k | | | 1M |
| 1M | | | 2M |
| 2M | | | |
| Price: 1 - 9 95 cents each | | | |

| LINEAR | linear | SINGLE GANG WITH SWITCH | LOG |
|--------------------------|--------|-------------------------|------|
| 25k | | | 1k |
| | | | 5k |
| | | | 10k |
| | | | 25k |
| | | | 50k |
| | | | 100k |
| | | | 250k |
| | | | 500k |
| | | | 1M |
| | | | 2M |
| Price: 1 - 9 \$2.25 each | | | |

| DUAL GANG LINEAR | linear | LOG |
|---|--------|------|
| 5k | | 5k |
| 10k | | 10k |
| 25k | | 25k |
| 50k | | 50k |
| 100k | | 100k |
| 250k | | 250k |
| 500k | | 500k |
| 1M | | 1M |
| Price: 1 - 9 \$2.25 each | | |
| Dual gang with switch 100k log (only) \$3.95 each | | |

CARBON SLIDER CONTROLS

45mm type - some with tap.

| LINEAR | linear | LOG |
|--------------------------------------|--------|-----|
| 2k | | 1k |
| 10k | | 5k |
| 50k | | 10k |
| 100k | | 25k |
| Price: 1 - 9 \$2.95 each | | |
| 30mm type 5k linear only \$1.95 each | | |
| 60mm type 10k log only \$3.95 each | | |

WIREWOUND POTS

3 watt rating (linear). 1/2" shaft, 3/8th bush.

10 ohms, 25 ohms, 50 ohms, 100 ohms, 500 ohms, 1k, 5k, 10k, 20k, 50k, 100k, 200k, 500k, 1M, 2M.

Price: 1 - 9 \$3.45 each

Jaycar stocks two types of trimmers. The low cost "skeleton" type and the high quality "PIHER" European made dust proof type.

MINIATURE VERTICAL (SKELETON)
100 ohms, 250 ohms, 500 ohms, 1k, 2k, 5k, 10k, 25k, 50k, 100k, 200k, 500k, 1M, 2M.

Price: 1 - 9 42 cents each

LARGE VERTICAL (PIHER TYPE)
100 ohms, 200 ohms, 500 ohms, 1k, 2k, 5k, 10k, 20k, 50k, 100k, 200k, 500k, 1M, 2M.

Price: 1 - 9 48 cents each

SMALL HORIZONTAL (PIHER)

100 ohms, 500 ohms, 1k, 2k, 5k, 10k, 20k, 50k, 100k
Price: 1 - 9 45 cents each

CERMET (Ceramic/Metal Film/Oxide)

High stability trimmers in a small package. 1/3rdW dissipation, dustproof, 100ppm T.C., 5mm lead spacing (1" grid).

100 ohms, 500 ohms, 1k, 2k, 5k, 10k, 20k, 50k, 100k
Price: 1 - 9 65 cents each

10+ 60 cents each

10+ 42 cents each

10+ 60 cents each

10+ 60 cents each

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CAPACITORS

TANTALUM BEADS also called "TAG" tantalums.

This is a brand name that has become a description.

VALUE VOLTS

0.1uF 35

0.22uF 35

0.33uF 35

0.47uF 35

1uF 35

1.5uF 35

2.2uF 35

3.3uF 35

4.7uF 35

6.8uF 35

10uF 25/35

22uF 25

33uF 16

47uF 10

100uF 3

10+ 50

10+ 50

10+ 50

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ALUMINIUM ELECTROLYTICS

Both axial (RT) and single ended radial (RB) types kept. We keep a large range of the now popular RB types. Each capacitor is fitted with a durable vinyl jacket which has the value and polarity clearly marked.

AXIAL TYPES

Capacitance Voltage

1uF 50

3.3uF 25

4.7uF 25

10uF 16

100uF 63

250uF 25

330uF 10

470uF 63

1000uF 10

2200uF 25

3300uF 63

4700uF 350

10000uF 10

22000uF 50

33000uF 10

47000uF 63

100000uF 16

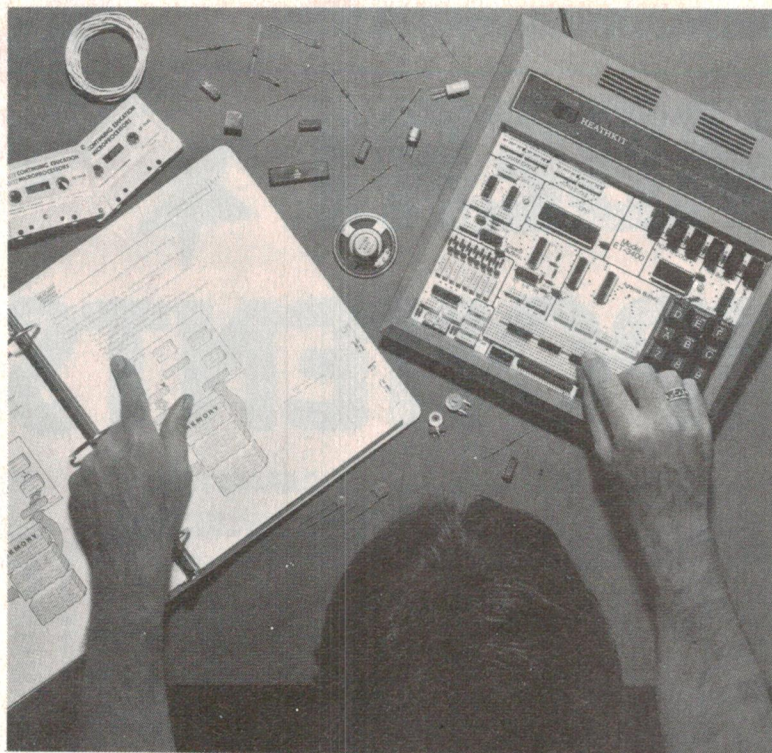
220000uF 35

330000uF 60

470000uF 16

1000000uF 35

2200000uF 35



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COMPUTING TODAY

'DAOS' program handles input from instruments or transducers

Computers in the laboratory are almost as common as microscopes. But all that hardware is not much good without the right program. 'DAOS' (Data Acquisition Operating System) from Laboratory Associates is part language, part level program and claimed to be an ideal tool for the researcher who lacks a specialist computing background.

Computer operations are simplified by the use of an extended form of BASIC and the incorporation of well over 1000 keyboard commands.

In the laboratory, 'DAOS' handles data arrays quickly and by single commands, segments of data can be multiplied, divided, rotated, integrated, differentiated, filtered, Fourier transformed and using the 'DAOS' Chebyshev transform, be fitted accurately with polynomials.

'DAOS' will allow the user to run programs and store data at the same time.

Every laboratory has its own needs and these can be catered for using the fully interactive capabilities of 'DAOS' it's claimed. Programs can be created and run without assembly, compiling or linking. They can then be stored on library files to be retrieved when needed. In this way the user can construct a powerful set of programs for any type of experiment and develop different libraries for different experiments. The system can also be linked to

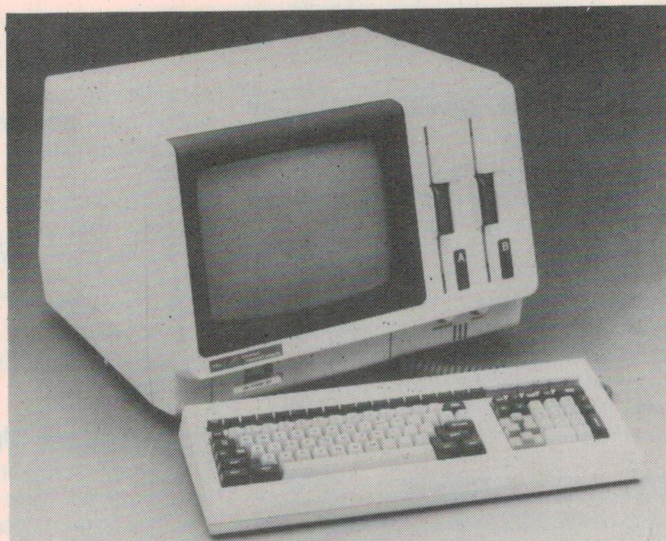
assembly language programs or FORTRAN subroutines.

The flexibility and expandability of 'DAOS' is claimed to be a real advantage, particularly in areas where research needs are liable to change eliminating the need for time consuming in-house programming which can cause costly delays in the extraction and analysis of data.

The system is available with an optional graphics facility, which provides an X-versus-t, X-versus-y and pseudo three dimensional facilities, full labelling, rotation, page allocation and interactive drawing.

At present, the program is only available on DEC processors (PDP-11 and LSI-11) but the company is in the process of adapting the system for use with Apple machines and hardware incorporating the 68000 range of microprocessors.

In Australia, 'DAOS' is distributed by Digital Electronics at a cost of approximately \$2500 including the cost of optional extras.



NEC 16-bit personal computer

NEC Information Systems Australia Pty Ltd recently announced a new 16-bit personal computer and software aimed at the business marketplace which will become available in November.

The Advanced Personal Computer is built around an NEC-manufactured, 16-bit 8086-compatible microprocessor. It is packaged in a compact, integrated enclosure with separate keyboard. The APC comes in two basic models and they both incorporate 12" (diagonal) monitors and display 25 lines of 80 characters plus a system status line.

The monochrome model, priced at around \$4000 combines a green/black high-resolution monitor, 128K of user memory, a single NEC-manufactured 1M, dual-sized 8" floppy disk drive upgradeable to a second diskette, keyboard and many other standard features.

The colour model, which includes two diskette drives, is priced at around \$6000. It is functionally identical in every way except that its high resolution monitor displays eight colours.

Standard I/O equipment includes a parallel printer controller and a serial communications controller that supports both asynchronous and synchronous communications at up to 19 200 bps.

Backed up by a 2-year life lithium battery are a 4K write-protectable CMOS RAM and hardware clock/calendar. Auto-Power-Off allows the system to turn itself off (or be instructed to do so by a remote host) at the completion of some predefined set of tasks.

Hardware options include a second 1M disk drive for the monochrome model, an additional 128K of memory, hardware graphics subsystem and a second communication controller.

Software announced for the NEC APC includes the CP/M-86 operating system from Digital Research. The company indicated that MSDOS from Microsoft would soon be offered as well as the Ryan-McFarland RM/COBOL compiler.

A large variety of application packages will be made available with first deliveries of the new product including the Benchmark word processor, Telecommunicator and Mailing List Manager from Metasoft and MicroPlan from Chang Laboratories.

For further information contact Jolyon Bone, NEC Information Systems Pty Ltd, (02)438-3544.

S100 DMA controller

AED Microcomputer Products have announced the release of a new S100 floppy disk controller which they claim has superior performance and features than the alternative designs.

The controller transfers operating instructions as well as data from and into the system memory by direct memory access. It therefore offers very fast speed and minimal use of processor time. The DMA channel that this controller operates on can be anywhere within the 24 address line space, so it is suitable for use in extended address systems, 16-bit systems, or multi-user systems.

The controller electronics is based on a Z80 instead of the more usual 1793 series of the Intel/Nec controller chips. This gives it compatibility with both double density sector header standards.

Because the controller electronics

is intelligent it is possible to change the controller parameters to suit any possible need by either reprogramming the EPROM or by down loading a controller routine directly into the on-board RAM and passing the on-board Z80 control to it. This card can control both 8" and 5¼" drives concurrently.

An advantage of this card is that it does not consume any memory space in the host computer even during the boot up procedure.

For further information about this product contact AED Microcomputer Products at 130 Military Rd, Guildford NSW. (02)681-4966.

Computer

Has the computer you want

Complete builtup computer

MicroBee is now ready built in a superb new case. And the best news of all – it costs no more.

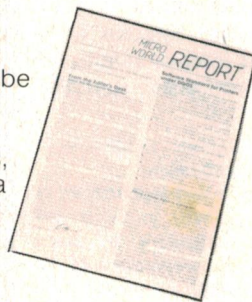
Response to the MicroBee has been so great that we are able to offer the builtup MicroBee at the same price as the kit. With over 2,000 MicroBees sold, and a government schools contract, we want to thank all those who have supported the MicroBee and made it a success.



receive the Newsletter and be in contact with other

MicroBee users, both in Australia and overseas.

MicroBee's BASIC makes it both easier to use yet more powerful than other Micros.



Powerful BASIC

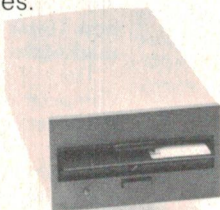
MicroBee has been developed as the finest low cost instructional computer you can buy. Vital to this is MicroBee's incredibly powerful ROM BASIC. It's advanced error reporting and editing features make running and debugging your programs quicker and more effective. The exclusive MicroBee BASIC manual means you've got a complete learning package.

You can quickly master elementary programming and move on to enter and run sample programs. And it doesn't stop there. As a MicroBee user, you are part of the MicroWorld Users Group. You will

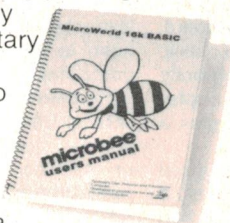


Expansion Power

Like most computer buyers, you're probably not sure just how you might be using your computer in the future. You want to keep your options open. This is where MicroBee is ahead of other machines.



If you need more memory to run longer, more advanced programs, your MicroBee's memory or 'core' board can be upgraded from 16K to 32K or 64K of RAM when you want to run longer more advanced programs. To run Z80 machine code, the Editor/Assembler ROMs



just plug straight in. Extra ROM (up to 32K can be plugged straight in to the memory board.

If you want to expand to disk drives or use the international standard S100 bus so you can use S100 products like our Digitalker (speech synthesis) or ROMBLASTER (EPROM programmer) or a host of other S100 products from other manufacturers you can. All necessary pins of the Z80 CPU chip are brought out at the back of the MicroBee. Just connect up the low cost MicroBee S100 interface and you can fit disk drives, (the disk controller board fits into the S100 expansion box, and up to three other S100 boards. To do this your MicroBee needs to be upgraded to 64K RAM. Then you can add disk drives. With your first drive you get CP/M 2.2. This opens up the huge library of world class software. Further drives can be added on later at low cost.



Continuous Memory

MicroBee's power and price come as a result of the new generation of super powerful LSI chips. The latest CMOS memory chips make battery backup of memory possible. Like your continuous memory calculator MicroBee can remember data and programs when you switch off or the power fails. Move to a new location and your program and data is still there.



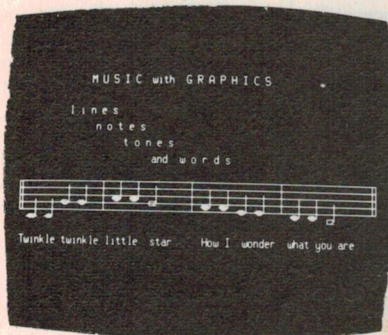
Graphics and Sound

Facilities such as the high and low resolution graphics, and built in sound are under BASIC control. So you can use them more easily. The inbuilt I/O ports are also under BASIC control so it's very easy to use a modem or printer with your MicroBee.



shopping?

got all these features?



When you want to go further, MicroWorld's Editor/Assembler helps you write in Z80 machine code. You'll be able to write USR subroutines. And the editor/assembler is in ROM for instant access.

Software Growth



Because MicroBee uses the Z80 it has a huge software base. Our expanding range of cassette software shows our commitment to the MicroBee as an inexpensive cassette based Micro. We believe cassette storage is the most effective for the enthusiast and educational environment. The cassette storage interface on the MicroBee works quickly and reliably. It is not a 'token effort' as it is on many Micros. MicroBee's interface works reliably with the sort of ordinary cassette recorder you have at home. It works at both 300 and 1200 baud, so you can store and recall your programs up to four

times more quickly than on many other Micros. The 6545 VDU controller chip gives you a fully programmable screen display with upper/lower case. The standard of 16 lines of 64 characters can be quickly and inexpensively changed to the professional 80x24 format if you want to move up to disk drives and CP/M. When you're ready, use MicroBee's optional parallel port and connect up joysticks or other peripherals to the optional parallel port.

Order Details

| | |
|--|--------------------|
| MicroBee 16K including manuals and 16K ROM BASIC | \$399.00 |
| MicroBee 32K | \$499.00 |
| MicroBee 64K | \$649.00 |
| New Moulded case | \$25.00 |
| Black and White monitor | \$139.50 |
| Green Screen Monitor | \$299.00 |
| Cassette Recorder | \$35.00 |
| MicroBee Editor/Assembler | \$49.50 |
| Cassette Software | each \$7.95 |
| Space Invaders program | \$14.75 |
| Conversion 16K to 32K | \$100.00 |
| Conversion 32K to 64K | \$155.00 |
| S100 Expansion Interface | \$299.00 |
| Disk Drive with CP/M and disk controller | \$799.00 |
| Add on disk drive | \$549.00 |

Cassette Programs

MicroBee is supported by a wide range of stimulation, games and educational cassette software. Contact us for full details. Listed below is some of our games and practice software.

Space Invaders

Concentration

Chess
Target
Star Shoot/Hangman
Eliza
MasterMind/Nim
Z Trek

Typing/Solitaire
Lunar Lander/Hurkle
Biorhythm/Calendar
Kids Game
Chase/Wumpus

New Stores

To help MicroBee customers, we've opened a number of new stores. Buying and getting support for your MicroBee is getting easier all the time. New Branches are:

Canberra: Applied Technology (ACT) Pty Ltd, 27 Colbee Court, Philip ACT. Phone (062)82 4611
Gosford 1, Debenham Road, West Gosford (043)24 1022

Artarmon: 35, Dickson Ave. Artarmon 2064 Phone (02)439 2322

Beeline

If you have technical questions on MicroBee, just phone on our new Beeline number:

(043) 24 1022

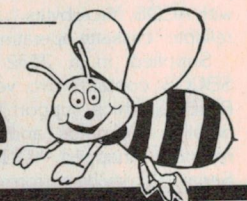


microbee apologizes

A sincere apology to all those people who have waited for their MicroBees over the past few months. It is your patience that has made MicroBee such a success. We are rapidly expanding our facilities to keep pace with the demand.

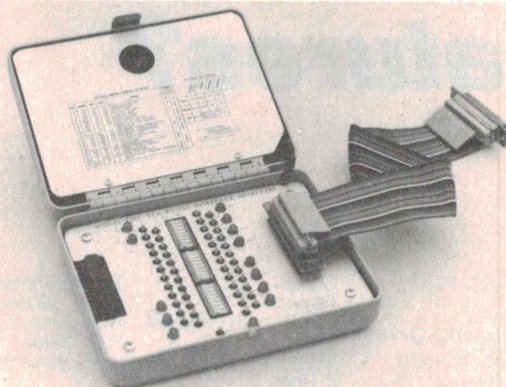
microbee

APPLIED TECHNOLOGY



Sydney Showrooms: 1A Pattison Ave. Waitara NSW 2077. Hour 9-5 Monday to Saturday. Phone (02)487 2711 (order queries), (02)487 3798 (mail orders). **Artarmon:** 35, Dickson Ave. Artarmon NSW 2064 Phone (02)439 2322
Gosford Showroom: 1, Debenham Road, West Gosford NSW 2250. Phone (043)24 1022 Mail Orders to: PO Box 311 Hornsby NSW 2077.

Print-out



Interface analyser

The Electro Standard Laboratory Model 700 EIA RS-232 interface analyser is a diagnostic tool designed for use at the standard EIA RS-232 or CCITT V.24 data interface of modems, multiplexers, terminals and computers.

It is inserted in series between the data terminal equipment (DTE) and the data communications equipment (DCE) to provide access to and monitoring of all data, timing and control signals.

The unit features state of the art

tri-state light emitting diodes to clearly display polarity, activity and validity of all interface signals.

Details from Scientific Devices Australia, 2 Jacks Rd, South Oakleigh Vic. 3167. (03)579-3622.

Colour video controller

The iSBX 270 video display controller provides low-cost, eight-colour display-terminal control for all 8 or 16-bit Multibus and iSBX-compatible systems.

The 76 x 177 mm iSBX multi-module board can interface with either colour or black and white display monitors at a 50 or 60 Hz frame rate. Up to 256 characters are contained in EPROM reprogrammable by the user for custom applications.

Three types of character font displays are supported by the iSBX 270 in matrixes of seven by nine, five by

seven, or six by eight dots.

The iSBX 270 contains a software package in its on-board 8741A microcomputer that initialises and monitors the various on-board ICs.

For further information contact T.J. Casey, Total Electronics, 9 Harker St, Burwood Vic. 3125. (03)288-4044.

Another monitor for the ETI-685

Microbyte has developed a monitor/cassette operating system for the ETI-685 2650-based S100 computer (Dec. '81 ETI), called SBCOS.

SBCOS contains the popular and proven BINBUG monitor together with ACOS, Microbytes "... fast and reliable" cassette operating system.

Supplied in a 2532 EPROM, SBCOS contains two versions of BINBUG which support serial and parallel keyboards and serial or memory mapped VDU devices. Serial I/O may be performed at 300, 1200 and 2400 baud.

The cassette operating system utilises a simple interface to reliably record named files at 3000 baud using a conventional audio cassette recorder.

Port C of the on-board PPI is used by SBCOS to establish the Monitor

I/O characteristics as well as for cassette recorder I/O and motor control, and details of the minor modifications required to install SBCOS on the SBC-2650 are supplied in the comprehensive user's guide.

The SBCOS EPROM including a manual with source listing and cassette tapes containing utility programs costs \$75. The pc board for the ACOS interface costs \$5 (45 x 105 mm board). An update 2532 EPROM is available to registered purchasers of ACOS and BINBUG for just \$25.

Enquiries to Microbyte, P.O. Box 274, Belconnen ACT 2616.

Microprocessor applications course

Following on from his popular radio courses on Microprocessor Fundamentals, Dr. David Mee, of the University of New South Wales School of Electrical Engineering, will broadcast over the University's radio station, Radio University, a new course on microprocessor applications, starting early in October.

These lectures will enable the student to use a microprocessor in his own system for measurement, monitoring or control. It will be assumed that students will have a basic knowledge of microprocessors derived from earlier courses or elsewhere.

The course covers applications of microprocessors to a variety of common situations, each exemplifying different aspects of microprocessor system design. The applications will include:

- A complete single chip analogue data acquisition, display and control system for water heating.
- EPROM programming for a single chip microcomputer.
- A raster scan terminal.
- Subsystem for communications and magnetic storage.

There will be eight lectures broadcast over the University's radio station VL2UV at 7 pm Tuesdays, repeated 8 pm Thursdays. Transistor radios adjusted

to pick up VL2UV can be purchased from Radio University for \$10 post paid or \$8 if collected. Students also receive instructions on how to modify their own radio.

One lecture will be transmitted over VITU, the University's television station. There will be two attended seminars held at the University.

The course fee of \$27.50 includes a comprehensive set of notes in addition to the lectures and seminars.

Following broadcast the course will be available on tape, at \$8 per radio lecture on audio cassette and \$30 or \$50, according to format, for the video cassette of the television programme. Three sets of notes are supplied with the tapes, making them ideal for training purposes.

For full particulars on this course, and others offered by the University of New South Wales by radio, television and tape, phone 622 2691 or write to P.O. Box 1, Kensington NSW 2033.

OKI 2350 printer

The OKI 2350 high-speed, high-quality, dot matrix printer has been released in Australia and New Zealand by Anderson Digital Equipment.

The durable nine-pin head boasts a head life of 500 million characters to handle the demands of office computer applications.

Capable of printing in two colours, the OKI 2350 offers standard, two condensed and three expanded fonts. It can also print subscripts and superscripts as well as underline. Any combination of these may be printed on the same line.

A 2K ROM allows users to store a full alternate character set in the unit.

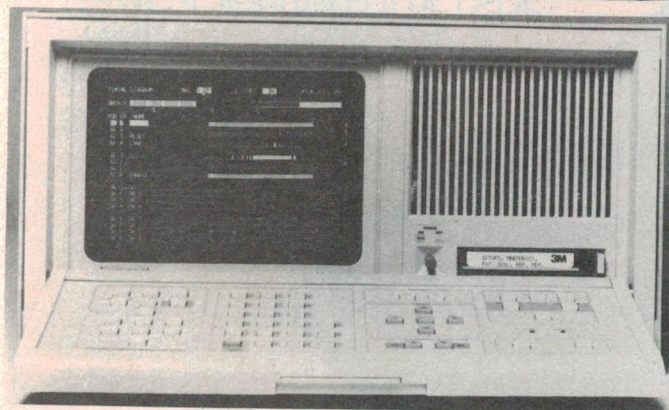
The OKI 2350 is able to print at speeds up to 350 character per second and it employs a bi-directional printing mechanism, short-line seeking logic, and high horizontal and vertical slew speeds for maximum throughput.

A print buffer employing a 2K RAM permits downline loading of an alternate character set from the host computer, so users can change character fonts electronically without changing any print element.

In the graphics mode of the OKI 2350 72 x 72 dot addressable graphics are offered, giving users the capability to generate tables, graphs and charts.

Form handling is provided by horizontal and vertical tabs, top of form, extended paper feed, and a fourteen-channel vertical format unit.

For further information contact Hugh Logie, Anderson Digital Equipment, 14 Whiteside Road, Clayton Vic. (03)544-3444.



Colour logic analyser

Tektronix recently introduced the colour version of its DAS 9100 family of Digital Analysis Systems that features a colour CRT display.

The DAS 9120 Series of colour logic analysers is a modular digital analysis system housing both data acquisition and pattern generation card modules in the same mainframe.

Offered in a variety of data widths to 104 channels and speeds to 660 MHz these modules are combined in the mainframe to match the user's applications needs. The interactive pattern generator allows simultaneous stimulation and acquisition from a device under test.

For data acquisition, three modules are available to meet specific design requirements. The 32-channel module provides 25 MHz sampling and the 8-channel mod-

ule provides 100 MHz sampling, both with 512 bits per channel memory. The 4-channel module provides 330 MHz sampling with 2048 bits per channel memory and a special high resolution mode provides 660 MHz (1.5 ns resolution) on two channels with 4096 bits per channel memory. These modules can be intermixed to support a variety of applications.

For pattern generation, there is a module with 16 data output channels at 25 MHz, expandable up to 80 channels.

For more information on the new DAS 9120 Series contact Tektronix Australia Pty Ltd, 80 Waterloo Rd, Nth Ryde NSW 2113. (02)888-7066.

Printer range

The first two products in a new range of intelligent matrix printers have been released to the Australian market by CASE Communication Systems.

Manufactured for CASE in the UK, the initial range constitutes the BD 136 with 240 cps capability and the WM 2000 with a speed of 120 cps. Both printers offer a bi-directional logic seeking mechanism for maximum throughput speed and to minimise printhead movement.

A 9 x 9 dot matrix provides high quality print and the ballistic printhead has a reliable, long life, according to CASE.

Both units include dot-addressable and character graphics as well as user programmable characters. The DB 136 has an alternative 96 character set, user-definable and capable of printing with a resolution of 60 dots per inch horizontally and 72 per inch vertically.

Large buffers are standard features

of both machines, with 10 000 characters on the BD 136 and a possible 1800 on the WM 2000.

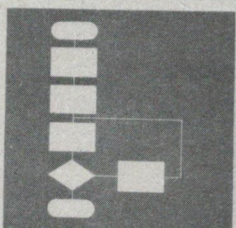
The BD 136's data formatting functions include centre justification, shift to right margin, decimal point and comma alignment, underlining, expanded characters, proportional spacing and right margin justification.

Other features include full forms control and horizontal and vertical tabbing, and the ability to define additional non-standard characters as well as down loading complete character sets.

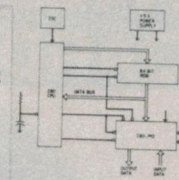
Interfaces available with both models include Centronics parallel, RS 232, Current Loop and IEEE.

For further information contact Barry Foster, CASE Communication Systems. (02)438-2400.

Microprocessors:
Your questions
answered Alec Wood



Newnes Microcomputer Books



Newnes Microcomputer Books

ZX81
BASIC BOOK
Robin Norman



Newnes Microcomputer Books

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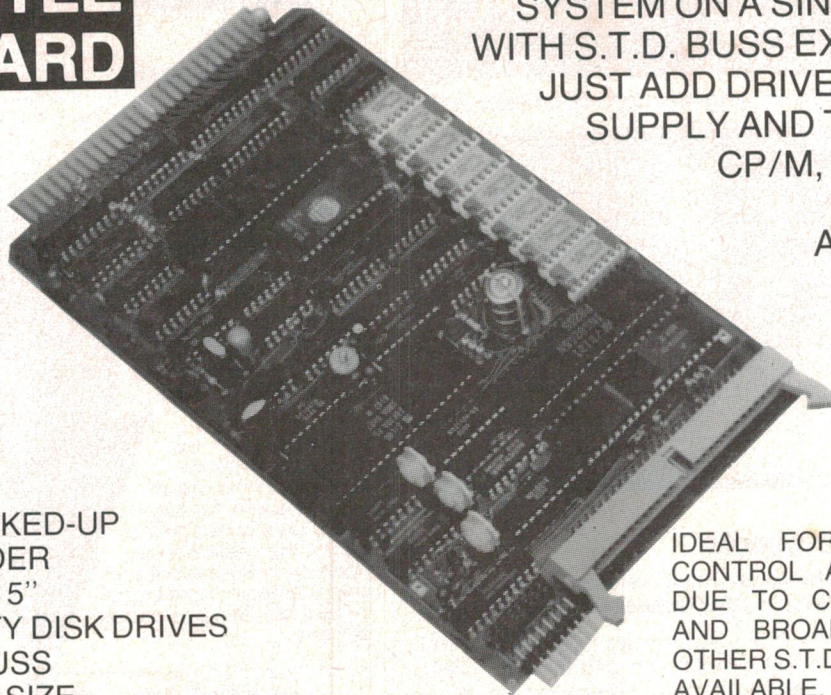
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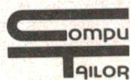
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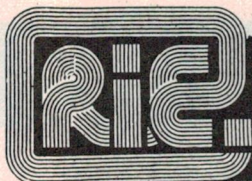


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| 7410 | .18 | .17 | .16B |
| 7421 | .15 | .12 | .11B |
| 74LS00 | .18 | .17 | .13B |
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"Escape from Rungistan" and "Flywars"

A fair indication of the worth of a computer game is the latest time in the wee small hours that you find yourself playing the thing. I call this factor the 'Cohen Rating', and "Escape from Rungistan" (by Sirius Software) has a CR of 3.

"Escape" is one of a family of computer games known as 'adventure games'. The idea is that a series of what the Americans call 'scenarios' are put onto the screen, and you have to find the right words to feed into the computer to get you out of trouble.

For example, the computer might come up with "YOU COME TO A CAVE, OUT OF WHICH RUNS A HUNGRY BEAR". You have to type in an answer to this situation, such as "SING TO BEAR", at which the computer will respond "THE BEAR LIKES YOUR SINGING SO MUCH THAT IT DECIDES NOT TO EAT YOU". Or perhaps, "THE BEAR IS TONE DEAF, AND SO IT EATS YOU — END OF GAME".

(If this example seems rather arbitrary to you, then I have to warn you that adventure games are all written in much this vein — with much humour and practically no logic. It's in trying to guess the mind of the writer that the fun lies.)

"Escape" is a rather special adventure in that it not only describes the scenario to you, it draws a picture of it too, using the Apple's high-resolution graphics. In fact, some of the pictures actually **move** — a snowstorm, for example, with falling snow, or a snake that slides from one side of the screen to the other.

The theme behind "Escape" is

that you start off in a prison cell in the country of Rungistan, and you have to escape from the cell, and then move from scenario to scenario across the face of the country, encountering all sorts of dangers, until you finally reach freedom.

If any of the dangers get the better of you — you die! Then it's back to the start of the adventure (in the cell) and go through it all again. One of the nice features of "Escape", though, is that you can 'save' the state of play at any stage, so that you can return to the spot in the next game, without going through all of the trials and tribulations you had to work your way through the first time.

The game doesn't only use pictures, it plays tunes too — each appropriate to the situation (the tunes can be turned off if you **are** playing at three in the morning, by the way). For example, in the prison cell, the first thing you hear is "Hang Down Your Head Tom Dooley".

"Flywars" is quite a different type of game — it's more like the 'slot machine' variety. That's not to say it's not as addictive (CR of 2.30).

The idea behind "Flywars" is that you are a spider (not one of those nasty poisonous ones — just a plain old huntsman, catching flies). As you crawl across the screen, you leave a web behind.

Flies cannot pass through this

Superpilot extends Apple software

Superpilot, a versatile extension of the Apple Pilot software language, has been introduced by Apple Computer Inc. The Pilot series helps educators and industrial trainers create lessons and illustrations for computer-aided instruction.

The features of Superpilot are graphic enhancement, easy debugging, and external video control.

"Lessons or training sessions created on Superpilot will make difficult concepts and processes easier to understand and retain," states Debra Janssen, product marketing manager. "Powerful computer simulations for industrial or academic learning situations can now be designed without a main-frame computer or its complicated software."

The Superpilot program controls external videodisc and videotape through user and computer command and response and presents

Turtle graphics for easy graphics programming and discovery learning.

Two support products in Apple's Pilot family which have just been released, are Co-Pilot and Superpilot Log. Co-Pilot is a completely self-contained, self-paced interactive tutorial on two diskettes which teaches how to program in Apple Pilot.

Superpilot Log works with Superpilot as an administrative record keeping program that automatically tracks test scores by item, student, or class and can also analyse non-computer test scores entered manually.

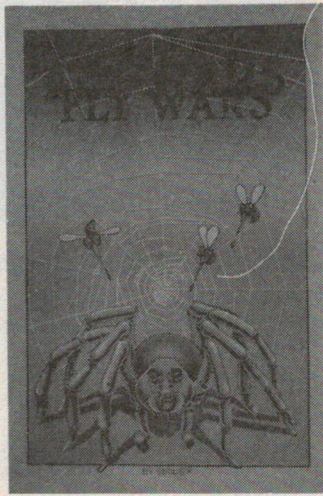
games for the Apple II — review



web, and you cross and recross the screen, trapping them in smaller and smaller areas of space, until finally you corner them and eat them (with a very satisfying 'squelch!'). As you eat more and more flies, your score goes up. And as the score goes up, your job becomes more and more difficult.

For example, a 'fly spray' appears that shoots clouds of chemicals at you — if you're hit by one, you die, and fall to the bottom of the screen.

If you still manage to improve your score while dodging the spray, other nasties appear, making it more and more difficult as you go on. This is a nice change from the usual game plan, where you choose a 'degree of difficulty' at the start of the game.



The "Flywars" programmer has provided an automatically adjusting degree of difficulty.

I never got past the flyspray, by the way — and there are about half a dozen degrees of difficulty past that!

"Flywars" (which comes with a dazzling plastic stick-on display, presumably for the side of your TV) is also by Sirius Software, and both games are distributed in this country by Imagineering, 22 Sir John Young Cr, Woolloomooloo NSW 2011 and available through Apple dealers.

Now if you'll excuse me, I think I'll get back to my computer and play another couple of rounds... after all, it's only two in the morning...

Phil Cohen

64K Dynamic RAM with pin one refresh

Motorola MOS Integrated Circuits Group has announced availability of the second generation 64K dynamic RAM with 'pin one refresh', the MCM6664A, offering improved performance.

The MCM6664A can do either RAS/CAS or RAS-only refresh cycles and has two additional refresh methods available to the user. These special functions are incorporated on pin 1 of the device and have been approved by JEDEC as an alternative function for that pin on the 64K dynamic memory. They are the auto-refresh and self-refresh modes.

The auto-refresh mode is accomplished by simply making pin 1 active during the time interval when a refresh cycle is desired. The refresh address is generated internally and is automatically incremented for the next refresh cycle. During pin 1 active low time, RAS and CAS

are at V_{IL} and all other inputs are 'don't care'.

The second refresh method is intended primarily for battery backup applications where pin 1 will be active longer than 2 μ s. This self-refresh mode generates internal refresh pulses in addition to the internal refresh addresses.

Both the auto-refresh methods simplify memory system designs, save system refresh overhead circuitry and reduce skew times associated with board level refresh address multiplexing.

For more information contact Motorola Semi Conductor Products, 250 Pacific Hwy, Crows Nest NSW 2065. (02)438-1955.

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Amazing space for Sinclair's little wonder

In 1981 Clive Sinclair's ZX81 was hailed as an important breakthrough in the micro world. But this amazing machine's lack of memory power was a big handicap for the serious computer user or advanced programmer.

The RAM pack offered with the ZX81 was very handy but its capacity was limited to 16K. Since the BASIC interpreter can work with 32K of memory, Vendale Pty Ltd are selling a 32K RAM expansion board (RP32) which puts this potential to good use.

Because it is difficult to achieve both performance and low power consumption, the RP32 uses dynamic RAM chips. They offer several advantages over static chips, being considerably denser and cheaper.

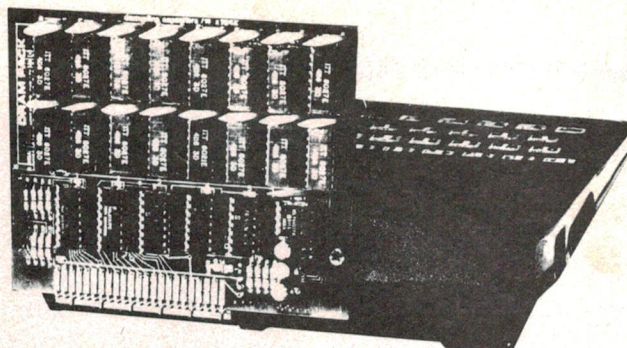
We gave readers a brief introduction to this RAM add-on on page 81 of the August issue, but as we didn't know much about it at the time, could not supply much detail. Shall we fill in the gaps now?

The RP32 is a self-contained, high performance 8-bit memory manufactured with high speed, low power MOS RAM. There are four

sections.

The address decoder uses address lines A15 and A14 and divides the memory space into four blocks of 16K bytes: block 0 is ROM, block 1 is unused, block 2 is the second RAM bank and block 3 is unused; the timing logic section provides refresh and address multiplexing. The refresh is performed at the end of FETCH OP CODE (M1 cycle) and is totally transparent to the user; a -5 V generator uses the Intersil MAXICOMOS ICL7660 to transform the +12 V into -5 V for biasing the DRAM; the DRAM array consists of two banks of 4116s, each bank containing eight chips, giving 32K. No extra power supply is required.

This 32K RAM pack is available from Vendale Pty Ltd, P.O. Box 456, Glen Waverley, Victoria 3150, at a cost of \$165, postage included (Australia).



Tektronix microprocessor selection

Tektronix Australia announces two one-day seminars on how to select the right 8 and 16-bit microprocessor for your application.

The topics discussed will cover fabrication technology, chip-architecture, development tools and other selection criteria.

David Ransier, a visiting Tek US expert on microprocessor development, will be conducting the seminars.

The seminars are intended for decision makers who need to know what the microprocessors can do and to better understand the chip selection process, experienced engineers who need to quickly learn the internals of a micro and how it affects the selection process and

less experienced engineers who are suddenly thrust with the task of developing hardware or software and must make the best decision on which micro to use.

The seminars will be full day events, one held in Sydney on Thursday, October 7 and the other in Melbourne on Tuesday, October 12. A fee of \$25 per person will cover a luncheon and the seminar materials.

For further information contact Sonya Stokell in Sydney by phoning (02)888-7066 or Jill Radford in Melbourne on (03)813-1455.

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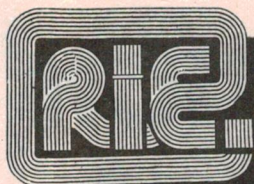
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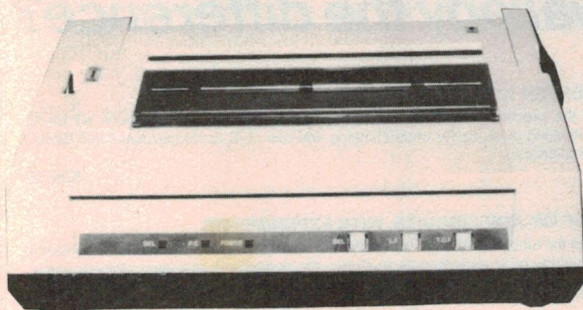
Transmitting Method—Half Duplex. Synchronization—Asynchronous.

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Physical dimensions: 398 mm W x 120 mm H x 285 mm D (15.7" W x 4.7" H x 11.2" D). **Weight:** 8.5 kg (18 lbs., 12 oz.)

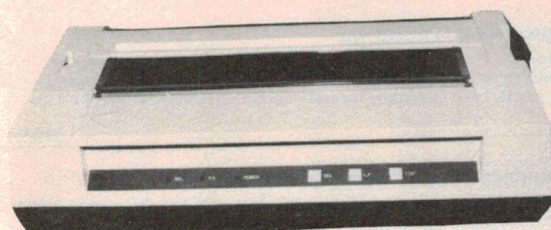


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The Model 1550 is a compact desk-top dot matrix serial impact printer used for data communication terminals, hardcopy of CRT displays, peripheral terminals for minicomputers and microcomputers, and small-sized business systems.

The character format is a dot matrix of 7(H) x 9(V). or 8(H) x 8(V). Print speed is 120 characters/second. Up to 136 characters can be printed per line at 10 CPI.

Its main features are: • Compact desk-top dot matrix printer • 136-column print • Lightweight • Low power-consumption • High-quality print • Bit image graphics • Graphic Symbols • Prints in six different languages • High reliability • Low cost.

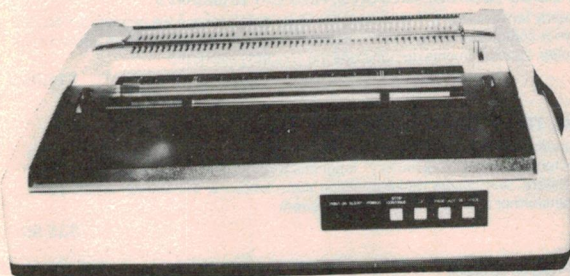


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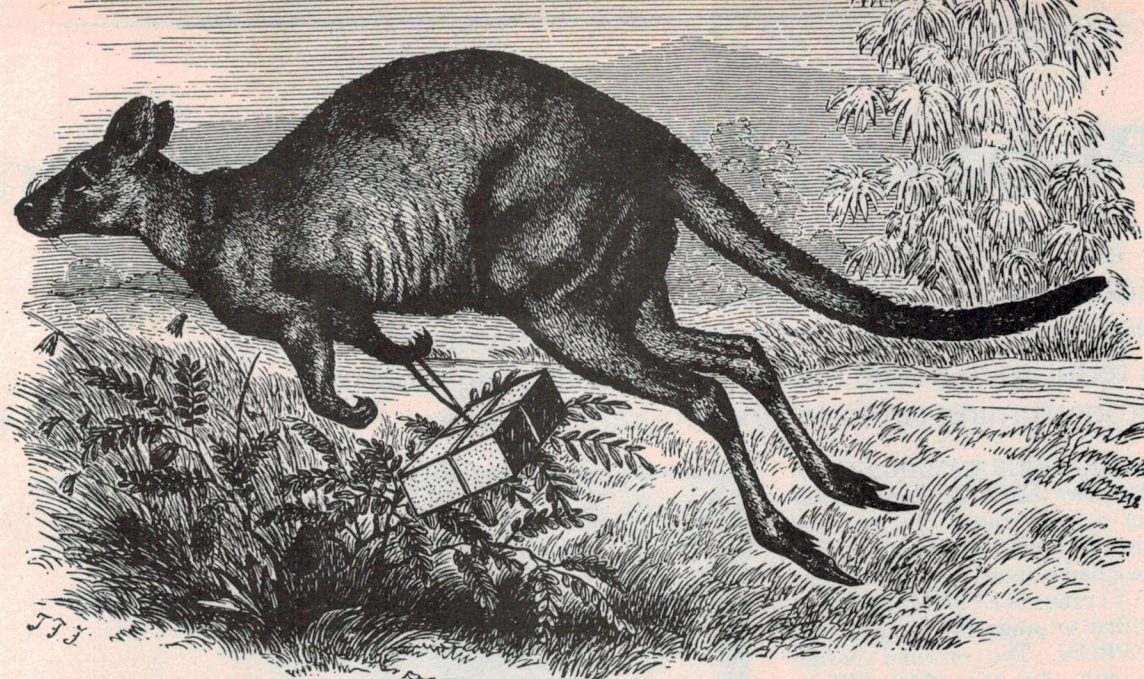
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PPI-based EPROM programmer

Here is a real application for the programmable ports provided on the ETI-685 processor board by the 8255 PPI. The hardware required for this EPROM programmer has been reduced to a bare minimum by transferring most of the programming control to the program software.

Ron Keonig

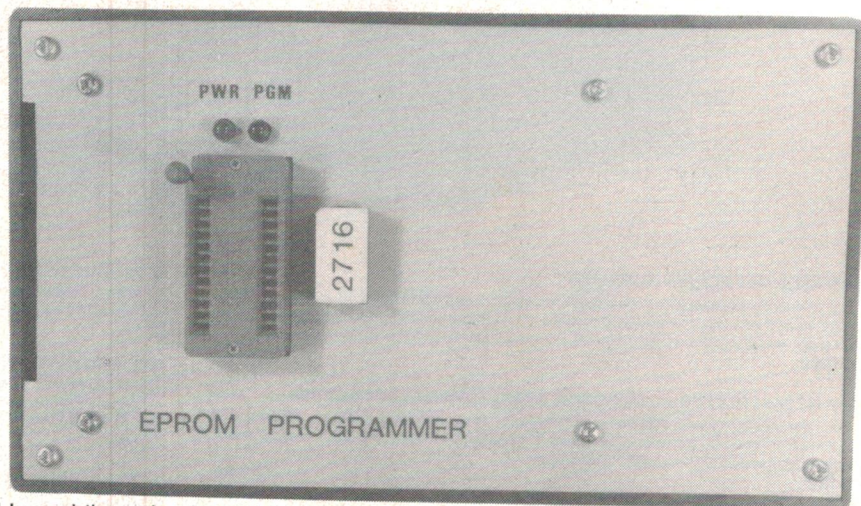
ERASABLE PROGRAMMABLE Read-Only Memories (EPROMs) are available today in a large range of types and capacities. This simple programmer has been designed primarily for the single-rail (+5 V) range of EPROMs but it can be modified to program the three-rail 2708 EPROM. The original design catered only for the 2716 2Kbyte EPROM which is by far the most popular type in use. The current design incorporates a 16-pin DIP 'personality module' to provide a means of rewiring the programmer to cater for all the 24-pin single-rail EPROMs.

The EPROM programmer is completely self-contained and can be used with any microcomputer which has three 8-bit ports available. A full source listing of a 2650 program to read, program and verify the programming of a 2716 EPROM has been supplied in this article. Of course, this program will have to be altered to cater for the other types of EPROMs.

Most computer systems use EPROMs to store the monitor program and to provide the VDU character generator ROM. With this programmer these EPROMs can be easily modified or replaced. EPROMs can also be used to store useful 'utility' programs and for the exchange of large programs from one system to another.

The ETI-685 2650 S100 Processor Board contains the unique feature of having RAM memory overlayed on the EPROM monitor. This feature allows new programs (either new monitors or OEM dedicated software) which are to reside at address '0000'H to be written on an assembler and assembled into the RAM. These programs can then be run and their operation verified before they are committed to EPROM. The following seven-byte program, when executed, will disable the monitor EPROM and transfer the CPU control to the program just assembled and stored in the on-board RAM at '0000'H. (Note: the two 2114 RAMs at IC14 and IC29 must be fitted to the ETI-685 board.)

| | | |
|---------|-----------------|--------------------------------|
| 04 01 | LODI,R0 H'01' | Bit to set EPROM disable latch |
| D4 07 | WRTE,R0 H'07' | I/O address of the latch |
| 1F 0000 | BCTA,UN H'0000' | Branch to Address '0000'H |



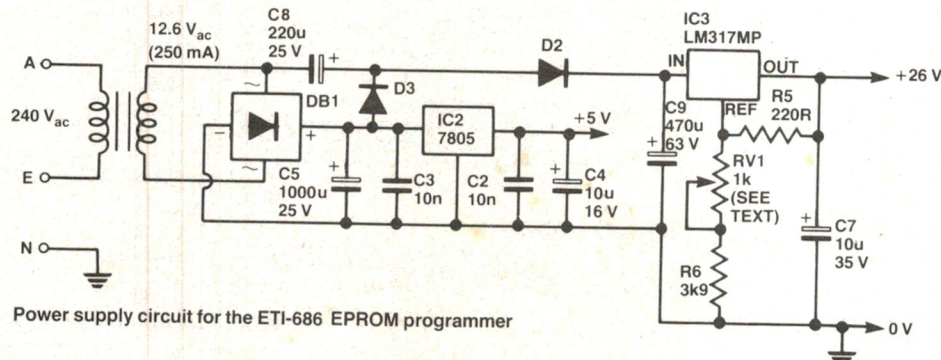
I housed the project in a jiffy box and lettered the front panel with Letraset. The 'ZIF' socket and 'personality' module DIP header are readily identified here.

The principal requirement for an EPROM programmer is to apply a 25 V programming voltage to EPROM pin V_{pp}, supply the appropriate address and eight bits of data to the EPROM socket and then generate the appropriate programming pulse. For the 2716 EPROM the addresses can be output in any order, and for each address a once-only 50 ms logic '1' pulse is applied to the CE input.

This programmer uses the B port and C port (lower) PPI lines to supply the required 11 address lines, and the A port is used to transport the data. The A port is set for output during programming, and for input during reading and verification. Two lines from the C port (upper) are used to switch on the V_{pp} programming voltage and to pulse the CE line during programming.

In most simple programmers the programming pulse duration is generated by the timing of a monostable multivibrator. As the pulse duration needs to be accurate for reliable programming these 'one-shots' must be set up using expensive test-equipment such as a frequency-period meter or a CRO. For this programmer the pulse duration is generated by the microprocessor executing a 'calibrated' delay subroutine. With all CPUs running from a crystal controlled 1 MHz clock the pulse length generated is extremely accurate, and no user timing calibration is required.

Interested software-buffs may like to know that the PPI's single bit set/reset feature has been used to generate the program pulse (see the PULSE subroutine in the program listing). ▶



Power supply circuit for the ETI-686 EPROM programmer



SERIES 5000

As designed by ETI



SERIES 5000 PREAMPLIFIER — SPECIFICATIONS

- Frequency response: High-level input: 15Hz-130 kHz, +0, -1 dB Low-level input — conforms to RIAA equalisation, ± 0.2 dB
- Distortion: 1kHz < 0.003% on all inputs (limit of resolution on measuring equipment due to noise limitation).
- S/N noise: High-level input, master full, with respect to 300 mV input signal at full output (1.2V): >92 dB flat >100 dB A-weighted.
MM input, master full, with respect to full output (1.2V) at 5 mV input, 50 ohm source resistance connected: >86 dB flat >92 dB A-weighted.
MC input, master full, with respect to full output (1.2V) and 200 μ V input signal: >71 dB flat >75 dB A-weighted.



N.B. Picture is only of original heatsink supplied with this project. Our one is tapped from the rear so that no screw heads are visible. New picture next month.

Please note that the "Superb" quality "Heatsink" for the power amp was designed and developed by Rod Irving Electronics and is being supplied to other kit suppliers. This product cost \$1,200 to develop so that your amplifier kit would have a professional finish as well as sound. We also have a new range of rack mounting boxes which will be released soon.

SERIES 5000 POWER AMPLIFIER — SPECIFICATIONS

- Power output: 100W RMS into 8 ohms (± 55 V supply).
- Frequency response: 8 Hz to 20 kHz, +0 -0.4 dB 2.8-Hz to 65 kHz, +0 -3 dB. NOTE: These figures are determined solely by passive filters.
- Input sensitivity: 1V RMS for 100W output.
- Hum: -100dB below full output (flat).
- Noise: -116 dB below full output (flat, 20 kHz bandwidth).
- 2nd harmonic distortion: <0.001% at 1 kHz (0.0007% on prototypes) at 100 W output using a ± 56 V supply rated at 4 A continuous. <0.003% at 10 kHz and 100 W.
- 3rd harmonic distortion: <0.0003% for all frequencies less than 10 kHz and all powers below clipping.
- Total harmonic distortion: Determined by 2nd harmonic distortion (see above).
- Intermodulation distortion: <0.003% at 100 W. (50 Hz and 7 kHz mixed 4:1).
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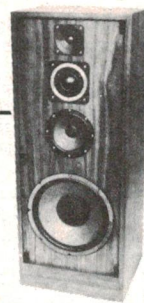
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ETI 10/82/7

PARTS LIST — ETI-686

Resistors all 1/2 W, 5%
 R1, R2 390R
 R3 2k2
 R4 10k
 R5 220R
 R6 3k9
 RN1, 2, 3 3k3 x 9 SIP resistor pack
 RV1 1k min. vert. trimpot

Capacitors
 C1, 2, 3 10n ceramic
 C4 10u/16 V tant.
 C5 1000u/25 V axial electro.
 C6 100n ceramic
 C7 10u/35 V tant.
 C8 220u/25 V axial electro.
 C9 470u/63 V axial electro.

Semiconductors
 D1 1N914, 1N4148
 D2, D3 EM401, 1N4001 etc.
 DB1 BY257 Philips bridge rect.
 IC1 7406
 IC2 7805
 IC3 LM317MP
 LED1 TIL220G green LED
 LED2 TIL220R red LED

Miscellaneous

ETI-686 pc board; three 16-pin solder-in type IC sockets (SK1, 2, 3) — optional; one 24-pin wire-wrap socket (SK4); one 24-pin ZIF socket; 4 x 18 mm spacers; one TO220 heatsink; one 2155 or PL12/20 VA transformer or similar (12 V/1 A); jiffy box, 196 x 113 x 60 mm (UB2 or similar); wire etc.

Price estimate \$42 — \$48

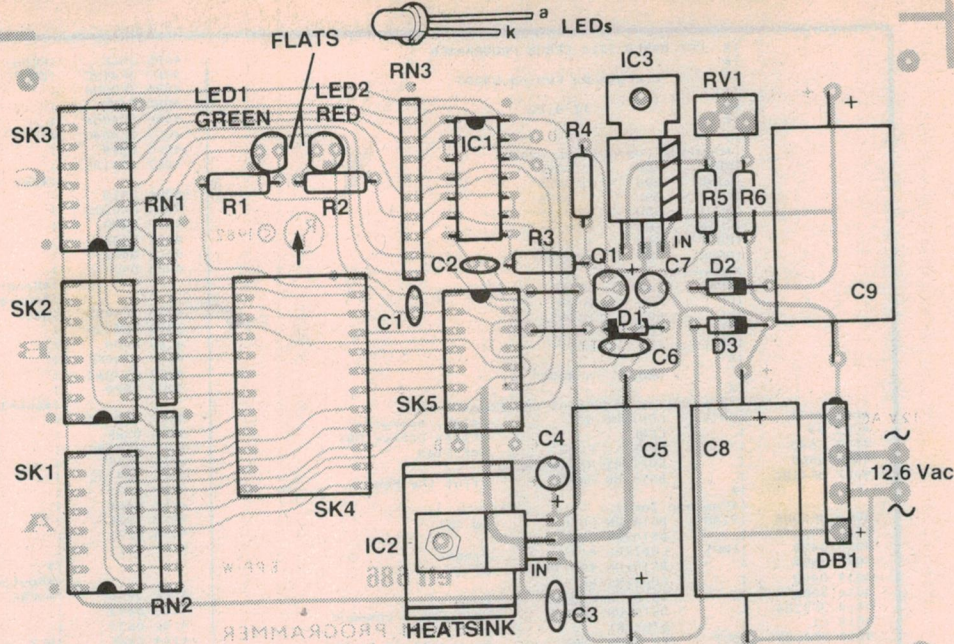
As you can see from the circuit diagram this must be just about the simplest EPROM programmer ever published, requiring only one TTL interface IC and two voltage regulators. But the software supplied makes this programmer a very versatile 2716 EPROM programming package.

Construction

All components for the programmer are mounted on a single sided printed circuit board which measures approximately 140 mm by 90 mm. The maximum height of components on the board is 18 mm. The board may be mounted permanently in the computer or assembled into a suitable case (as illustrated in the accompanying photograph). If you intend to only program 2716 EPROMs then the 'programming module' socket can be replaced with five wire links. If you intend to mount the board behind a panel or box lid you can use a 24 and 16-pin wire-wrap socket to raise the height of the sockets above the other board.

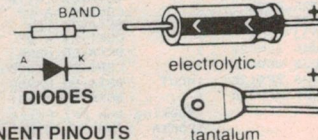
It is important that you use a good quality 24-pin socket for the EPROM. Ideally, a 'zero insertion force' type (ZIF) should be used. If you purchase the Textool brand ZIF you may find that it will not fit into the holes on the board or plug into any type of IC socket. I solved this problem by soldering a 24-pin component header (similar to the 16-pin module header) to the pins of the Textool ZIF.

Mount all the components on the board paying close attention to the polarity of



Component overlay. Test point T1 is on the link adjacent to the cathode of D1. The ● at one end of RN1, 2 and 3 shows the common pin (+5 V).

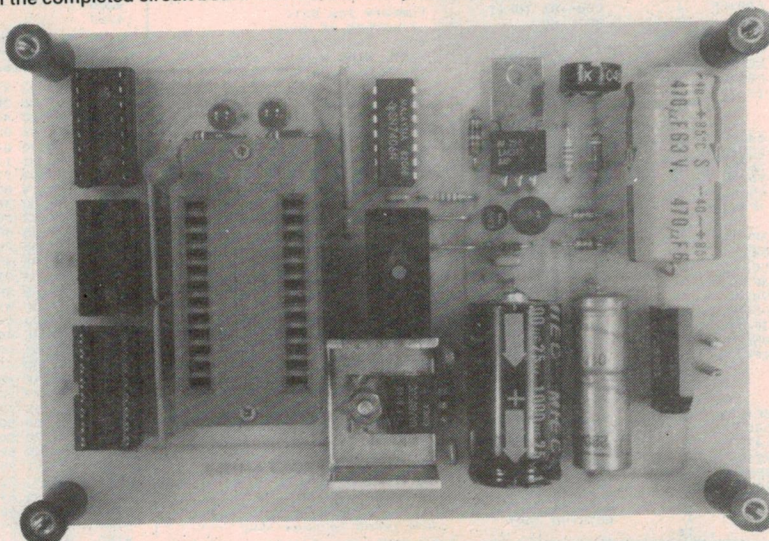
View of the board showing mounting of the wire-wrap and ZIF sockets.



COMPONENT PINOUTS

the electrolytic capacitors and the orientation of the resistor networks. Install the six resistors, three diodes and two wire links first. Next install Q1, then the diode bridge, the resistor networks and the capacitors. The two voltage regulators and the TTL IC are installed next and, after the board has been fitted to the panel, the LEDs can be installed. The three 16-pin IC sockets for the port A, B and C cables can be installed next.

View of the completed circuit board with the standoff pillars installed.



The wire-wrap sockets should be fitted after the assembled board has been secured to the front panel so that you can accurately set their height. A suitable heatsink for the 5 V regulator can be fabricated by bending up two 12 mm sides from a 45 mm x 20 mm x 1.2 mm piece of aluminium. When the board is completed it can be secured to the front panel using four 18 mm to 20 mm long spacers. Plug the zero insertion force socket into the 24-pin wire-wrap socket and fit it to the assembly. Check that the release lever of the ZIF is not resting on the front panel and solder the wire-wrap pins to the board. Next locate the 16-pin module socket to be just proud of the front panel and solder it in place. ▶

PPI BASED 2716 EPROM PROGRAMMER

Execute by keying G4000

11-4-82

Monitor Subroutines used:

```

:MBUG EQU H'0022'
:LINE EQU H'005B'
:CR LF EQU H'008A'
:BDOUT EQU H'0269'
:CHIN EQU H'0286'
:CDOUT EQU H'02B4'
:GNUM EQU H'02DB'

```

```

:Constants
:SPAC EQU H'20'
:CR EQU H'0D'
:LF EQU H'0A'
:BS EQU H'08'
:BELL EQU H'07'

```

ORG H'4000'

Initialise and Print the Command Menu

```

4000 0402 :ENTRY LODI,R0 02      Arith. compare
4002 93 :LPSL          Clear Carry, RS1
4003 7640 :PFSU FLAG      Set Flag
4005 0500 :LODI,R1 MSSG0-MSSG0
4007 3F412C :BSTA,UN PMSGG      Print the Menu

```

Command Input

```

400A 3F008A :START BSTA,UN CRLF      to CRT
400D 3F4040 :BSTA,UN INIT
4010 042B :PMPT LODI,R0 A+'      Prompt
4012 3B89 :BSTR,UN *OUT+1      Cout
4014 0408 :LODI,R0 BS
4016 3B85 :BSTR,UN *OUT+1      Cout
4018 3F0286 :BSTA,UN CHIN      Wait for a command
401B C1 :STRZ,R1
401C 3F02B4 :OUT BSTA,UN COUT      Echo command to CRT
401F E543 :COMI,R1 A'C'
4021 1C408B :BCTA,EQ COPY      Copy EPROM into Buffer
4024 E550 :COMI,R1 A'P'
4026 1C40A2 :BCTA,EQ PROG      Programme EPROM from Buffer
4029 E549 :COMI,R1 A'I'
402B 1C407D :BCTA,EQ INITB      Initialise Buffer with 'FF'
402E E556 :COMI,R1 A'V'
4030 1C410F :BCTA,EQ VER      Verify EPROM with Buffer
4033 E551 :COMI,R1 A'O'
4035 1C0022 :BCTA,EQ MBUG      EXIT to Monitor
4038 05BA :LODI,R1 MSSG8-MSSG0 Print 'INVALID' command
403A 3F412C :MOUT BSTA,UN PMSGG
403D 1F400A :BCTA,UN START

```

Define the PPI Ports (A,B and C) used

```

:PORTA EQU H'00'      for Data (in and out)
:PORTB EQU H'01'      for A0 to A7
:PORTC EQU H'02'      PC0-PC3 for A8-A11
:CNTRL EQU H'03'

```

```

:SETPG EQU H'0B'      Set Programme Pulse (PC5)
:RSTPG EQU H'0A'      Reset Pulse
:SETOE EQU H'09'      Tristate PROM, apply Vpp (PC4)
:PGEN EQU H'10'      Programme Enable flag

```

Routine to Initialise the PPI

```

4040 0490 :INIT LODI,R0 H'90'      A Input, B and C Output
4042 D403 :WRT,R0 CNTRL
4044 20 :EORZ,R0
4045 CC0202 :STRA,R0 PGENF      Clear flag

```

Initialise the Buffer Pointer TEMP

```

4048 0C01FD :INITP LODA,R0 BUFST      to the value stored at BUFST
404B CC0200 :STRA,R0 TEMP      High byte
404E 20 :EORZ,R0
404F CC0201 :STRA,R0 TEMP+1      Low byte
4052 17 :RETC,UN

```

Routine to Write Byte Address to PPI

```

4053 0C0201 :WRITE LODA,R0 TEMP+1      Low byte
4056 D401 :WRT,R0 PORTB
405B 0C0200 :LODA,R0 TEMP      High byte
405D 440F :ANDI,R0 H'0F'      Strip top nibble
405D 6C0202 :IORA,R0 PGENF      add Enable flag
4060 D402 :WRT,R0 PORTC
4062 17 :RETC,UN

```

Routine to Increment TEMP until TOP address

```

4063 0C0200 :INCR LODA,R0 TEMP      High byte
4066 0D0201 :LODA,R1 TEMP+1      Low byte
4069 EC01FE :COMA,R0 TOP      Compare high byte
406C 1A04 :BCTR,LT INCT
406E ED01FF :COMA,R1 TOP+1      Compare low byte
4071 14 :RETC,EQ      Finished if equal
4072 D902 :INCT BIRR,R1 SAVE      Incr. low byte and
4074 D800 :BIRR,R0 SAVE      incr. high byte if R1=0
4076 CC0200 :SAVE STRA,R0 TEMP      Store the new address
4079 CD0201 :STRA,R1 TEMP+1
407C 17 :RETC,UN      Not finished

```

Routine to Initialise the Data Buffer with 'FF' Hex.

```

407D 04FF :INITB LODI,R0 H'FF'
407F CC8200 :STRA,R0 *TEMP      Data byte to buffer
4082 3B5F :BSTR,UN INCR      Next address
4084 9B77 :BCFR,EQ INITB      Loop till end
4086 05BF :LODI,R1 MSSG3-MSSG0
4088 1F403A :BCTA,UN MOUT      Print 'Completed'

```

Routine to Copy EPROM into Buffer

```

408B 3F4053 :COPY BSTA,UN WRITE      Address to PPI
408E 5400 :REDE,R0 PORTA      Address from EPROM
4090 CC8200 :STRA,R0 *TEMP      to buffer
4093 3F4063 :BSTA,UN INCR      Next address
4096 9B73 :BCFR,EQ COPY      Loop till end
4098 05BA :LODI,R1 MSSG2-MSSG0
409A 3F412C :BSTA,UN PMSGG      Print 'Copy'
409D 05BF :LODI,R1 MSSG3-MSSG0
409F 1F403A :BCTA,UN MOUT      'Completed'

```

Routine to Programme EPROM from buffer

```

40A2 3F4053 :PROG BSTA,UN WRITE      Address to PPI
40A5 5400 :REDE,R0 PORTA      EPROM Data
40A7 E4FF :COMI,R0 H'FF'      Error?
40A9 9C40B5 :BCFA,EQ PERR      Error if not 'FF'
40AC 3F4063 :BSTA,UN INCR      Next address
40AF 9B71 :BCFR,EQ PROG      Loop till end
40B1 057E :LODI,R1 MSSG1A-MSSG0 'PROM Erased'
40B3 1B02 :BCTR,UN CONT      and ready to programme

```

```

40B5 05AE :PERR LODI,R1 MSSG1-MSSG0 'PROM not Erased'
40B7 3F412C :CONT BSTA,UN PMSGG
40BA 3F0286 :BSTA,UN CHIN      Command?
40BD E40D :COMI,R0 CR      'CR' to continue
40BF 9C400A :BCFA,EQ START      Abort if not 'CR'
40C2 3F4040 :BSTA,UN INIT      Reset pointers
40C5 0599 :LODI,R1 MSSG4-MSSG0
40C7 3F412C :BSTA,UN PMSGG      Print 'Programming'

```

```

40CA 0410 :Set up PPI for Programming
40CC CC0202 :LODI,R0 PGEN      Set PGEN flag bit
40CF 0480 :STRA,R0 PGENF
40D1 D403 :LODI,R0 H'80'
40D3 0409 :WRT,R0 CNTRL      Port A output mode
40D5 D403 :LODI,R0 SETOE

```

```

40D7 3F4053 :Programme the Buffer contents into the EPROM
40DA 0C8200 :NXTP BSTA,UN WRITE      RUFST (start) and TOP (end)
40DD D400 :LODA,R0 *TEMP      Address to PPI
40DF 3F40FE :WRT,R0 PORTA      Byte to EPROM
40E2 3F4063 :BSTA,UN PULSE      Pulse programme bit
40E5 9B70 :BCFR,EQ NXTP      Next address

```

```

40E7 3F4040 :Completed programming
40EA 05BF :BSTA,UN INIT      Clear Enable, Port A input
40EC 3F412C :LODI,R1 MSSG3-MSSG0
40EF 05A5 :BSTA,UN PMSGG      Print 'Completed'
40F1 3F412C :LODI,R1 MSSG5-MSSG0
40F4 0532 :BSTA,UN PMSGG      'and'
40F6 0600 :LODI,R2 0
40F8 FA7E :BDRR,R2 $      delay before verify
40FA F97C :BDRR,R1 $-2
40FC 1B11 :BCTR,UN VER      Go to Verify Programme

```

```

40FE 040B :Routine to generate a 50ms Programme Pulse
4100 D403 :PULSE LODI,R0 SETPG
4102 0632 :WRT,R0 CNTRL      Set Programme Pulse
4104 046E :LODI,R2 50
4106 F87E :DLY LODI,R0 110
4108 FA7A :BDRR,R0 $      1mS delay
410A 040A :BDRR,R2 DLY      x 50
410C D403 :LODI,R0 RSTPG
410E 17 :WRT,R0 CNTRL      Reset Pulse

```

```

410F 0600 :Routine to Verify EPROM with the buffer
4111 3F4053 :VER LODI,R2 0      Loop of 256
4114 5400 :VIFY BSTA,UN WRITE      Address to PPI
4116 EC8200 :REDE,R0 PORTA      EPROM data byte
4119 9B0C :COMA,R0 *TEMP      Check byte
411B FA77 :BCFR,EQ VFERR      Error if not equal
411D 3F4063 :BDRR,R2 VIFY      Read 256 times
4120 9B6F :BSTA,UN INCR      Next address
4122 05B0 :BCFR,EQ VFY      Loop till TOP
4124 1F403A :LODI,R1 MSSG7-MSSG0
4127 05A9 :BCTA,UN MOUT      Print 'Verified'
4129 1F403A :LODI,R1 MSSG6-MSSG0 or

```

```

412C 0420 :BCTA,UN MOUT      Print 'ERROR'
412E 3F02B4 :PMSGG LODI,R0 SPAC
4131 0D613A :BSTA,UN COUT      Print a space
4134 14 :LODA,R0 MSSG0,R1      fetch character
4135 3F02B4 :RETC,Z      end if a Null (0)
4138 D977 :BSTA,UN COUT      else print character

```

```

413A 0D0A550 :Menu Table
414B 432D403 :MSSG0 DATA CR,LF,A'EPP COMMANDS',CR,LF,LF
4165 502D2050 :DATA A'C- Copy PROM into Buffer',CR,LF
4178 562D2056 :DATA A'P- Programme PROM',CR,LF
4188 492D2049 :DATA A'V- Verify PROM',CR,LF
419E 512D2051 :DATA A'I- Initialise Buffer',CR,LF
41A7 00 :DATA A'Q- Quit',CR,LF

```

```

41AB 50524F4D :MSSG1 DATA A'PROM not Erased',0
41BB 50524F4D :MSSG1A DATA A'PROM Erased',0
41C4 436F7079 :MSSG2 DATA A'Copy',0
41C9 436F6D70 :MSSG3 DATA A'Completed',0
41D3 50726F67 :MSSG4 DATA A'Programms',0
41DF 416E6400 :MSSG5 DATA A'and',0
41E3 4552524F :MSSG6 DATA A'ERROR',BELL,0
41EA 56657269 :MSSG7 DATA A'Verified',BELL,0
41F4 494E5641 :MSSG8 DATA A'INVALID',BELL,0

```

```

41F0 00 :Data Buffer Address Set to 5000-57FF, length 2K
41FD 50 :BUFST DATA H'50'      Sets Buffer Start Address
41FE 57 :TOP DATA H'57'      Sets Buffer High End Add.
41FF FF :DATA H'FF'      Sets Buffer Low End Add.

```

```

4200 :TEMP RES 2      Temporary address
4202 :PGENF RES 1      OE flas register
00 ERRORS DETECTED
EPROM Programmer HEX Listings.
4000 04 02 93 76 40 05 00 3F 41 2C 3F 00 8A 3F 40 40
4010 04 2B 3B 89 04 0B 3B 85 3F 02 86 C1 3F 02 B4 E5
4020 43 1C 40 8B E5 50 1C 40 A2 E5 49 1C 40 7D E5 56
4030 1C 41 0F E5 51 1C 00 22 05 BA 3F 41 2C 1F 40 0A
4040 08 90 D4 03 20 CC 02 02 0C 01 FD CC 02 00 20 CC
4050 02 01 17 0C 02 01 D4 01 0C 02 00 44 0F 6C 02 02
4060 D4 02 17 0C 02 00 0B 02 01 0C 02 01 17 04 FF 01
4070 FF 14 D9 02 D8 00 CC 02 00 CD 02 01 17 04 FF 01
4080 82 00 3B 5F 9B 77 05 8F 1F 40 3A 3F 40 53 54 00
4090 CC 82 00 3F 40 63 98 73 05 BA 3F 41 2C 05 BF 1F
40A0 40 3A 3F 40 53 54 00 E4 FF 9C 40 85 3F 40 63 98
40B0 71 05 7E 1B 02 05 6E 3F 41 2C 3F 02 86 E4 0D 9C
40C0 40 0A 3F 40 05 09 99 3F 41 2C 04 1C 0C 02 02 04
40D0 80 D4 03 04 09 D4 03 3F 40 53 0C 82 00 D4 00 3F
40E0 40 FE 3F 40 63 98 70 3F 40 40 05 BF 3F 41 2C 05
40F0 A5 3F 41 2C 05 32 06 00 FA 7E F9 7C 1B 11 04 0B
4100 D4 03 06 32 04 6E F8 7E FA 7A 04 0A D4 03 17 06
4110 00 3F 40 53 54 00 EC 82 00 9B 0C FA 7F 3F 40 63
4120 9B 6F 05 B0 1F 40 3A 05 A9 1F 40 4A 45 50 50 20
4130 B4 0D 61 3A 14 3F 02 D4 B9 77 0D 3A 04 20 3F 02
4140 43 4F 4D 4D 41 4E 44 53 0D 0A 0A 43 2D 20 43 6F
4150 70 79 20 50 52 4F 4D 20 69 6E 74 6F 20 42 75 6F
4160 66 65 72 0D 0A 50 2D 20 50 72 6F 67 72 61 6D 6D
4170 65 20 50 52 4F 4D 0D 0A 56 2D 20 56 65 72 69 66
4180 79 20 50 52 4F 4D 0D 0A 49 2D 20 49 6E 69 74 69
4190 61 6C 69 73 65 20 42 75 66 65 72 0D 0A 51 2D
41A0 20 51 75 69 74 0D 0A 00 50 52 4F 4D 20 6E 6F 74
41B0 20 45 72 61 73 65 64 00 50 52 4F 4D 20 65 72 61
41C0 73 65 64 00 43 6F 70 79 00 43 6F 6D 70 6C 65 74
41D0 65 64 00 50 72 6F 67 72 61 6D 6D 69 67 00 61
41E0 6E 64 00 45 52 4F 52 07 00 56 65 72 69 66 69
41F0 65 64 07 00 49 4E 56 41 4C 49 44 07 00 50 57 FF

```

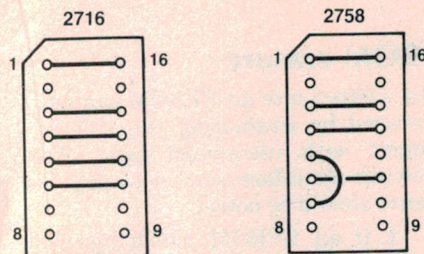



Figure 1 (left). Strapping of the personality module for 2716 programming.

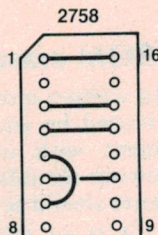


Figure 2 (right). Strapping of the personality module for 2758 programming.

Wiring the programming module plug

The program module required for the 2716 type EPROM is shown in Figure 1. The top and bottom link supply +5 V and the V_{pp} voltage to the EPROM socket. The other links supply the OE, CE/PG and A10 signals. Figure 2 shows the required module for a 2758 where the A10 input has been wired to 0 V.

The inclusion of the programming module has allowed this EPROM programmer to cater for a wide variety of EPROMs. It was not my intention that this project would be the design of a 'universal' EPROM programmer so I will not, at this time, go into the diverse software and module changes required to satisfy the programming requirements of the other types of EPROMs.

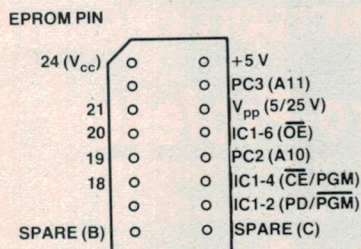


Figure 3. Module signals.

For those interested in experimenting I have included the following additional information. Figure 3 shows the signals available at the programming module and Figure 4 shows the timing of the signals generated by this software package. Most EPROMs require a similar 50 ms 'program pulse', but the timing and polarity of this and the other signals vary.

EPP Software

The program supplied occupies only 512 bytes of memory and uses a 2K block of RAM for a buffer. This buffer may be loaded from tape, disc or from another EPROM, or it may be loaded by the assembly of a source program under the control of an editor-assembler. Using

the Microbyte 2650 Assembler, the command sequence to assemble a source program into the buffer memory at 5000 is to enter 'M.5000' to the prompt 'OPTION?'. The assembler will then assemble the source program using the 'ORG' statement for address calculations, and insert the generated code into the EPROM buffer at H'5000'.

The location and length of the EPROM buffer is defined by three bytes at the end of the program. The start of the buffer is defined by the BUFST byte at 41FD, and the end is defined by the two bytes labelled TOP at 41FE and 41FF. These initially define a 2K buffer which commences at H'5000' and ends at H'57FF'. These locations can be altered if necessary and, to copy an EPROM already installed, they can be set to the address of the resident EPROM.

When the EPP program is run a menu is displayed illustrating the available software commands. The required EPROM programmer (EPP) command is selected by entering the appropriate single letter label. The following is a brief description of the operation of the five available commands in this simple EPP program.

The COPY routine

The Copy EPROM routine is used to load the contents of a pre-programmed EPROM into the buffer memory. As supplied, the routine reads the entire 2K into the buffer and the previous contents are lost. The copy routine can be used when copying EPROMs, disassembling programs or for listing the contents using a HEX lister. The routine is selected by keying in 'C' and, when the copy is complete, the routine will display the message 'Completed' and prompt with a '+' for another command.

The INITIALISE routine

This routine will erase the entire 2K buffer memory and fill it with the value H'FF'. When the routine is finished the message 'Completed' is displayed on the screen and the program will prompt for

a new command. The Initialise routine is mainly used when part-programming is required (see following section) to prepare the buffer. This routine is selected by keying an 'I' and, when finished, the routine will display 'Completed' and return to the command level.

The PROGRAM routine

This routine is actually a dual purpose program and it includes a safeguard to prevent the inadvertent programming of an EPROM which you may have only wanted to copy. When selected with the command 'P' the routine first checks to see if the EPROM is erased. At the end of this test either the message 'PROM Erased' or 'PROM not Erased' will be displayed and the program will pause and await the entry of the 'Return' key. If any other key is entered the program routine is aborted and the prompt for a new command is displayed.

If the Return key is entered the program cycle commences. The message 'Programming' is appended to the previous message and, on the EPROM programmer board, the red LED is illuminated. The entire contents of the buffer memory (as defined by the start and end address) is now programmed into the EPROM.

At the end of the programming cycle the LED is extinguished and the message 'and' is appended to the previous message. The routine now enters the verify cycle where the contents of the EPROM at every address is read and compared 256 times with the appropriate contents of the buffer. This level of checking has been chosen for the detection of poorly programmed EPROMs. The loop count can be changed by altering the value of the byte at H'4110'. If an error is located the message 'ERROR' is displayed and the routine returns to the command level. A reprogram cycle can be introduced by reselecting the program cycle.

If the programming has been successful the message 'Verified' is displayed and the program returns to the command level.

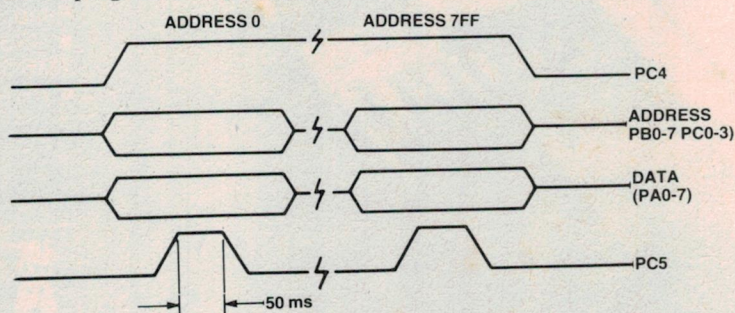


Figure 4. 2716 program cycle.

Project 686

The VERIFY routine

The verify routine used at the end of the program routine can be selected separately with the command 'V'. This routine will either display the message 'Verified' or 'ERROR' and return to the command level. The verify program is particularly useful in comparing the contents of programmed EPROMs to determine if they are at the same revision level.

The QUIT routine

The input of the command 'Q' will cause the computer to exit from the EPROM programmer and return to the system monitor at H'0022'. The EPP program can be re-entered at any time with a G4000 command and, in both instances, the contents of the buffer memory will not be altered.

Programming example

The following is a typical display generated during the copying of a programmed EPROM. The first verify sequence has been selected to check that the correct data has been read into the buffer memory. As the COPY routine only reads each address once I have found that poor contact in the EPROM

socket, due to bent or dirty pins, has caused occasional reading errors, but the verify has found them every time. Better safe than sorry. The second verify, after programming, is actually quite unnecessary.

EPP COMMANDS

C — Copy PROM into buffer
P — Program PROM
V — Verify PROM
I — Initialise buffer
Q — Quit

Part programming

Erased EPROMs contain the data byte 'FF' in every location and only logic 0s are actually programmed into an EPROM. Any attempt to program an 'FF' into an already programmed EPROM will not alter the contents. It is therefore possible to add to or alter the contents of a programmed EPROM, or to part-program an erased one by filling the buffer with 'FF' in the addresses where programming is not required.

The Initialise routine can be used to completely fill the buffer with 'FF' before the part-program is loaded into it. Of course, if the EPROM and the buffer contain different data, the Verify routine will report the message 'ERROR'.


EPROM erasure

Data written into an EPROM can only be erased by irradiating the memory element with ultraviolet light. The following manufacturers' warnings on erasure should be noted.

1. If an EPROM which has not been properly erased is programmed and used, writing problems and operating problems are likely to arise.
2. Excessively long erasure times (of several hours duration) can also result in failure.
3. Lengthy exposure to direct sunlight can result in programmed bit changes. Although normal fluorescent lights have practically no effect, it is recommended that the glass face be covered with a screening label.

With these warnings in mind, you should carefully read the operating instructions supplied with the EPROM eraser which you are using.

With all the information provided in this article you should now find programming your own EPROMs quick and efficient, and you will soon discover the advantages of having your very own computerised EPROM programmer. ●



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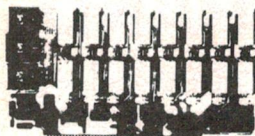
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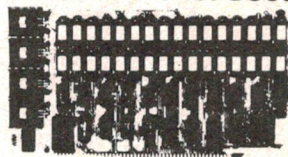
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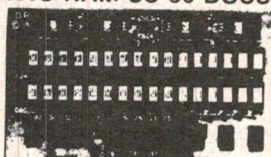
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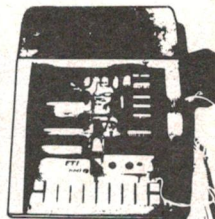


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Part 2

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NOW THAT you have your speech synthesiser up and talking with a set of switches, as described in the last article, you'll want to know how to connect it to your computer firstly, and then how to program it. We can only describe interfacing in a general way as computer I/O ports differ from machine to machine.

The MM54104 speech processor requires a number of control signals as well as data representing the word to be said. The ETI-647 board has circuitry allowing interfacing to the majority of computer parallel ports. There are two completely different ways that you can connect to the board. You simply connect to different ports on the board and include or exclude ICs depending on which way suits you. The two methods are called 'direct control' and 'control data'.

Direct control

Most parallel ports on computers supply sufficient data, address and control signals to the user to drive the Turtle Talk Speech Synthesiser directly. If your computer has the following signals supplied to pins of the user port or expansion port then you will be able to connect the board without any requirements other than a short length of cable and a soldering iron.

- (i) eight data bits
- (ii) five address bits
- (iii) read/write line
- (iv) device enable, port enable, in/out request lines or more address bits.

Connection is via J2 and J3 located either side of IC16's position on the board. ICs 18, 17, 16 and 15 are not used and should be removed from their sockets if you have them in place. Table 2 lists the signals and functions of each line in J2.

These signals from the data buss of your computer will dictate what word is

to be said (no pun intended). They are unidirectional signals going to the speech synthesiser board so computer ports having separate in and out data busses will only use the out data signals.

Tables 3 and 4 show the functions and operations of each line in J3.

The *speak* signal comes from the write signal from your computer and must be active low. If you have a read/write or a \overline{WR} signal then this can be used. You have to invert the signal from your computer if the signal is read/write or \overline{WR} or \overline{RD} .

The *busy* (or 'talking') signal on line 1 of J3 is used for handshaking between the speech synthesiser board and the

computer. Since any speech machine will obviously be operating in real time and a computer could run through a thousand word addresses in a fraction of a second, it is necessary to slow the computer up. i.e. "Don't interrupt while I'm still talking". If you have an interrupt facility and want to use it then connect this busy signal to your computer's interrupt line and use a wait routine in your software.

Alternatively, you might have separate in-data and out-data signals available on your computer parallel port. Use the in-data 0 pin so that simpler programming can do the handshaking. If neither of the above options are suitable

| J2 (line no.) | FUNCTION (word add. bit) | SIGNAL (from computer) |
|------------------|-----------------------------|---------------------------|
| 1 | SW8 | D7 |
| 2 | SW7 | D6 |
| 3 | SW6 | D5 |
| 4 | SW5 | D4 |
| 5 | SW4 | D3 |
| 6 | SW3 | D2 |
| 7 | SW2 | D1 |
| 8 | SW1 | D0 |

Table 2. The functions of the J2 interface lines and the signals required. SW1-8 refer to the word address lines in Table 1 (Part 1, Master Word List). As the table shows, these can be hooked directly to your computer's data buss.

| J3 (line no.) | FUNCTION performed | SIGNAL (from computer) |
|------------------|-----------------------|---------------------------|
| 1 | Busy (talking) | Data in or Int. |
| 2 | Board select | A6 |
| 3 | Command select | A5 |
| 4 | Speak | read/write |
| 5 | ROM select | A3 |
| 6 | C ROM | A2 |
| 7 | B ROM | A1 |
| 8 | Mute | A0 |

Table 3. The functions of the J3 interface lines and the signals produced or required. The ROM B, C and select lines determine which of the four word lists is selected. Each word list is contained in two ROMs and up to eight ROMs, making a total of four word lists, can be accommodated on-board. Table 4 shows the logic for word list selection.

| ROM Select (J3/5) | C ROM (J3/6) | B ROM (J3/7) | List selected |
|-------------------|--------------|--------------|---------------|
| H | X | X | Clear |
| L | L | L | List 1 |
| L | H | L | List 2 |
| L | L | H | List 3 |
| L | H | H | List 4 |

Table 4. The logic for selecting the word list ROMs. Two bits (J3, lines 6 and 7) are used to select the appropriate ROM pair while the ROM select line (J3, line 5) is held low. When this line is held high, it doesn't matter what signals appear on lines 6 and 7 (hence the X in the first row of Band C ROM columns). The ROMs are then readied for the list to be selected.

speech synthesiser

then don't connect anything to J3/1. It is still extremely easy to generate delays in the programming without handshaking. See the 'programming' section for these details.

Control data

This method will allow a bidirectional data buss to be used to control all of the speech synthesiser functions. The data signals are routed through unidirectional buffers to respective points by selecting different addresses. Connection this time is through J1 and ICs 18, 17, 16 and 15 must be plugged into their respective sockets. Table 5 lists the signals required by, and functions of, each line in J1.

Most computers will be connected this way since interfacing and programming are easy for the beginner and most ports on hobby computers are bidirectional.

An 8-bit code is presented at J1/1 to J1/8 to select the word required. J1/9-J1/11 used to select the correct address for this data to be stored by IC16. Next, another code, selected according to all the control signals required (as for J3), is presented to J1 and the correct address for this data is determined by J1/9-J1/11 so that it is latched by IC15. Speech is initiated when a third code, also a control code, is presented and latched in such a way that the \overline{WR} line of the speech processor (pin 4) is toggled. i.e: taken low then high.

The 8-bit data bytes are directed to various destinations by the control

| AO | R/W | \overline{DE} | |
|----|-----|-----------------|-----------------------|
| X | X | H | board not selected |
| L | L | L | control byte (data 2) |
| H | L | L | word byte (data 1) |
| L | H | L | busy byte (data 3) |
| H | H | L | not used |

Table 6. Control signal logic on J1. (Note that X = 'don't care'.)

signals A0, R/W and \overline{DE} according to Table 6.

The control byte (data 2) is the set of functions normally selected through J3, and the word byte (data 1) is used to select the word to be spoken. The busy byte (data 3) uses a single bit for the handshaking between the speech board and computer. All of these commands are introduced above in the Direct Control section and will be covered thoroughly in the programming section. The terms data 1, data 2 and data 3 for each of the three bytes are from the way they are used when programming. For example: POKE W+1,X will select the

| J1 (line no.) | FUNCTION | SIGNAL (from computer) |
|------------------|---------------|---------------------------------|
| 1 | data I/O 0 | D0 |
| 2 | data I/O 1 | D1 |
| 3 | data I/O 2 | D2 |
| 4 | data I/O 3 | D3 |
| 5 | data I/O 4 | D4 |
| 6 | data I/O 5 | D5 |
| 7 | data I/O 6 | D6 |
| 8 | data I/O 7 | D7 |
| 9 | address 0 | A0 |
| 10 | read/write | R/W |
| 11 | device enable | \overline{DE} , I/O sel. etc. |

Table 5. Signals and functions for J1. This is the 'general purpose' interface port for the ETI-647 Turtle Talk speech synthesiser.

word with code number X from the vocabulary list when the speech synthesiser board is located at address W. i.e: POKE W + 1,0 will cause it to say "This is Digitalker". Or, POKE W + 1,128 will have it say "right".

Programming

Even the most inexperienced programmer will have few problems getting the ETI-647 Turtle Talk Speech Synthesiser to talk. All programming depends on simple POKE and PEEK or OUT and INP instructions. To say a word simply requires the programmer to know the code number for the word and which vocabulary list it is on (which ROM pair). This requires only three instructions, all of which can be placed on the one line:

POKE W, A: POKE W + 1, 8;

POKE W + 1, 16 + (2 * B)

Where A is the code number for the word and B is the number of the ROMs the word is in. B will be 0, 1, 2 or 3 and the standard vocabulary list is ROM pair number 0. Hence, to say the word "GREAT" requires:

POKE W, 88: POKE W + 1, 8;

POKE W + 1, 16

(W is the address of the Turtle Talk board.)

To prevent the Turtle Talk from trying to say words on top of each other is also a simple programming task:

100 IF PEEK (W + 1) < 255

THEN 100

This program line will cause the program to stay on the line until the word is finished being said.

The Turtle Talk board offers many other options which are discussed fully later, but they are just as easy to use, making the synthesiser very versatile indeed. The above examples provide the

means to get the board up and talking immediately.

Command modes

The ETI-647 board utilises the CMS (Command Mode Select) facility of the Digitalker speech processor. This allows the programmer to select one of two modes:

- Reset the speech and initiate a new word,
- Reset the speech only.

The interconnection of different word parts becomes possible by using the start of the word only.

Interrupt

Similarly, the interrupt facility is available to the programmer so that the next word can follow closely after the finish of a word. A programmer can also use delay loops to achieve the required timing for spacing between words.

Mute

The ETI-647 is the only speech synthesiser in the world, so far as we know, allowing the programmer access to the end or middle parts of words generated by the Digitalker speech processor. The versatility available with this facility makes the Turtle Talk possibly the most powerful speech generation system available to date.

Using the mute facility, words other than those in the ROM vocabulary can be 'constructed' from the existing vocabulary words.

Direct control programming

How you program the speech synthesiser depends somewhat on how you've interfaced it to your computer. Direct Control interfacing provides you with a range of options. The 'full' system, detailed in Figure 7, requires eight data bits at J2 — called data 1, which is your word byte or word data — plus seven address inputs at J3 (data 2, the SPC control data) and an output — data 3, an interrupt signal ('busy'). This is the 'all singing, all dancing' way to go about it. There is a gut-simple way to do it, because:

- The MUTE function is optional
- The INTERRUPT function is optional
- The COMMAND select function is optional
- If only the Master Word List is used, the two ROM address select

★ **STOP PRESS!** — see page 51 for kit availability ★

lines and the ROM select line (A1/A2/A3, J3 5/6/7) are unused.

Thus, your minimal system will look like Figure 8. To program a system like Figure 8 you simply POKE the desired word number at the board address, as follows:

POKE W,A

where W = board (I/O) address

A = word code number

(from Table 1)

To string words together, you program a word, then a wait loop, then the next word, etc. It's sort of cumbersome, but for short 'messages', it's fine and fulfils all the requirements of the KISS theory of machine design — '...keep it simple, Sam'.

To program a system implemented as in Figure 7 requires a bit more thought, but gives much more power. Firstly, your speech synthesiser board has two WRITE addresses and a READ address. Table 7 shows how the various functions are related to the write and read address forms.

| ADDRESS | DATA | FUNCTION | R/W |
|---------|--------|-------------|-------|
| W | data 1 | word select | write |
| W + 1 | data 2 | control SPC | write |
| R | data 3 | interrupt | read |

Table 7. Relation between the three data groups, their functions and read/write addresses.

The role of data 1 has already been mentioned. Data 2 is used to control the speech synthesiser chip (SPC). The relationship between the data 2 lines in J3, their functions, values for addressing and states are illustrated in Table 8. Following is how to use the instructions (W is the address of the board):

To initiate speech

You set the ROM select and write lines on J3 high. Like this:

POKE W + 1, 8

POKE W + 1, 16

All other lines will be low and thus, word list 1 is selected and the mute is off.

To select word list 2 when initiating speech:

POKE W + 1, 8

POKE W + 1, 18

This sets your ROM select, B ROM and write lines.

Initiate speech, mute on

You set the ROM select, write and mute select lines high:

POKE W + 1, 8

POKE W + 1, 17

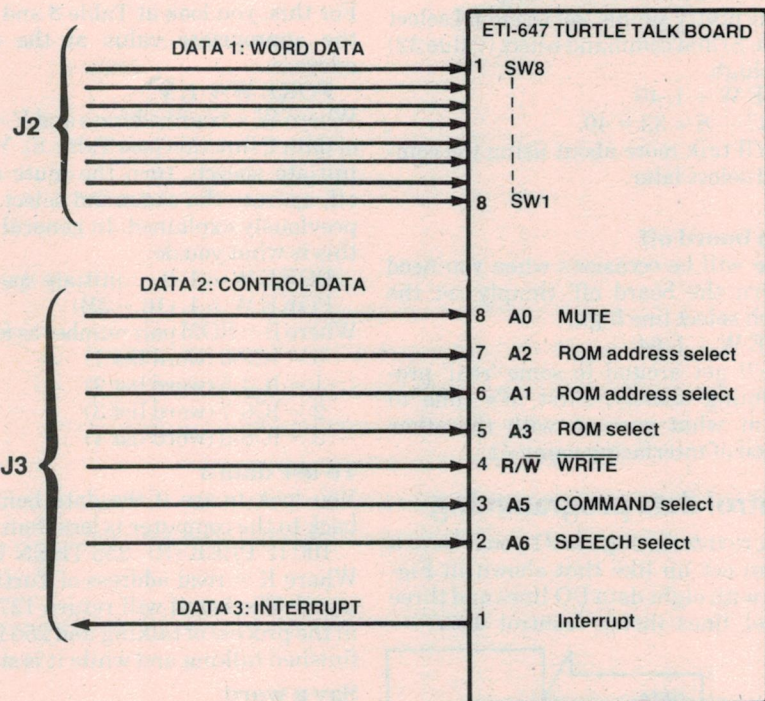


Figure 7. The synthesiser interfaced via J2 and J3, showing the role of each line.

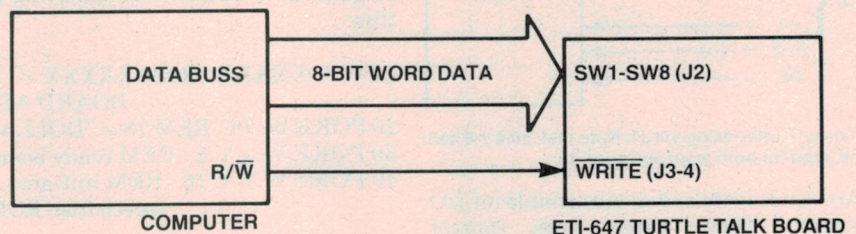


Figure 8. A 'minimal' system.

| DATA 2 FUNCTION | DATA VALUE | STATE | ACTION |
|--------------------|---------------|-------------|------------------------|
| Mute select | 1 | 0 | speech on |
| | | 1 | speech muted |
| B ROM address | 2 | see table 4 | |
| C ROM address | 4 | see table 4 | |
| ROM select | 8 | 0 | set |
| | | 1 | ready to select |
| Write | 16 | 0 | write reset |
| | | 1 | write set |
| Command select | 32 | 0 | reset and start speech |
| | | 1 | reset only |
| Speech select | 64 | 0 | speech on |
| | | 1 | speech off |

Table 8. Using data 2. Note the following:

(a) The ROM select bit must be ready to select then set (see Table 4).

(b) The Write bit must be reset then set every time.

(c) The Speech select bit must be low during operation of the board.

(d) ROM address, Command select and Mute are able to be used whenever needed during, after or before speech generation.

Command select, reset only

This is pretty simple, set the ROM select (value 8) and command select (value 32) lines high:

POKE W + 1, 40

Get it? — $8 + 32 = 40$.

We'll talk more about using the command select later.

Turn board off

There will be occasions when you need to turn the board off. Simply set the speech select line high:

POKE W + 1, 64

We'll get around to some 'real' programming shortly. First, it's time to look at what goes on with the other method of interfacing using J1.

Control data programming

If you're interfacing via J1 you'll have a system set up like that shown in Figure 9 with eight data I/O lines and three control lines (hence 'control data') —

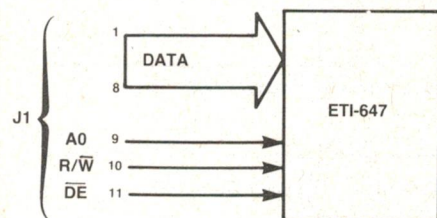


Figure 9. Interfacing via J1. Note that lines 1-8 can be used for both input and output.

A0, read/write and device enable (or I/O select). These three lines control whether you're sending data to the Turtle Talk board — data 1 (word code) or data 2 (SPC control), or reading data from the board — data 3, the interrupt or busy signal. Table 6 shows the logic. Note the following:

(i) It is possible to tie the DE line low so that the Turtle Talk board is on call at all times.

(ii) If the interrupt facility is not needed then the read/write line can be tied low setting the board such that it is always ready to accept data.

(iii) The A0 line can be any line from your computer, or even a switch, that can toggle J1/9 high or low as required. Programming requirements are simple.

To write to data 1

Your simply POKE the appropriate word code number at the board's address:

POKE W, A

Where W = board address and A = word code number from the appropriate word list. To select the appropriate word list (ROM pair), you write to data 2.

To write to data 2

For this, you look at Table 8 and POKE the appropriate value at the data 2 address:

POKE W + 1, V

Where W = board address and V = value of data 2 function (see Table 8). You can initiate speech, turn the mute on and off, operate the command select etc, as previously explained. In general terms, this is what you do:

POKE W + 1, 8 . . . initiate speech

POKE W + 1, (16 + 2B)

Where B = ROM pair number as follows:

0 = IC2-3 (word list 1)

1 = IC4-5 (word list 2)

2 = IC6-7 (word list 3)

3 = IC8-9 (word list 4)

To use data 3

You look to see if the data being sent back to the computer is less than 255:

100 IF PEEK (R) < 255 THEN 100

Where R = read address of Turtle Talk board. The board will return 127 if it is in the process of talking and 255 if it has finished talking and while it is silent.

Say a word

A complete program to cause a word to be generated would look something like this:

```

10 W = XXXXX : REM XXXXX = BOARD ADD.
20 POKE W, 78 : REM 78 = "DOLLAR"
30 POKE W + 1, 8 : REM ready board
40 POKE W + 1, 16 : REM initiates speech from ROM 0
  
```

Command modes

The ETI-647 Turtle Talk board provides the programmer with the facility to use the CMS (Command Mode Select) function of the Digitalker system. The facility provides the ability to reset the speech processor, hence stopping whatever speech was occurring. The programmer can then either start a new utterance immediately, thereby generating a new word by compounding words and parts of words or just cease further utterance, generating a new word from part of an existing word. e.g:

```

COMMA
  POUND
= COMPOUND
or
AMPERE
= AMP
  
```

When a second word is programmed after using the command selection, a new word is formed by the 'phoneme reconstruction' method or by using the first word part as a prefix. For example:

```

HAVE
VOLT
  SS
  PETER
= HOLSTER
  
```

When using the Command Mode Selection function the ROM select line must be held high. In BASIC, the instructions for CMS on and CMS off are like this:

For convenience, if CMS is on, we say that the command mode is on.

To page 88

(a) with CMS off

```

POKE W, 74          — "COMMA"
POKE W + 1, 8       — CMS off, initiate speech
POKE W + 1, 16      — ROM pair 0
FOR T = 1 TO TT: NEXT T — adjust TT for "COMM"
POKE W, 123         — "POUND"
POKE W + 1, 8       — CMS off, initiate speech
POKE W + 1, 16      — ROM pair 0
  
```

(b) with CMS on

```

POKE W, 74          — "COMMA"
POKE W + 1, 8       — CMS off, initiate speech
POKE W + 1, 16      — ROM pair 0
FOR T = 1 TO TT: NEXT T — adjust TT for "COMM"
POKE W + 1, 40      — CMS on, ROM select on
POKE W + 1, 123     — "POUND"
  
```


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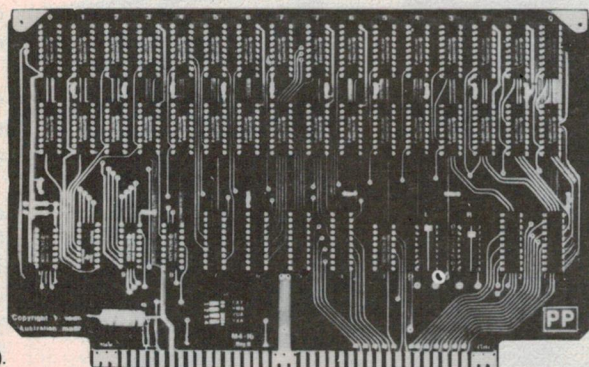
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speech synthesiser

Mute

Using the mute you can select parts of words other than at the start of a word. The Command Mode selection allows the programmer to determine what length of the start of a word is spoken. When this is used in conjunction with the mute control, any part of a word, from someplace in the word to the end of the word, can be selected. There is no other system available for general use with these features associated with set vocabularies.

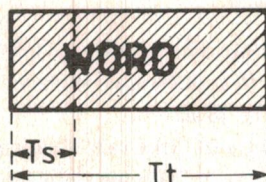
With synthesisers employing phoneme reconstruction techniques, the programmer soon finds that a wide knowledge of linguistics is required before any progress can be made. Phonemes, by definition, are part of words which are perceived as discreet units.

Quite different sounds are often perceived as being identical when spoken within a word and therefore constitute a single perception called a phoneme. This can be demonstrated quite dramatically by editing spoken words from tape recordings to isolate parts of the words.

To construct a word from another word using the mute control you need know nothing about phonemes, linguistics or articulation. Simply select the particular sound by vocalising the word you want, then look for a word in the

vocabulary having that sound.

The method of actually constructing the word is then quite simple. Suppose a word is represented by a rectangle:



The speech synthesiser will take time, T_t , to say the word. To mute a period, T_s , at the start of the word, first generate a time delay with a simple FOR-NEXT loop:

```
FOR T = 1 TO TT: NEXT T
```

The variable TT is adjusted until T_s is of the right length. The mute is operated by setting the Mute Select function in data 2 high, like this:

```
POKE W + 1, 1
```

After the required time you need to turn the mute off again, like this:

```
POKE W + 1, 0
```

Now, the normal instructions for initiating a word are:

```
POKE W, A: REM A=WORD NO.
```

```
POKE W + 1, 8
```

```
POKE W + 1, 16 + (2*B):
```

```
REM B=ROM PAIR
```

With the mute control you can initiate a word but keep it silent like this:

```
POKE W, A
```

```
POKE W + 1, 8
```

```
POKE W + 1, 16 + (2*B) + 1: REM +1  
SETS MUTE
```

The mute can be on for whatever time is required by the use of a FOR-NEXT loop. To remove the mute during the processing of a word you simply write the last line again, less the + 1. Hence, a complete program would look like this:

```
POKE W, A
```

```
POKE W + 1, 8
```

```
POKE W + 1, 17
```

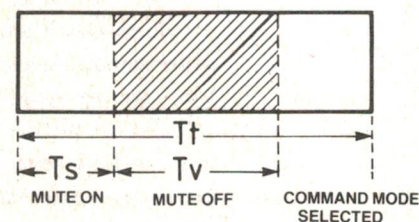
```
FOR T = 1 TO TT: NEXT T
```

```
POKE W + 1, 16
```

For $A = 0$ and $TT = 500$, the above program will isolate the word "talker" from "This is Digitalker" in the master word list (ROM pair 0).

When the Command select is turned on, the speech processor is reset, thus ending the word sharply. This facility can be used with the mute control to isolate the central part of a word out of the total time normally taken to speak the word, call this time T_t (as before),

you need to vocalise only a central portion, call this T_v , muting the first part, call this T_s (as before), and cutting the end portion with the Command select (CMS). If the word is again represented by a rectangle, the sequence of events becomes clear from this diagram:



Where T_t = total time of original word

T_s = period at start

T_v = period of vocalisation

A program to perform this operation looks like this:

```
M = 1
```

```
POKE W, A
```

```
POKE W + 1, 8
```

```
POKE W + 1, 16 + (2*B) + M
```

```
FOR T = 1 TO T1: NEXT T1:
```

```
REM T1 SELECTS Ts
```

```
M = 0: REM MUTE OFF
```

```
POKE W + 1, 16 + (2*B) + M
```

```
FOR T = 1 TO T2: NEXT T2:
```

```
REM T2 SELECTS Tv
```

```
POKE W + 1, 40: REM COMMAND  
MODE SELECTED
```

If a word is required to follow the selected part of the word in this program, the Command select is left off. Like this:

```
M = 1
```

```
POKE W, A1: REM A1 = CODE
```

```
FOR 1ST WORD
```

```
POKE W + 1, 8
```

```
POKE W + 1, 16 + (2*B1) + M:
```

```
REM B1 IS CODE FOR 1ST WORD'S  
ROM PAIR
```

```
FOR T = 1 TO T1: NEXT T
```

```
M = 0
```

```
POKE W + 1, 16 + (2*B1) + M
```

```
FOR T = 1 TO T2: NEXT T
```

```
POKE W, A2: REM A2 = CODE FOR  
2ND WORD
```

```
POKE W + 1, 8
```

```
POKE W + 1, 16 + (2*B2) + M:
```

```
REM B2 IS CODE FOR 2ND WORD'S  
ROM PAIR
```

With a little practice and imagination, you'll soon have your ETI-647 Turtle Talk Synthesiser saying all manner of things! (At this point, we would like readers to know that we take no responsibility for what any constructor's project might say... — Ed.)

ROM EXPANSION

The ETI-647 Turtle Talk Speech Synthesiser employs a method of speech generation which utilises whole words from a vocabulary list stored in pairs of ROMs. To expand the vocabulary it is only a matter of purchasing further ROMs. The Turtle Talk board has provision for expansion to a total of eight ROMs, that is — four word lists giving a vocabulary of around 550 words plus prefixes, suffixes, tones and silences. The kits are supplied with the 'standard vocabulary' ROMs (Master Word List, Table 1, Part 1) which includes all the letters of the alphabet and all the numbers needed to form any finite set of numbers. Having this vocabulary allows all words which are not on the list, to at least be spelt out.

National Semiconductor also have available ROMs containing vocabularies of foreign languages — German, French, Italian, etc. Associated with this large potential vocabulary, is the Turtle Talk's special power saving circuitry to cut down heating and power consumption by chopping the supply to the ROMs when they are not being used.

A COUPLE OF DEMONSTRATION PROGRAMS

These programs were written for a Tasman Turtle Robot fitted with a General Purpose Interface board and a Turtle Talk board, interfaced to an Apple II via slot 2. In this configuration, the robot has address W (where W = -16224) and the speech synthesiser addresses become W + 2 and W + 3.

LOAD COUNT TO 100
LIST

```
1 W = -16224
2 DEF FN DD(X) = INT (X / 10)
3 TL = 600
4 PRINT I
100 FOR I = 1 TO 100: GOSUB 200: NEXT I: END
200 IF I = 0 THEN WD = 31: GOSUB 1000: GOTO 230
210 IF I < 21 THEN WD = I: GOSUB 1000: GOTO 230
220 IF I < 100 THEN WD = INT (I / 10) + 10: GOSUB 1000: IF NOT FN DD(I)
    THEN WD = (I / 10) - INT (I / 10) * 10: GOSUB 1000
230 FOR T = 1 TO TL: NEXT T: RETURN
1000 IF PEEK (W + 3) < 255 THEN 1000
1005 IF I < J THEN PRINT I
1010 POKE W + 2, WD: POKE W + 3, 8: POKE W + 3, 16
1015 J = I
1020 RETURN
```

LOAD TO SAY A SENTENCE
LIST

```
10 W = -16224: REM TURTLE IN SLOT 2 OF APPLE
20 READ A,B
25 IF A = 0 THEN END
30 POKE W + 2, A
40 POKE W + 3, 8
50 POKE W + 3, 16 + (2 * B)
60 IF PEEK (W + 3) < 255 THEN 60
70 GOTO 20
100 DATA 65,0,65,0,66,0,71,0,18,0,115,0,96,0,102,1,61,0,126,1,87,1,116,
    1,71,0,26,1,78,0,129,0,61,0,128,1,
110 DATA 0,0
```

LOAD SPELL
LIST

```
5 T1 = 130
10 W = -16224
20 A$(1) = "BUTTON": C(1) = 10
21 A$(2) = "COMPLETE": C(2) = 18
22 A$(3) = "HELP": C(3) = 58
23 A$(4) = "NEXT": C(4) = 76
24 A$(5) = "SOUTH": C(5) = 110
45 FOR K = 1 TO 5
60 GOSUB 200
70 POKE W + 2, C(K)
80 POKE W + 3, 8
90 POKE W + 3, 16
100 INPUT B$
110 IF B$ = A$(K) THEN 300
120 GOSUB 400
130 GOTO 60
150 NEXT K
160 END
200 POKE W + 2, 133
210 POKE W + 3, 8
220 POKE W + 3, 16
230 FOR T = 1 TO T1: NEXT T
240 POKE W + 2, 43
250 POKE W + 3, 8
260 POKE W + 3, 16
270 IF PEEK (W + 3) < 255 THEN 270
280 RETURN
300 POKE W + 2, 22
310 POKE W + 3, 8
320 POKE W + 3, 16
330 IF PEEK (W + 3) < 255 THEN 330
340 GOTO 150
400 POKE W + 2, 61
410 POKE W + 3, 8
420 POKE W + 3, 16
430 IF PEEK (W + 3) < 255 THEN 430
440 POKE W + 2, 140
450 POKE W + 3, 8
460 POKE W + 3, 16
470 IF PEEK (W + 3) < 255 THEN 470
480 POKE W + 2, 58
490 POKE W + 3, 8
500 POKE W + 3, 16
510 IF PEEK (W + 3) < 255 THEN 510
520 RETURN
LOAD SENTENCE
LIST
10 W = -16224: REM TURTLE IN SLOT 2 OF APPLE
20 READ A,B
25 IF A = 0 THEN END
30 POKE W + 2, A
40 POKE W + 3, 8
50 POKE W + 3, 16 + (2 * B)
60 IF PEEK (W + 3) < 255 THEN 60
70 GOTO 20
100 DATA 73,0,76,1,73,1,2,0,86,0,
110 DATA 0,0
```

LOAD VOCABULARY
LIST

```
5 PRINT A:
10 W = -16224
20 POKE W + 2, A
30 POKE W + 3, 8
40 POKE W + 3, 16 + (B * 2)
45 IF PEEK (W + 3) < 255 THEN 45
50 A = A + 1
52 IF C = 1 AND A = 131 THEN END
55 IF A = 143 THEN C = 1: A = 0: C = 1
60 PRINT A: " "
70 GOTO 20
```

LOAD TALKING KEY BOARD
LIST

```
2 REM TALKING KEYBOARD BY ALLAN BRANCH 1982
5 W = -16224
6 CALL -936
10 GET A#
15 TT = 100
20 X = ASC (A#)
30 IF (64 < X) AND (X < 91) THEN 100
40 IF (48 < X) AND (X < 58) THEN 200
50 IF X = 32 THEN X = 132: GOTO 100
60 IF X = 13 THEN END
99 GOTO 10
100 X = X - 33
105 PRINT A#
110 POKE W + 2, X
120 POKE W + 3, 8
130 POKE W + 3, 16
140 GOTO 10
200 X = X - 48
210 GOTO 100
```

WORD LIST 2

| Word | 8-Bit Binary Address | | Word | 8-Bit Binary Address | | Word | 8-Bit Binary Address | |
|-------------|----------------------|-----|-----------|----------------------|-----|--------------|----------------------|-----|
| | SW8 | SW1 | | SW8 | SW1 | | SW8 | SW1 |
| ABORT | 00000000 | | FARAD | 00101100 | | PER | 01011000 | |
| ADD | 00000001 | | FAST | 00101101 | | PICO | 01011001 | |
| ADJUST | 00000010 | | FASTER | 00101110 | | PLACE | 01011010 | |
| ALARM | 00000011 | | FIFTH | 00101111 | | PRESS | 01011011 | |
| ALERT | 00000100 | | FIRE | 00110000 | | PRESSURE | 01011100 | |
| ALL | 00000101 | | FIRST | 00110001 | | QUARTER | 01011101 | |
| ASK | 00000110 | | FLOOR | 00110010 | | RANGE | 01011110 | |
| ASSISTANCE | 00000111 | | FORWARD | 00110011 | | REACH | 01011111 | |
| ATTENTION | 00001000 | | FROM | 00110100 | | RECEIVE | 01100000 | |
| BRAKE | 00001001 | | GAS | 00110101 | | RECORD | 01100001 | |
| BUTTON | 00001010 | | GET | 00110110 | | REPLACE | 01100010 | |
| BUY | 00001011 | | GOING | 00110111 | | REVERSE | 01100011 | |
| CALL | 00001100 | | HALF | 00111000 | | ROOM | 01100100 | |
| CAUTION | 00001101 | | HELLO | 00111001 | | SAFE | 01100101 | |
| CHANGE | 00001110 | | HELP | 00111010 | | SECURE | 01100110 | |
| CIRCUIT | 00001111 | | HERTZ | 00111011 | | SELECT | 01100111 | |
| CLEAR | 00010000 | | HOLD | 00111100 | | SEND | 01101000 | |
| CLOSE | 00010001 | | INCORRECT | 00111101 | | SERVICE | 01101001 | |
| COMPLETE | 00010010 | | INCREASE | 00111110 | | SIDE | 01101010 | |
| CONNECT | 00010011 | | INTRUDER | 00111111 | | SLOW | 01101011 | |
| CONTINUE | 00010100 | | JUST | 01000000 | | SLOWER | 01101100 | |
| COPY | 00010101 | | KEY | 01000001 | | SMOKE | 01101101 | |
| CORRECT | 00010110 | | LEVEL | 01000010 | | SOUTH | 01101110 | |
| DATE | 00010111 | | LOAD | 01000011 | | STATION | 01101111 | |
| DAY | 00011000 | | LOCK | 01000100 | | SWITCH | 01110000 | |
| DECREASE | 00011001 | | MEG | 01000101 | | SYSTEM | 01110001 | |
| DEPOSIT | 00011010 | | MEGA | 01000110 | | TEST | 01110010 | |
| DIAL | 00011011 | | MICRO | 01000111 | | TH (NOTE 2) | 01110011 | |
| DIVIDE | 00011100 | | MORE | 01001000 | | THANK | 01110100 | |
| DOOR | 00011101 | | MOVE | 01001001 | | THIRD | 01110101 | |
| EAST | 00011110 | | NANO | 01001010 | | THIS | 01110110 | |
| ED (NOTE 1) | 00011111 | | NEED | 01001011 | | TOTAL | 01110111 | |
| ED (NOTE 1) | 00100000 | | NEXT | 01001100 | | TURN | 01110000 | |
| ED (NOTE 1) | 00100001 | | NO | 01001101 | | USE | 01110001 | |
| ED (NOTE 1) | 00100010 | | NORMAL | 01001110 | | UTH (NOTE 3) | 01110101 | |
| EMERGENCY | 00100011 | | NORTH | 01001111 | | WAITING | 01110111 | |
| END | 00100100 | | NOT | 01010000 | | WARNING | 01111100 | |
| ENTER | 00100101 | | NOTICE | 01010001 | | WATER | 01111101 | |
| ENTRY | 00100110 | | OHMS | 01010010 | | WEST | 01111110 | |
| ER | 00100111 | | ONWARD | 01010011 | | SWITCH | 01111111 | |
| EVACUATE | 00101000 | | OPEN | 01010100 | | WINDOW | 10000000 | |
| EXIT | 00101001 | | OPERATOR | 01010101 | | YES | 10000001 | |
| FAIL | 00101010 | | OR | 01010110 | | ZONE | 10000010 | |
| FAILURE | 00101011 | | PASS | 01010111 | | | | |

The MPF-1 Micro-Professor — tool, or toy?

Jonathan Scott

Described in the brochures as “a learning tool for hobbyists, students and microprocessor enthusiasts...”, the Micro-Professor is a new concept in computer or microprocessor products. Is it a tool, as the makers claim, or a toy?

THE MPF-1 Micro-Professor, manufactured by Multitech Industrial Corporation, is a Z80-based microcomputer system which fills a need which no other system we have yet to see or hear about can. It is more than an ‘evaluation kit’, but it is definitely not a ‘home’ or personal computer in the usual sense. It is a most marvellous device for actually *teaching* someone what a microprocessor is and how to use it. One emerges from a relationship with this little fellow with familiarity and skill enough to tackle the design of a microprocessor-based appliance, scientific instrument or even a pinball machine. These are the fundamental commercial uses of microprocessors, where the designer must have close contact with hardware, software and layout of the system. In these areas the MPF-1 will support a student in the processes of acquiring familiarity and technique. The MPF-1 is neither a games-player nor a computer for a serious computational task. If you want to play games go buy a VIC-20 or some such, and if you need to do serious calculations, go buy an HP. The MPF-1 can be fun, but only in the way that learning can be a pleasurable challenge for its own sake. If you seriously intend to expand your commercial ability in the digital area, or if you like to get at the roots of microprocessing, rather than fool about in pure software, this is probably the best system we have ever seen for you. Having now discharged my duty to ward off the games players and the heavy number crunchers, let me proceed to describe this fascinating system.

The basic board

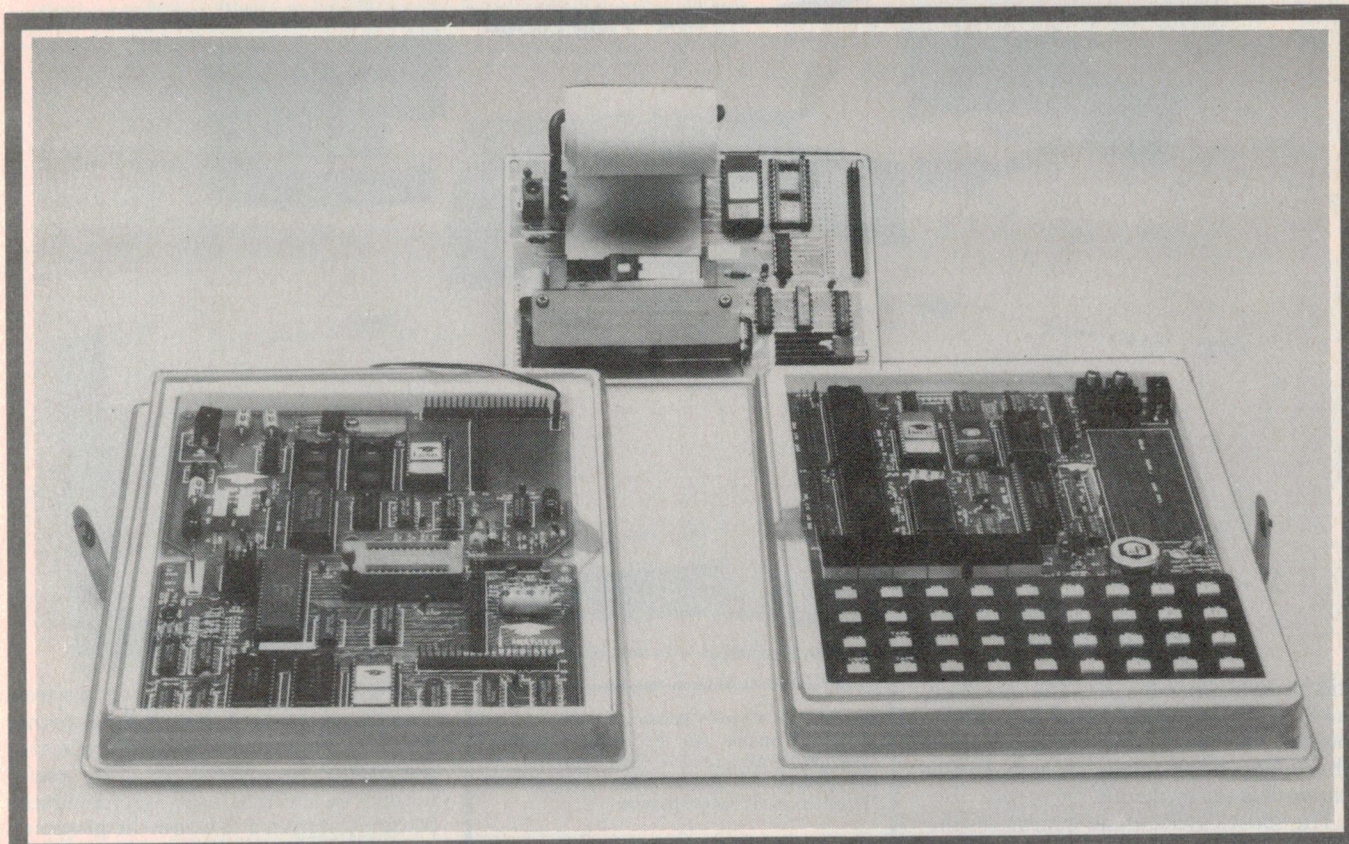
The MPF-1 itself comes as a single pc board about 150 x 220 mm. This is powered by a plugpack and sports a 9 x 4 keyboard and a six-character 7-segment display. A cassette interface, a speaker and two separate indicator LEDs are provided. There is an area for user hardware expansion which would support about half a dozen 16-pin chips. There are four IC sockets unoccupied on the board. Two multipin connectors are mounted on the left side of the board. The keyboard is labelled with the monitor commands, and is also designed to accept overlays with new key designations, rather like an HP41. It has a nice feel and the keys are spaced quite comfortably. It is supplied with three books — The User's Manual, an Experiment Manual and the Monitor Source Listing.

Before going on to discuss the beastie, let me list all the options which we received along with the MPF-1 itself. The four empty sockets were filled with a CTC (Counter-Timer-Chip), a peripheral interface adaptor (PIA), which used the second connector on the pc board, an EPROM of BASIC, and further RAM or EPROM which could be used as desired. Chained on to the first connector we had the EPROM programmer option (EPB-MPF), a separate board half the size of the first, also powered by another plugpack. From this we chained the SSB-MPF speech synthesiser — same size, same power arrangement. Beyond this was the PRT-MPF printer, which carried more software in EPROM, and

room for expansion. The speech synthesiser had room for further vocabulary, with which we were not supplied. This space could also presumably support RAM or whatever you wanted. Two of the added boards could be housed with the MPF-1 in its book-like mounting arrangement, though this turns out to be not very useful. Each additional board came with a small booklet, as did the BASIC EPROM.

The MPF-1 is very well built, in hardware terms. The pc board is good quality and has the component labels screened onto it. The keyboard, as I have said, is neat and has a good positive ‘feel’. The only complaint we might level at the makers is that the LED display does not have an antiglare cover sufficient to eliminate reflections from bright lamps or to maintain contrast in a very light room. The display, speaker and keyboard are controlled by an on-board PIA, as is the cassette interface. The interrupt line and a reset facility are included in the keypad, so you can get at the hardware even before you add things to the user's addition area.

The monitor with which it is supplied is *excellent*. It provides easy and ergonomic inspection of registers and memory, and simple alteration facility. It permits single stepping, an important function absent from many monitors. It allows the setting of a breakpoint. It also automatically calculates and stores relative addresses, which is so useful an enhancement for anyone practicing hand assembly of machine level programs. Finally, it will insert, delete and move blocks of instructions. For a



The 'full kit & kaboodle' Micro-Professor system. On the right, is the MPF-1, on the left are the EPB-MPF EPROM programmer (foreground) and SSB-MPF speech synthesiser (behind the EPROM programmer). Right at the rear is the PRT-MPF printer module. The case in which the MPF-1 arrives is ingeniously designed so that manuals or the other boards (as shown here) can be accommodated.

student who is using hand assembly rather than an assembler program initially, this is superbly handy. I realise that no-one these days develops software seriously without assembler support, but it is nevertheless the best way to familiarise oneself with the inner workings of a processor. With all the options available, the best task a student can have on the MPF-1 is, in any case, to write and install in EPROM an assembler. For this grass-roots warm-up to microprocessors the monitor supplied is absolutely ideal. It is also well documented and has a good repertoire of utility subroutines to facilitate the user, including standard functions from the monitor in personally developed utility software.

The User's Manual and Experiment handbook are indeed comprehensive and cover the subjects well, but there is one serious flaw. They are, like every manual supplied with the MPF boards, written in ghastly English. I think they were translated hurriedly using a tourist's phrasebook. True, the Sydney agents, Emona, did supply a replacement BASIC manual, which was a great improvement over the preceding comedy exercise, but it still used clumsy constructions. The original was frequency incomprehensible, and the humour wears off soon if you are trying to extract information. For this reason, I

hesitate to recommend that anyone try to learn from the MPF course if there is not someone accessible who knows what is going on already, and is willing to be helpful when the task gets too hard.

The PIA and CTC integrated circuits were supplied without any documentation, as they quite reasonably state that the Zilog data sheets for the various ICs are the necessary and best instructions for using them. With the addition of these you are ready to embark on the jobs such as installation of custom hardware on the user expansion board area. This might mean the connection of analogue-to-digital (A/D) converters or position sensors, or whatever you need. A stepper motor controller and some sensor inputs might allow the computer control of a toy truck or model train.

BASIC

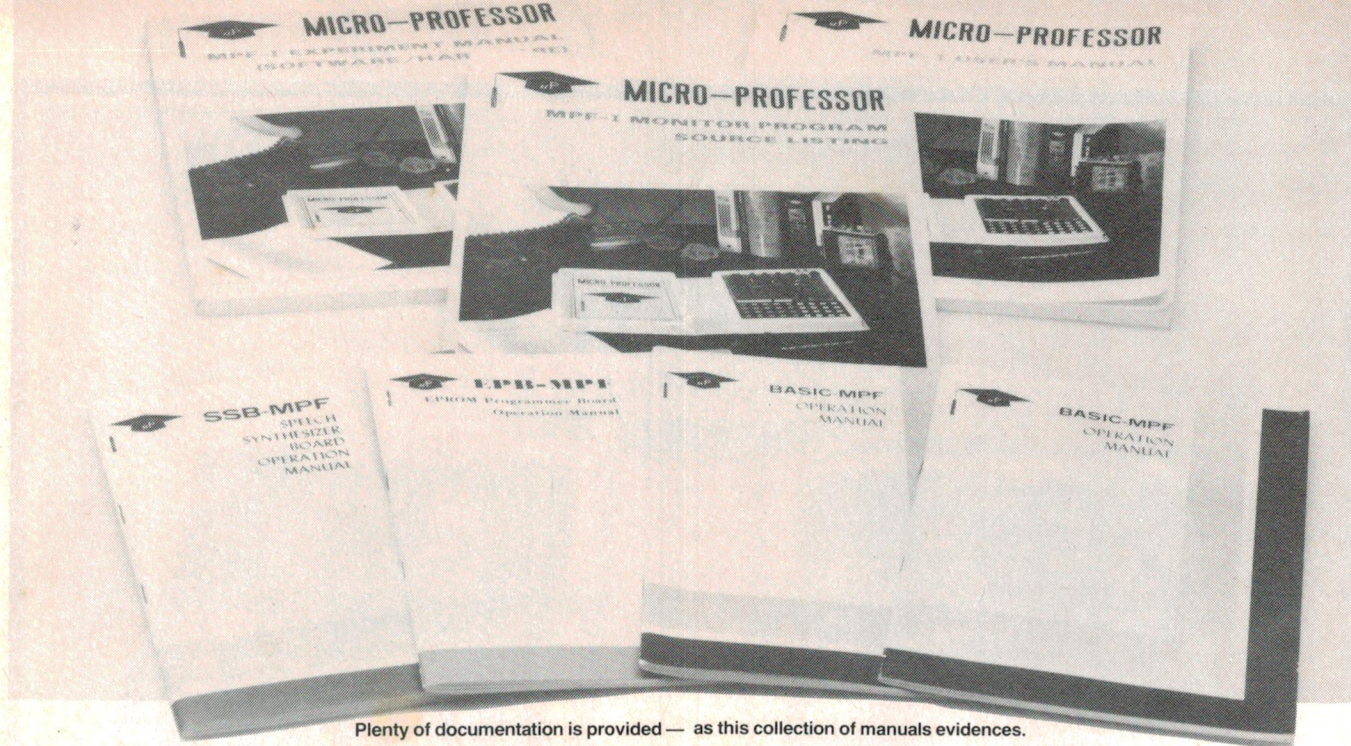
The BASIC EPROM contains a 4K tiny BASIC interpreter. It is supplied with a (hopefully revised) manual. As you might imagine, BASIC is pretty limited when you have only a 6-digit, 7-segment display. However, a very credible attempt has been made to allow full BASIC operation. You can edit and list, albeit rather clumsily. This software is enough to allow one to see how a machine expands from assembly language to a 'full high level language', as such. This

is the only option which I do not recommend you buy with the MPF-1, and this is purely because the MPF-1 is plainly not a BASIC-orientated machine. For those who suspect that they will have to have a slight BASIC familiarity, it will do, as one could subsequently migrate to a proper machine solely by reading the list of instructions available on the machine's operating system and the list of functions supported by the system's BASIC. There is no criticism here against Multitech's programmers.

'Expansion' boards

All three expansion boards (the EPROM programmer, the speech synthesiser and printer) are constructed with the same high quality as the MPF-1 host board, and are well matched and integrated with it. So much so, they merge in the mind to become part of it, more than peripherals.

The EPROM programmer (EPB-MPF) has the standard zero insertion force (ZIF) socket, and supports a range of EPROMs, including the 2716, 2532, etc. It is supplied with a well set-out, if not fluent, instruction manual and a keypad overlay. It also comes with a plug-pack supplying the appropriate voltages. It allows programming and verification of EPROMs and listing of data in the buffer. On the whole it is a very neat and



Plenty of documentation is provided — as this collection of manuals evidences.

adequate assembly. All the relevant main monitor subroutines are implemented on the buffer (4K) on-board. This RAM can, of course, be regarded as expanding the system RAM to 6K.

The speech synthesiser board uses the Texas Instruments chip set. This uses the 'linear predictive coding' method, and sounds just like a TI speak-and-spell. It is supplied with plugpack and instruction booklet. There is an EPROM on the board which contains a speak-the-time type of clock as a demonstration, as well as a very useful utility for quickly getting the system working with your software. It comes with the ability to say the numbers 'one, two, ...' up to twenty, then by tens to fifty, etc, as well as things like 'good morning/afternoon', etc. This is, of course, exactly the vocabulary a clock requires. More words are listed in the booklet as being available, and expansion sockets are ready on the pc board to accept the new chips but we were not supplied with any others. There seems to be insufficient space on the board to hold all the words available at once, which might prove tricky as the complete vocabulary comes in alphabetical chunks, so perhaps you would have to either transfer appropriate data around by using the excellent EPROM programmer and put all you wanted in one of your own EPROMs, or limit yourself to what part of the vocabulary fits in three chips. Otherwise, you might use some of the expansion space on the other boards to hold the excess chips. All in all, this is a brilliant fun board, and provides a perfect introduction to this method of speech synthesis. The documentation is OK, though takes a little getting used to. It is clear that further vocabulary is necessary if you don't want to get bored

The MPF-1 Micro-Professor, accessories and expansion boards are marketed in Australia by Emona Enterprises P/L, CBC Bank Building, 661 George St, Sydney 2000. (02)212-4815. Prices are as follows (all plus tax):

| | |
|---|---------|
| MPF-1 'host' board, with manuals (plugpack \$11.50) | \$115 |
| MPF-CPK CTC and PIO chip kit | \$17 |
| MPF-2KRAM, 2Kx8 6116 RAM or equiv. | \$12.50 |
| MPF-2KROM, blank 2K EPROM | \$8 |
| MPF-4KROM, blank 4K EPROM | \$12.50 |
| EPB-MPF EPROM programmer board with manual (plugpack extra) | \$145 |
| SSB-MPF speech synthesiser board with manual (plugpack extra) | \$145 |
| PRT-MPF printer (plugpack extra) | \$95 |

The Micro-Professor equipment is distributed in Victoria by the Radio Parts Group, 562 Spencer St, West Melbourne 3003. (03)329-7888.

quickly. I would have liked to see if the further vocabulary could be manipulated around using the other facilities. There seems to be no reason why this could not be achieved readily with the EPB board.

Finally, the printer (PRT-MPF). This is a marvellously compact unit, being all up the same size as the other options. It is supplied with a book and plugpack, as are the others. It prints, albeit at less than one line per second, up to 20 characters per line on thermal paper. The result is black on white and very readable. The board has space for the addition of one further EPROM or RAM chip, just as a method of using board space economically. It comes supplied with software to disassemble Z80 code (very nicely) and to list BASIC, as it is not of course limited to 7-segments as is the MPF-1 display. These, as well as a memory dump program, are contained with the driving utility routines in the

EPROM on-board. Because of this freedom to print in dot matrix form rather than the inherent hardcopy output, this particular peripheral is very useful indeed. In addition, sufficient documentation is supplied to allow the programmer to get down to the very dot delivery level, so that one can obtain a complete hardware familiarity with this species of printer.

Summary

As I said when I started, the MPF-1 system is fundamentally a teaching system which exposes its designer's thoughts and techniques well to the student. It is comprehensive and detailed in its design and programming. It is also particularly good monetary value. Schools and other teaching institutions can afford several of these I am sure. While the ever popular 'home' computer plays games and BASIC is the language that people like kids to get into, games are not the stuff microprocessors are practically used for in commercial situations and BASIC is a language on the way out, I hope. Further, one printer, one EPROM programmer and a single speaking box can service several MPF-1 boards. Two of each mobile peripherals would support up to a dozen MPF-1s. This is the way schools and universities are going to have to go, to teach *micro-processing* as different from computing.

In conclusion, the MPF-1 system is a beautifully designed specific purpose gadget. I recommend it wholeheartedly for the purpose for which it is intended, but I must reiterate my warning that it is not general purpose and hence not to be regarded as the basis for an expandable system for playing or number crunching. In short — for the dedicated student, not the frivolous.

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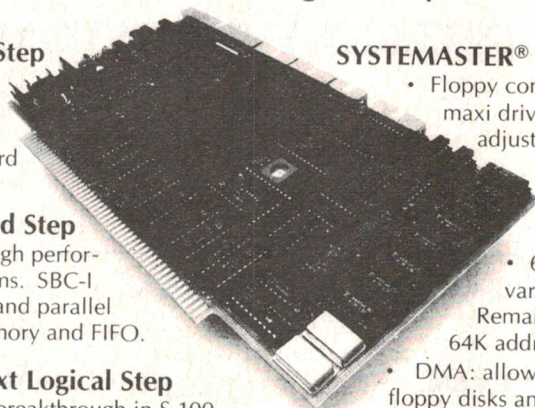
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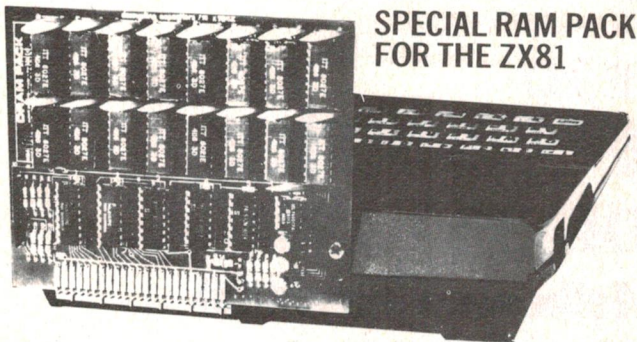
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VENDALE

The '660 Program Potpourri has had to be held over due to lack of space, but here's a couple of good programs to keep you amused in the meantime. Both are listed in compact form,

WIPEOUT '660 STYLE

W.F. Kreykes
St. Albans, Vic.

This is not a remake of 'One-handed Pong' as published in the April '82 ETI. Firstly, this program is in colour and will absorb the player for much longer. Sound effects are included. You'll need to have installed the colour components and have 3K of memory on your '660.

You start the game with 200 dots and a large bat. A ball moves around the screen and the object is to position your bat so that the ball bounces from it, and in doing so, wipes out as many dots as possible. For this, you score points. If the ball moves toward you and you don't get the bat under it, the ball goes off screen. Your progressive score is then displayed and you get another ball. You start the game with 15 balls. You get a bonus of one extra ball if 10 dots are wiped out with one ball. If you wipe out at least 20 dots with one ball you get a bonus of three extra balls. When you have wiped out a total of 100 dots, the bat size automatically decreases 25% — making the game from that point on a little harder! The game ends when you either run out of balls or run

showing program code only. We'll leave it to you to disassemble them and find out how they work.

out of dots. The game restarts automatically after ending and the highest score to beat is displayed on-screen.

Use the following keys to control the bat:

KEY A — LEFT
KEY D — RIGHT

The bat does not go off-screen at extreme left or right travel.

If you want a monochrome game (or white on blue background), change 2822 at 0600 to 00FF then type in the program from 0600 to 07AC and disregard the rest (and for this you only need 1K of memory). Here are the colour setups:

| | |
|-----------|--|
| 00600 | 2822: calls colour routines. Set to 00FF for mono. |
| 0822 | 07F7: sets background to black. For blue set to 00FF |
| 0824 | 07B8: initiates colour operation |
| 0828 | 6007: white — number of hits |
| 082C | 6006: pale blue — first divider |
| 0830 | 6003: violet — highest score to beat |
| 0836 | 6004: green — progressive score |
| 083C | 6005: yellow — third divider |
| 0840 | 6001: red — number of balls left |
| 0844-085C | colours the dots but jumps colour black |
| 085A | 6001: red — bat |

TRADITIONAL SPACE INVADERS

Master P. Easdown
Kew, Vic.

Every computer must have a version of the traditional arcade game of 'Space Invaders'. This one has the familiar row of invading characters marching across and down the screen dropping missiles on you and which you fire at from behind shields at the bottom of the screen. This is a monochrome program and requires only 1K of memory. Sound effects are included.

Four invaders march across and down the screen, firing at you at random intervals. Your cannon is located at the bottom of the screen, just above which are four shields which help protect you from the invaders' bombs. The bombs dropped by the invaders and your cannon shots are single pixels on the screen.

Cannon control is as follows:

| | |
|-------|-------|
| KEY 4 | LEFT |
| KEY 5 | FIRE |
| KEY 6 | RIGHT |

The game can end in two ways: one is, if you get shot, the other is if the invaders get so low that it is impossible for you to shoot them — i.e. you've been invaded!

When the game ends, your score is shown on the screen.

Save this program on tape and save yourself a fortune!

WIPEOUT '660 STYLE

| | | | | | | | | |
|------|------|------|------|------|------|------|------|------|
| 0600 | 2822 | 6000 | 2798 | 00E0 | 6C0A | 6D00 | A7A6 | DCD6 |
| 0610 | 6C1E | DCD6 | 6C32 | DCD6 | A7A4 | 6D06 | 1620 | A6B5 |
| 0620 | 6C00 | DCD1 | 7C08 | FD00 | FD18 | 3C40 | 1622 | 7D03 |
| 0630 | 3D27 | 161E | 60FE | A7A5 | F055 | A7A5 | 681C | 692E |
| 0640 | D891 | 670F | 6D00 | 279C | 6600 | 2758 | 632C | 6B3E |
| 0650 | 8B82 | 4700 | 1760 | F70A | 4F0A | 1660 | 3F0D | 1656 |
| 0660 | 274A | 6D00 | 2758 | 77FF | 2758 | A6D2 | DB31 | 65FF |
| 0670 | C401 | 3401 | 64FF | A7A5 | 6C00 | 6E0A | EEA1 | 1690 |
| 0680 | 6E0D | EE9E | 169C | 483A | 169C | D891 | 7C02 | 1698 |
| 0690 | 4800 | 169C | D891 | 7CPE | 88C4 | D891 | 4F01 | 1704 |
| 06A0 | 4B00 | 6401 | 4B3E | 64FF | 4306 | 6501 | 432E | 1712 |
| 06B0 | A6D2 | DB31 | 8E44 | 8354 | DE31 | 3F01 | 1676 | 432E |
| 06C0 | 1704 | FE00 | 6A02 | FA18 | 7D01 | 4D0A | 2750 | 4D14 |
| 06D0 | 2750 | 80D0 | 8064 | 40C8 | 175E | 3064 | 16P2 | A7A5 |
| 06E0 | D891 | 6A05 | 6110 | FA00 | F118 | 60F8 | F055 | A7A5 |
| 06F0 | D891 | A6D2 | DB31 | C401 | 3401 | 64FF | C501 | 3501 |

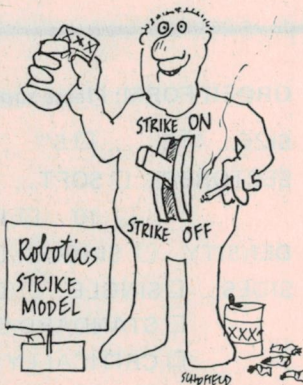
| | | | | | | | | |
|------|------|------|------|------|------|------|------|------|
| 0700 | 65FF | 1676 | F900 | 6A03 | FA18 | A6D2 | DB31 | 73FF |
| 0710 | 166A | A6D2 | DB31 | 271A | 164C | 274A | 2720 | 86D4 |
| 0720 | 6E23 | 8060 | 6A00 | A7AD | F033 | F265 | 4000 | 1740 |
| 0730 | F029 | 273A | F129 | 273A | F229 | DEA5 | 7E05 | 00EE |
| 0740 | 3100 | 1734 | 3200 | 1738 | 00EE | 6E00 | 80D0 | 1724 |
| 0750 | 2758 | 3D0A | 7701 | 7701 | 6E38 | 8070 | 1724 | 271A |
| 0760 | A7A5 | D891 | A7AC | F065 | 8065 | 4F00 | 2794 | 6A00 |
| 0770 | A7A5 | DA91 | 7A08 | 3A40 | 1772 | 6703 | F715 | F707 |
| 0780 | FA00 | FA18 | 3700 | 177E | A6D2 | DA91 | 7AFF | 3A00 |
| 0790 | 177A | 1606 | 279C | 8060 | A7AC | F055 | A7AC | F065 |
| 07A0 | 6E0F | 1724 | BAAA | 60F0 | F0F0 | 6000 | 0000 | 0000 |

| | | | | | | | | |
|------|------|------|------|------|------|------|------|------|
| 07B0 | A7AD | F255 | 07C8 | 00EE | F839 | AF96 | BFEF | F82C |
| 07C0 | 5F62 | 2FF8 | 205F | 62D4 | F807 | BEF8 | ADAE | EE72 |
| 07D0 | FA07 | BFF0 | FA07 | 5E1E | F0FA | 1FFE | FEFE | FE5E |
| 07E0 | F80C | 7C00 | BDF8 | 80F4 | AF9D | 7C00 | BD8F | 2EF4 |
| 07F0 | ADED | 9F5D | 63E2 | D4E9 | 61D4 | 6F01 | 6100 | 27B0 |

| | | | | | | | | |
|------|------|------|------|------|------|------|------|------|
| 0800 | 7101 | 3108 | 17FE | 7201 | 7FFF | 3F00 | 17FC | 00EE |
| 0810 | 6200 | 6F03 | 27B0 | 7201 | 7FFF | 3F00 | 1814 | 7101 |
| 0820 | 00EE | 07F7 | 07B8 | 6100 | 6007 | 2810 | 6006 | 2810 |
| 0830 | 6003 | 2810 | 2810 | 6004 | 2810 | 2810 | 6005 | 2810 |
| 0840 | 6001 | 2810 | 6203 | 6007 | 27FA | 70FF | 4213 | 1856 |
| 0850 | 3000 | 1848 | 1846 | 6F04 | 27FC | 6001 | 17FA | |

SPACE INVADERS PROGRAM

| | | | | | | | | | | | |
|------|---|------|------|------|------|-----|---|------|------|------|------|
| 0600 | - | 6100 | 6202 | 6C00 | 6B00 | 708 | - | 00E0 | 6410 | A7FC | FD33 |
| 608 | - | C530 | 6629 | 6D00 | 6004 | 710 | - | F265 | F029 | D345 | 7304 |
| 610 | - | 6B19 | A72C | DOE8 | 700F | 718 | - | F129 | D345 | 7304 | F229 |
| 618 | - | DOE8 | 26D6 | A724 | D568 | 720 | - | D345 | 0000 | 1010 | 387C |
| 620 | - | 8A10 | 8920 | 2762 | DA98 | 728 | - | 387C | FE00 | 3C7E | FFE7 |
| 628 | - | 2744 | 7A0A | 2762 | DA98 | 730 | - | E300 | 0000 | 387C | D6FE |
| 630 | - | 2744 | 7A0A | 2762 | DA98 | 738 | - | 2844 | 8200 | 387C | D6FE |
| 638 | - | 2744 | 7A0A | 2762 | DA98 | 740 | - | 2844 | 2800 | CC03 | 4C01 |
| 640 | - | CC03 | 4C01 | 2688 | A724 | 748 | - | 2688 | 00EE | 78FD | A770 |
| 648 | - | 6E04 | D568 | EEA1 | 75FE | 750 | - | D788 | 6B10 | FB15 | FB07 |
| 650 | - | 6E05 | EEA1 | 26AE | 6E06 | 758 | - | 3B00 | 1756 | D788 | A724 |
| 658 | - | EEA1 | 7502 | D568 | 8A10 | 760 | - | 00EE | 4000 | 176A | A73C |
| 660 | - | 8920 | 2762 | DA98 | 7A0A | 768 | - | 00EE | A734 | 00EE | 7001 |
| 668 | - | 2762 | DA98 | 7A0A | 2762 | 770 | - | 4002 | 6000 | 1620 | 4299 |
| 670 | - | DA98 | 7A0A | 2762 | DA98 | 778 | - | FB1A | E500 | 0775 | 00FF |
| 678 | - | 7101 | 311B | 176E | 7203 | 780 | - | 6B10 | FE00 | FB18 | 7B01 |
| 680 | - | 6100 | 4220 | 1708 | 176E | 788 | - | 4B15 | 00EE | 1782 | 00FF |
| 688 | - | 83A0 | 8490 | 7407 | A724 | | | | | | |
| 690 | - | 7401 | D341 | 4F01 | 16E2 | | | | | | |
| 698 | - | D341 | 442F | 00EE | 1690 | | | | | | |
| 6A0 | - | 4429 | 1708 | 00FF | 6B05 | | | | | | |
| 6A8 | - | F900 | FB18 | 00EE | 2780 | | | | | | |
| 6B0 | - | 00FF | 00FF | 8750 | 8860 | | | | | | |
| 6B8 | - | 78F8 | A724 | 78FF | D781 | | | | | | |
| 6C0 | - | 4501 | 160C | D781 | 4800 | | | | | | |
| 6C8 | - | 00EE | 16BC | 16EE | A720 | | | | | | |
| 6D0 | - | 274C | 7D01 | 00EE | 700F | | | | | | |
| 6D8 | - | DOE8 | 700F | DOE8 | 6000 | | | | | | |
| 6E0 | - | 00EE | 6B22 | 8C40 | 8BC5 | | | | | | |
| 6E8 | - | 3F01 | 1708 | 16A4 | 6B18 | | | | | | |
| 6F0 | - | 8C80 | 8BC5 | 3F01 | 16A6 | | | | | | |
| 6F8 | - | D781 | 6B20 | FE00 | FC18 | | | | | | |
| 700 | - | 7BFF | 4B10 | 16D0 | 16FC | | | | | | |



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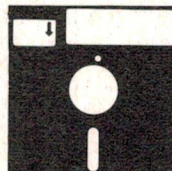
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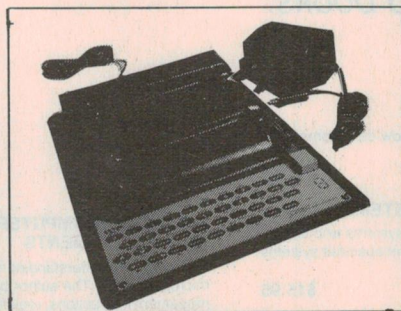
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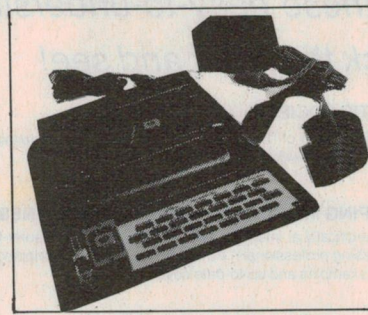


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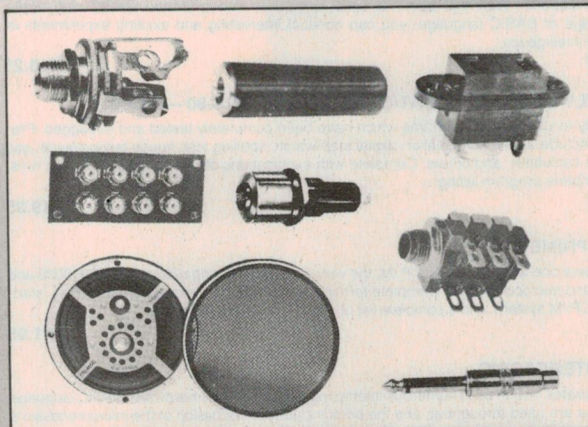
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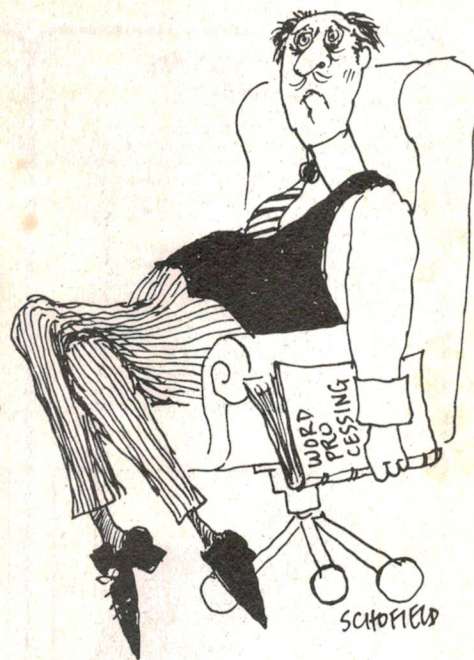
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ELECTRONIC LIFESTYLE

Row brewing over Sydney FM broadcasting

An argument is brewing between Sydney's FM broadcasters and the Federal Government over siting of the transmitters, the broadcasters claiming that a 'towering headache' faces the Government if a suitable site is not found where all services can share an antenna mast.

Sydney, unlike the other capitals, does not have a Black Mountain, Mount Dandenong, Mount Lofty or Mount Cootha etc and siting of broadcast antenna towers is a planners' headache if adequate coverage of what is Australia's biggest audience area is to be obtained.

For some reason, as yet unrevealed, the Federal Government appears to have passed up the opportunity to acquire what would appear to be an excellent site to locate all the FM broadcasters — Centrepont tower.

Ideally, all services should be located on a single site, with allowance made to add further broadcasters as they become licensed. This means every listener can point their antennas in the one direction for reception of all the FM stations.

At present, FM station transmitting antennas are scattered between city sites, North Sydney and Artarmon. Some stations cannot be received over wide areas owing to Sydney's terrain.

In March, the Department of Communications proposed the cositing of FM transmissions at Centrepont, with one antenna, shared by all stations, mounted on the tower. Discussions between the stations and the Department in subsequent months considered and

compared three sites: the ABC site at Gore Hill, the Northpoint building in North Sydney and the Centrepont tower. For a variety of reasons, Centrepont was the broadcasters' preferred site and discussions with D.O.C. proceeded.

Late in July the Department of Communications suddenly switched, naming Northpoint as their preferred option.

2MBS-FM immediately argued strongly against this switch, making submissions to the Minister for Communications against the Northpoint site. The station also objected to the failure of the Department to call for public comment on its proposal or to indicate the attitude to its plans of other affected authorities such as North Sydney council — Northpoint would have to have a 48 m tower built on its roof.

2MBS met with the Minister for Communications, Mr Neil Brown, on the 2nd of August and made submissions on the subject on 4th August. 2MBS says the Northpoint site, favoured by the D.O.C. is an expedient and makeshift decision and will only lead to further problems and expense for Sydney's FM broadcasters.

One wonders why the D.O.C. is being so coy about the switch to Northpoint and why they have made no move to call for public comment on the various proposals.

B & W's leisure monitor

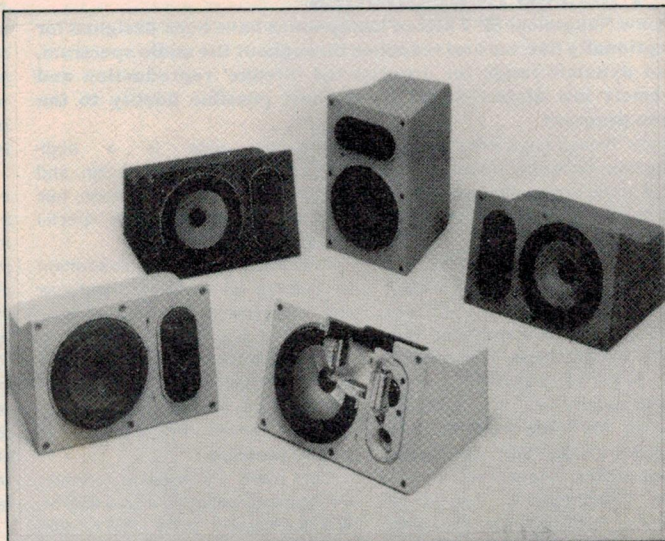
B & W's new LM1 'leisure monitor' design is aimed at providing a high-fidelity loudspeaker for use in cars, boats and caravans etc, as well as the home.

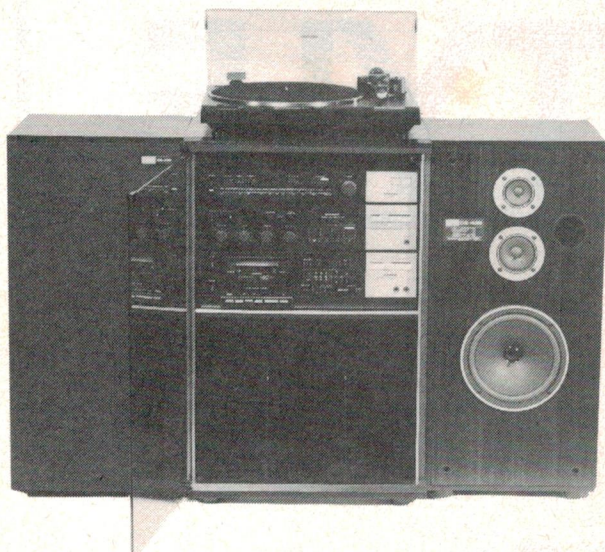
Its modest dimensions (240 mm x 195 mm x 155 mm) enable it to be placed unobtrusively in small rooms, on bookshelves etc, or on the rear window-shelf of a car. The LM1 is also available in a panel mounting version.

The LM1 has a number of unique design features for its reliable operation in the high ambient temperatures often experienced in cars. These features include the use of a die-cast alloy enclosure lined with a heavy vibration-damping material

and B & W's Kevlar construction for the bass/mid-range loudspeaker cone. The crossover network is provided with a switch which enables the frequency response characteristic to be adjusted either for use in normal rooms or for use in the car.

To protect the LM1 from accidental overload when used with high-power amplifiers the crossover network is fitted with B & W's patented automatic overload protection device, APOC.





New range of Pioneer products

Pioneer has just released the new Black Avante range of equipment. There are six hi-fi systems which all include an amplifier, AM/FM tuner, cassette deck, turntable, speakers and a matching general purpose glass door cabinet.

The top-of-the-range system also includes a graphic equalizer for achieving a flat frequency response and you can adjust the output sound to suit your own individual needs. Seven separate controls for each channel handle tonal adjustments and the signal-to-noise ratio is 100 dB.

There are five Black Avante amplifiers with power outputs ranging from 20 watts RMS per channel to 70 watts RMS per channel. Three tuners and three cassette decks in this range provide a choice in terms of quality and price. The cassette decks feature Dolby noise re-

duction, music search and automatic reverse on the top performing deck.

Pioneer is also releasing new products in the Champagne Gold range. These are two new hi-fi systems, Syscom 5 and Syscom 6, and a range of add-on equipment. The add-on components include three graphic equalizers, the SG-9, SG-3 and SG-300, the RG-9 dynamic processor and the SR9 reverberation amp. These components can be used with the accurate and convenient DT-5 and DT-510 digital timer/clocks.

SP-7 stereo headphones

The new Nakamichi SP-7 stereo headphones have been designed for exceptionally flat, natural response throughout the audio spectrum, broad dynamic range for unrestricted 'lifelike' reproduction and extremely low distortion for the highest possible fidelity to the source program.

These headphones offer the reproduction accuracy required for critical monitoring applications, as well as comfort and freedom from listening fatigue.

The original driver units are dynamic types featuring ferrite magnets 46 mm in diameter, 14.5 mm thick, and having a magnetic flux density of 5500 Gauss. This remarkably powerful magnetic circuit helps to achieve flat, resonance-free response from the extreme low-frequency range right up to the highest audible frequencies.

The diaphragm is formed of 25-micron thick polyester film with an effective diameter of 40.5 mm and a large 18.3 mm voice coil. The

diaphragm edge is a high-performance tangential design, and diaphragm breakup distortion has been eliminated through a special double coating process.

Nakamichi claim that, in addition to heightened transient response and minimized harmonic distortion, there is a significant reduction in intermodulation distortion. Further, special earpad design has reduced the variations in subjective response usually caused by changes in earpad pressure.

For more information contact Convoy International, 4 Dowling St, Woolloomooloo NSW 2011. (02) 358-2088.

Kiss the hiss goodbye!

Well, so say BASF with the launch of their new Chromdioxid II high bias tape, claimed to have the world's lowest background noise combined with outstanding sensitivity in the critical high frequency range.

BASF, which invented tape almost 50 years ago, has designed the Chromdioxid II for hi-fi applications where high output at all frequencies is required with low modulation noise.

The tape consists of perfectly-shaped and uniformly-sized particles of pure chromium dioxide, free from the imperfections that can plague ferric oxide tapes.

The new Chromdioxid II is the industry standard reference tape, as judged by the International Electro-technical Commission (IEC) for bias II — type magnetic recording

tape.

The Mobile Fidelity Sound Lab chose Chromdioxid II tape and the latest BASF-developed cassette shell for their Original Master Recording Series. These state-of-the-art pre-recorded cassettes are duplicated in real time from the original recording studio master tapes of some of the world's most prominent recording artists.

The cassette shell housing the new formula tape is BASF's ultra-precision model where tolerances are measured in micrometres.

Tandy to private-label Intellivision

John Roach, president and chief executive of Tandy Corp, has confirmed that Tandy will private-label Mattel's Intellivision video game system.

"We will be stocking up our stores in the next few months," he said, noting that the retail price has been set at US\$249. Tandy has purchased an estimated 100 000 systems from Mattel, and will market them under the Tandy Vision/Radio Shack label.

Mr. Roach said the court ruling enjoining Mattel from marketing six video game cartridges for the system "is not going to change our plans

at all." Mattel was enjoined from marketing six cartridges for violating a patent that had been licensed to N.A. Phillips.

Meanwhile, one source said that Mattel planned to initiate a US\$50 consumer rebate for its Intellivision video game which, when discounting is accounted for, would bring the price of an Intellivision game component down to around \$190.

Filth and degradation!

Dirt degrades cassette deck performance. We all know that. What you mightn't know is that Allsop, makers of the Allsop 3 cassette deck cleaning system, have introduced a new model cleaner, called the 'Ultraline'.

The original Allsop 3 cleaner consisted of a cassette case housing a cam, driven by the deck's take-up spindle, driving a felt pad that wiped across the heads. A second felt pad was held against the capstan and pinch wheel. It could only be inserted one way into the deck.

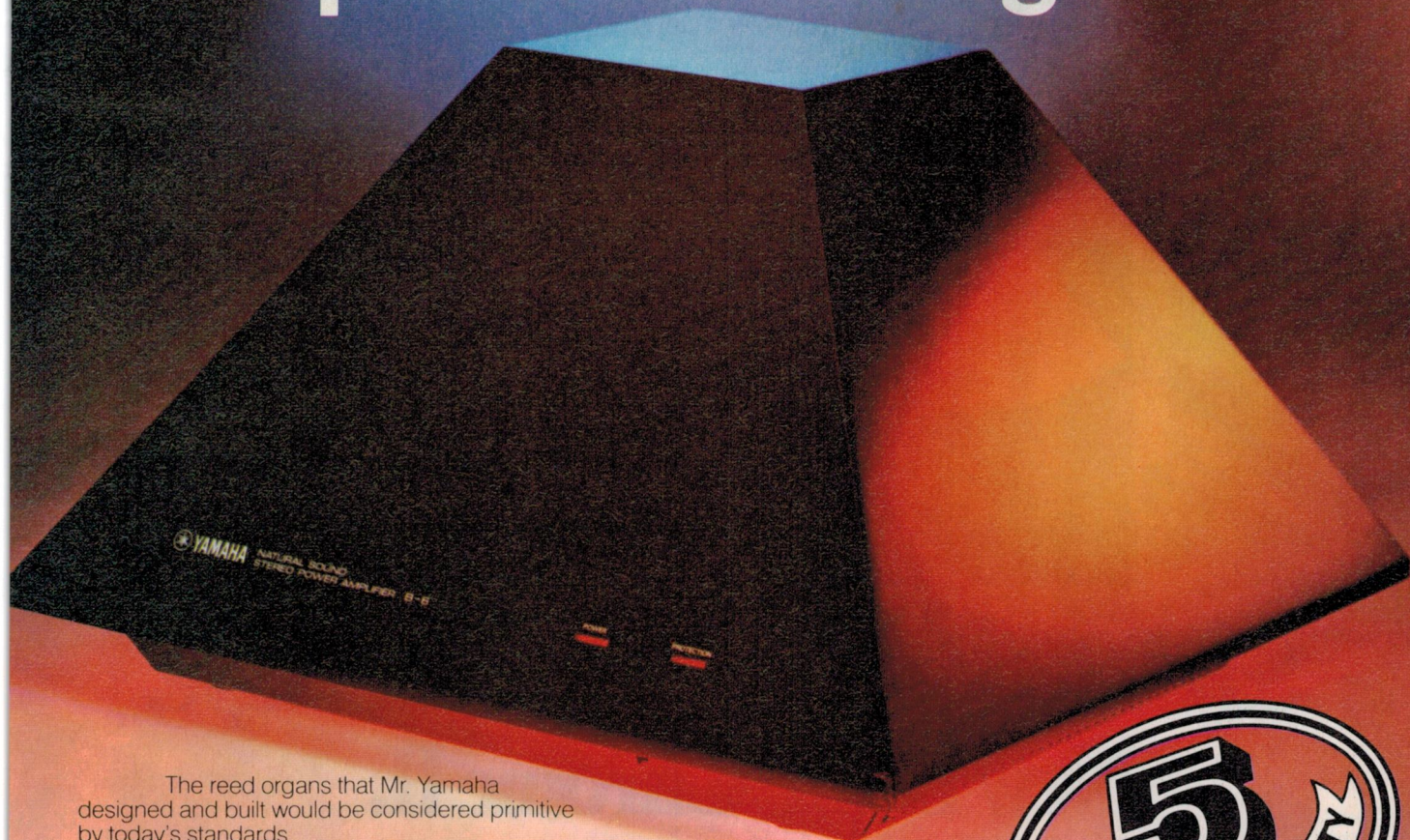
The new 'Ultraline' Allsop 3 cleaner can be inserted either way into the deck and incorporates improved cleaning facilities, according to Allsop. Firstly, the head cleaning pad now has a longer stroke through an improved epicyclic gear drive mechanism. Two felt pads are included for improved contact with the capstan and pinch roller. All pads are held in removable clips so that, when their life is exhausted they can be replaced. Replacement packs are readily available.

The new Allsop 3 Ultraline is available through hi-fi stores and record bars. Distributed in Australia



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The pyramid-shaped B-6 amplifier illustrated above for instance, is just as much 'state-of-the-art' now as Mr. Yamaha's reed organs were 95 years ago. And though technology has changed, the Yamaha principle hasn't.

All of our audio equipment, just like our fine musical instruments, is designed, crafted and ruthlessly tested by musicians. Just like our reed organs almost a century ago.

Indeed, the trained ear rather than a computer will always be the final arbiter of perfection.

And naturally the perfection that our musicians require and that our heritage demands, cannot be achieved by cutting corners or trimming costs.

Which may explain Yamaha's premium pricing and the full five year warranty we give all our audio equipment.

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For the wealthy—exclusive toys.

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The Marantz TT1000 (around \$2500), with its precision-made high density glass and golden aluminium sandwich structure, is justifiably described as one of the most beautiful turntables ever.

'Playing a series of direct-recorded discs, warped discs, discs with nasty low frequency content and discs requiring unusual trackability performance, showed clearly that this system borders on the superlative in areas where even most good turntables only provide good to above average performance.'

'...the resonance characteristics of the TT1000 are the lowest we have yet seen from any turntable irrespective of its selling price.'

'This is top-of-the-line equipment for people who rate hi-fi as their greatest pleasure in life'.
— Louis Challis, *Electronics Today International*, April 1981.

Similarly, the Marantz SM1000 Stereo Amplifier (around \$5000) is designed to be the ultimate in luxury and performance.

When it was bench-tested by ETI Magazine in an exhaustive lab study, Louis Challis stated 'The Marantz SM1000 Amplifier has the capability to provide superlative performance at home, in a laboratory, in a studio, or in a rock band with the ease and panache of a professional.'

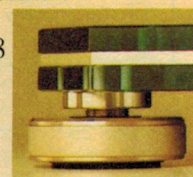
'The power output claims are modest for the unit is readily capable of producing 625 watts into an 8 ohm load with both channels driven...'

And when the Marantz ST8 FM/AM Tuner (around \$700) was

put through its paces so technically surprising was its performance that a second series of tests was devised to check the first results.

'As a result... the Marantz ST8... far ahead of any tuner we have ever measured and better than any tuner we have ever seen reviewed in any other magazine, either local or overseas.'

— Paul de Noskowski, *Electronics Australia*, April 1981.



The TT1000's adjustable, high-absorption air suspension audio insulator feet.

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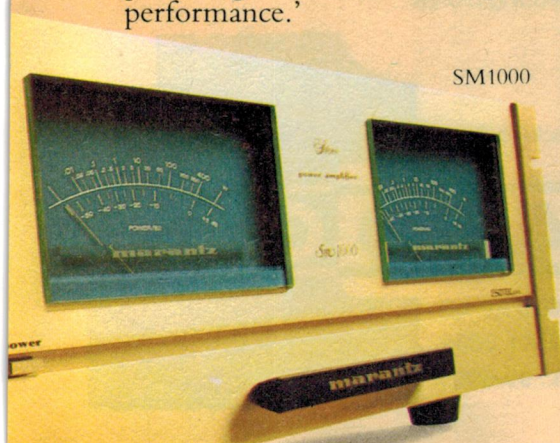


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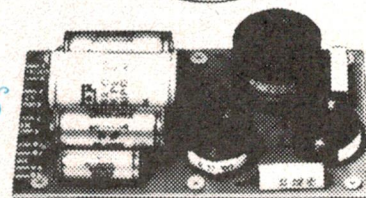
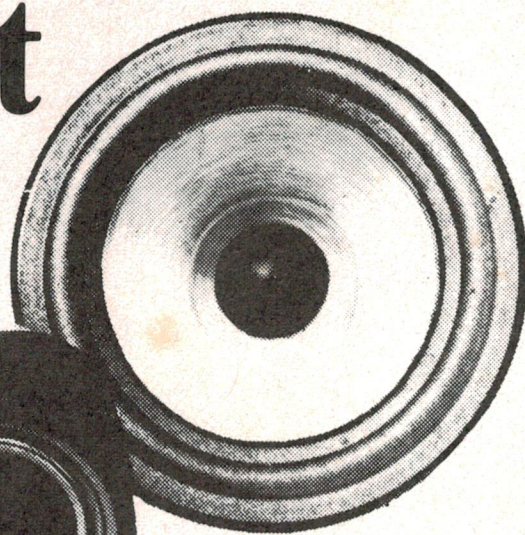
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LIFESTYLE NEWS

New videodisc record/replay system

Although videodiscs and their associated equipment are generally much cheaper than video cassette recorders, they suffer from the great disadvantage that it is normally impossible to alter the material recorded on them during manufacture.

The Japanese Sharp Company has already announced (ETI May 82, p. 121) a videodisc system using a laser which can record and it is interesting to note that the Japan Broadcasting Corporation (NHK) has now also developed a prototype of another type of video disc which can be recorded and erased any number of times.

In the NHK system a neon-helium laser is used to focus 10 mW of light onto the disc which employs the thermomagnetic characteristics of a 0.2 μm thick gadolinium-cobalt amorphous film for recording. A small magnetic field causes local reversal of the magnetisation of the film at points where the laser beam strikes the disc. The whole disc can be erased by a strong magnetic field, but selective erasure is also possible at any selected points.

Most optical disc recording techniques use the writing laser light to vapourise a thin film so that subsequent erasure for re-use is impossible. However, in the NHK system the laser light is attenuated to only

1.6 mW during playback which uses the Kerr optomagnetic effect. Disadvantages of the current system under development are occasional dropouts caused by imperfections in the disc surface and the poor signal-to-noise ratio of only 38 dB. However, NHK are confident that improved amorphous gadolinium-cobalt thin film coatings on the discs will result in a greatly improved performance.

Domestic users naturally wonder whether the equipment required for such recordable disc systems will ever enable them to be cost-competitive with video cassette recorders. If cheap recordable discs are developed, they could have great appeal to business users who would keep them in large numbers, but I cannot foresee the current types of recordable videodisc systems appealing to the home user owing to the high cost of the record/replay equipment. Clearly there is a great market if someone can solve this problem.

Brian Dance



Portable digital electronic time switch

Wattmaster Alco Pty Ltd has added a new model to its range of digital electronic time switches.

The 'Digital 2' is the first portable digital electronic time switch on the Australian market, claim Wattmaster Alco.

The unit has capacity for 12 programmable switching instructions which can be selected on a daily or weekly basis, and offers the added flexibility of programming in day-blocks. The provision of day-blocks means that several days' switching instructions can be programmed without affecting memory capacity.

Portability is achieved by using a revolutionary adaptor plug which provides power for the switch as well as power for the unit to be switched.

Switching capacity is 10 A 240 V, and the time is displayed with green LED's which are brightness self-

compensating to adjust for ambient light.

The 'Digital 2' can switch any period from a minimum of only one minute up to 24 hours, and can be programmed in one minute increments.

Wattmaster Alco says that the 'Digital 2' is ideal where accurate time switching is required up to one week in advance, such as for recording radio programmes, for switching medical equipment and other equipment where absentee switching is a benefit, as in the case of equipment with a long warm up period.

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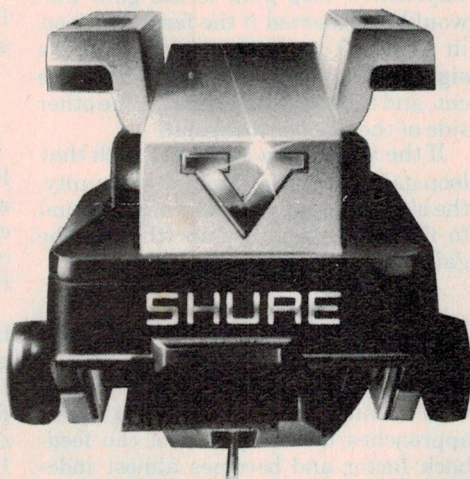
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Audio amplifiers using nested differentiating feedback loops

Part 1 — The present state of the art

The use of nested differentiating feedback loops (NDFLs) is a new technique for reducing audible-frequency distortion in an amplifier to a vanishingly low level. As the name implies, NDFLs rely on negative feedback, but they use it in a new way.

Edward M. Cherry

Associate Professor
Department of Electrical Engineering
Monash University

IN ORDER TO UNDERSTAND just how far the new NDFL technique can improve an amplifier, we first need to know the fundamental limits to the reduction of distortion that can be achieved with conventional techniques. In this first of three articles we survey familiar negative-feedback theory.

Figure 1 is a block diagram of an amplifier with negative feedback. In this diagram, the forward path corresponds to the amplifier before feedback is applied, and its gain is traditionally designated by the Greek letter μ . The feedback network returns a fraction β of the output to the input circuit, where it is in some way subtracted from the true input to provide the actual input to the forward path.

In many practical amplifiers, the subtraction is accomplished by applying the input and feedback signals to the two inputs of a balanced differential first stage of the forward path. Figure 2 is an outline practical circuit. In this circuit the feedback factor β is the attenuation of the network comprising R_{F1} and R_{F2}

$$\beta = \frac{R_{F1}}{R_{F1} + R_{F2}} \quad (1)$$

A typical value for an audio power amplifier might be 1/20. The forward-path gain μ in Figure 2 corresponds to gain from input to output when the feedback network is removed. A typical value for a simple audio power amplifier might be 1000.

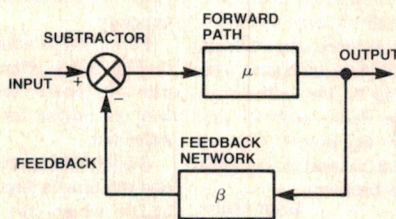


Figure 1. Block diagram of a feedback amplifier.

For Figure 1, the overall closed-loop gain A is given precisely by

$$A = \frac{\text{Output}}{\text{Input}} = \frac{\mu}{1 + \mu\beta} \quad (2)$$

The quantity $\mu\beta$ is called the loop gain. Physically, loop gain is the gain that would be observed if the feedback 'loop' in Figure 1 was cut at some point, a signal was injected into one side of the cut, and the resulting signal at the other side of the cut was measured.

If the values of μ and β are such that loop gain is small compared with unity, the closed-loop gain is very nearly equal to the forward-path gain (that is, the gain without feedback)

$$A \longrightarrow \mu \quad \text{as } \mu\beta \ll 1 \quad (3)$$

However, if loop gain is large compared with unity, the closed-loop gain approaches the reciprocal of the feedback factor and becomes almost independent of the forward-path gain

$$A \longrightarrow 1/\beta \quad \text{as } \mu\beta \gg 1 \quad (4)$$

The quantity $1/\beta$ is often called the demanded gain, as it is the value the overall closed-loop gain would take in ideal circumstances.

As a numerical example, if we substitute the above values $\mu = 1000$ and $\beta = 1/20$ into Equation 2, the gain of our 'typical' audio power amplifier works out as $A = 19.6$. The approximate Equation 4 predicts $A \rightarrow 20$, within 2% of the correct answer.

The quantity $1 + \mu\beta$ occurs often in feedback theory. It is called the return difference F

$$F = 1 + \mu\beta \quad (5)$$

Physically, return difference has the significance

$$F = \frac{\text{forward-path gain}}{\text{closed-loop gain}} \quad (6)$$

For values of loop gain greater than about 10, loop gain and return difference are almost equal — in our 'typical' example the values are 50 and 51 respectively.

Simplified treatments of feedback theory show that, if the distortion generated in the forward path (that is, the amplifier without feedback) at a particular output signal amplitude is D_μ , then the resulting closed-loop distortion D_A at the same output signal amplitude is

$$D_A = D_\mu / F \quad (7)$$

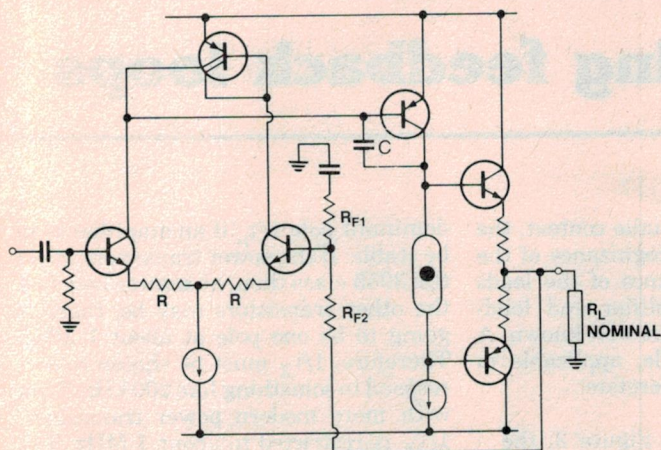


Figure 2. Outline circuit of an audio power amplifier.

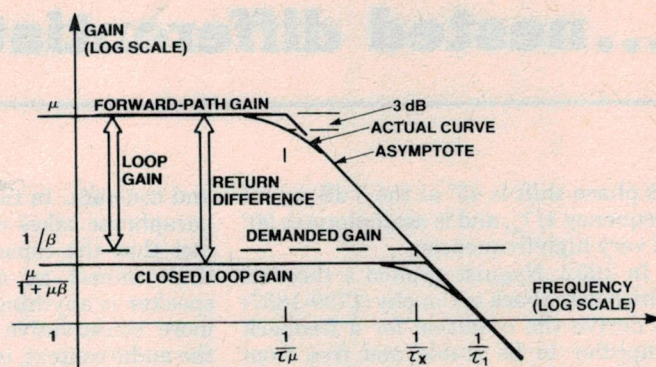


Figure 3. Logarithmic plots of gain versus frequency for Figure 1.

Distortion is improved when feedback is applied to an amplifier by a factor equal to the return difference. In our 'typical' amplifier, $F = 51$; if the distortion without feedback happened to be 10%, then feedback should reduce the distortion to 0.196%.

More rigorous treatments of feedback theory show that Equation 7 is no more than a poor approximation to the truth. In the first place, real amplifiers are far more complicated than Figure 1 suggests, because several different feedback paths (not all intentional!) can be identified. For example, the collector-base capacitances of transistors inevitably provide some unintended feedback at high frequencies. There is a very real problem in interpreting just what loop gain and return difference mean when there is more than one feedback loop. Once the correct interpretation is established, return difference invariably turns out to be a function of frequency, and the reduction of distortion corresponding to Equation 7 depends on the value of return difference at the frequency of the distortion, not the frequency of the input. Feedback therefore, does not reduce all distortion components equally.

Finally, it is found that the closed-loop distortion of an amplifier can contain new components that were not present in the distortion that existed in the forward path before feedback was applied. These new distortion components initially increase as loop gain is increased, but they fall away again towards zero as loop gain is made large.

Despite all these complications, the fact remains that adequate negative feedback, properly applied, does reduce distortion. Why, then, do amplifier designers not simply apply some arbitrarily large amount of feedback and reduce amplifier distortion to the vanishing point?

TIM, IIM, PIM, . . .

In the last 10 years or so, readers of audio magazines have been made aware of a conjecture that goes something like this:

"Harmonic distortion and the usual intermodulation distortion decrease with increasing feedback. Transient intermodulation distortion (TIM) increases with increasing feedback, and is approximately directly proportional to the feedback. Therefore, there is an optimum value for the feedback at which the subjective distortion sensation is least. This optimum feedback is unlikely to exceed about 20 dB."

More recently, there has been conjecture that heavy overall feedback should be applied with caution if interface intermodulation distortion (IIM) is to be avoided. An amplifier should provide a low open-loop output impedance so that the need for feedback-generated loudspeaker damping is minimised.

There has also been conjecture that negative feedback, which reduces the usual intermodulation distortion, may increase phase intermodulation distortion (PIM) by converting amplitude nonlinearities into phase nonlinearities.

Unequivocally, none of these conjectures has any basis in the new NDFL amplifiers. As an aside, there is a substantial body of opinion that none of these conjectures has any basis, full stop; interested readers should refer to References 1 — 12.

Instability and oscillation

A fundamental limit to the amount of feedback that can be applied to an amplifier is set by the onset of instability and oscillation.

If the magnitudes of the forward-path gain and demanded gain of the idealised Figure 1 are plotted versus angular frequency ω (in radian/second) on logarithmic scales, the resulting graph looks something like Figure 3. The 3 dB bandwidth of the amplifier without feedback is $1/\tau_\mu$, and the gain-bandwidth product (at which gain drops to unity) is $1/\tau_1$.

Because the graph is on logarithmic scales, the separation between the curves of forward-path gain and demanded gain is the loop gain (remember that, to divide two numbers, you subtract their logarithms; if you divide μ by $1/\beta$, you get $\mu\beta$). The magnitude of loop gain falls to unity at the frequency $1/\tau_x$ where the curves intersect and their separation is zero (remember that the logarithm of unity is zero).

By a similar argument, return difference is the separation between the curves of forward-path gain and closed-loop gain, as indicated in Figure 3.

We could make a similar graph to Figure 3, showing the phases of μ and $1/\beta$. Again, the phase of loop gain would turn out to be the separation between the two curves. However, there is a remarkable piece of mathematics due to Bode, who used a transformation evolved by Hilbert (1862-1943), which shows that there is a relation between the magnitude and phase of the response of any linear system. Subject to some qualifications, our proposed graph of the phases is completely predictable from Figure 3 and contains no new information. Interested readers may refer to Chapter 14 of Bode's book (Reference 13), but are warned that it is anything but easy going!

As an example, many readers will know that, if the forward-path in Figures 1 and 3 has a high-frequency cut-off rate variously described as single pole, 20 dB/decade, or 6 dB/octave, then

...nested differentiating feedback loops

its phase shift is 45° at the 3 dB cut-off frequency $1/\tau_\mu$, and is asymptotic to 90° at very high frequencies.

In 1932, Nyquist applied a theorem which dates back to Cauchy (1789-1857) to derive the condition for a feedback amplifier to be stable and free from oscillation. If a polar plot is made of the magnitude and phase of return difference as frequency is varied, a vaguely 'snail-shaped' curve results. Such a polar plot is called a Nyquist diagram. Subject again to some qualifications, the stability criterion for a feedback amplifier is that its polar plot of return difference should not enclose the origin. Figure 4 shows one example each of a stable situation and an unstable situation.

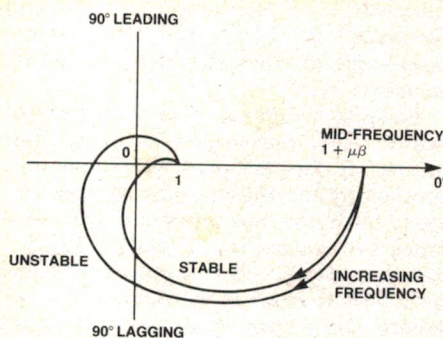


Figure 4. Nyquist's stability criterion. The curves are polar plots of return difference for changing frequency.

Because the phase of return difference can be predicted from Figure 3 via Bode's result a Nyquist diagram can also be constructed from Figure 3 and the onset of instability can be predicted. In 1945 Bode showed that Nyquist's criterion could in fact be expressed in terms of the gradients of the curves in Figure 3, thereby eliminating the work of finding the phase explicitly and plotting the Nyquist diagram. Bode's exact rule is complicated, but a useful paraphrase is

"If in graphs such as Figure 3 the separation between the forward-path gain and demanded gain decreases toward zero at a rate not exceeding 30 dB/decade, the amplifier is unlikely to oscillate."

This paraphrase makes no allowance for the tolerances on components. It assumes, in effect, that everything about the forward path is well known

and constant. In the audio context, the paraphrase takes no cognizance of the fact that the capacitance of the leads that connect an amplifier and loudspeaker is anything but well known. A more conservative rule, applicable to the audio context, is therefore

"In graphs such as Figure 3, the separation between the forward-path gain and demanded gain should not decrease towards zero at a rate exceeding 20 dB/decade."

The practical consequence is that the forward path of an audio amplifier with conventional resistive feedback should have a single dominant pole which sets the fall-off of gain at frequencies above $1/\tau_\mu$. The second and subsequent poles should all lie at frequencies substantially above $1/\tau_X$ (the frequency where the separation reaches zero), because each pole contributes a 20 dB/decade downwards slope to the graph of forward-path gain.

Maximum available feedback

In Figure 2, the first stage is a long-tailed pair with a current mirror at its output; the input and feedback signals are applied to the two bases to perform the subtraction process of Figure 1. The second stage provides a large voltage gain, and the lag compensating capacitor C provides the dominant pole of the forward path corresponding to $1/\tau_\mu$ in Figure 3. The third stage is a complementary class-B emitter follower whose function is to transfer the output voltage from the second stage to the loudspeaker load. In practice, the transistors in the second and third stages are often Darlington's, and the input transistors are often replaced by FETs.

In any amplifier, there is at least one pole associated with the finite transit time of electrons through each transistor. The transit time for typical small-signal transistors is a fraction of a nanosecond, but for power transistors of the ubiquitous 2N3055 class the transit time may be as long as a few tenths of a microsecond. Thus, the output stage of Figure 2 may have a pole in the vicinity of 1 MHz.

As we saw in the previous section, the unity-loop-gain frequency $1/\tau_X$ in Figure 3 must be substantially less than the frequency of all poles except the

dominant pole $1/\tau_\mu$ if an amplifier is to be stable. If the power transistors are of the 3055 class then, no matter how fast the other transistors may be, there is going to be one pole at about 1 MHz. Therefore, $1/\tau_X$ must be chosen to correspond to something like 200 kHz. Even with more modern power transistors, $1/\tau_X$ is restricted to about 1 MHz. The art of designing a stable power amplifier involves choosing the lag compensating capacitor C such that $1/\tau_X$ is appropriate to the transistors actually used.

The geometry of Figure 3 is such that, no matter how μ , β and τ_μ are separately chosen, the return difference $F(\omega)$ at any angular frequency ω cannot exceed

$$F(\omega) \leq 1/\omega\tau_X \quad (8)$$

Thus, if $1/\tau_X$ is designed to correspond to 200 kHz, return difference at 20 kHz cannot exceed 10 (= 20 dB), and cannot exceed 200 (= 46 dB) at 1 kHz. An amplifier that boasts 80 dB of feedback ($F = 10\,000$ at low frequencies) must have $1/\tau_\mu$ corresponding to about 20 Hz; return difference must begin falling above 20 Hz, and the former values at 1 kHz and 20 kHz (46 dB and 20 dB) still apply.

Returning now to Equation 7, the effectiveness of feedback in reducing distortion is set by the frequency of the distortion, not the frequency of the input. The audible frequency range is generally reckoned to extend to about 20 kHz and, with the foregoing constraints, return difference at this frequency cannot exceed 10. Remembering that 20 kHz is the third harmonic of 6.667 kHz, we see that feedback cannot reduce offensive odd-harmonic distortion of mid-treble input signals by more than a factor of 10. Remembering too that 20 kHz is the seventh harmonic of 2.857 kHz, we see that feedback cannot reduce crossover distortion of mid-range input signals by more than a factor of 10.

Until recently there has been no way around this problem except to increase the unity-loop-gain frequency $1/\tau_X$, and this demands that the frequencies of the transistor poles must be increased if stability is to be preserved. Fragile, expensive power transistors, with narrow bases to achieve short transit times, become mandatory.

(... to be continued)

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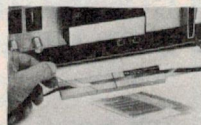
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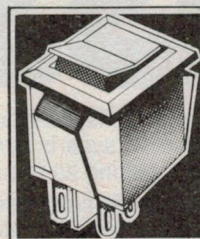
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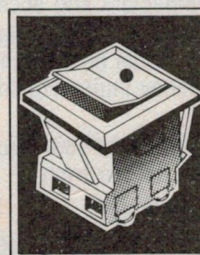
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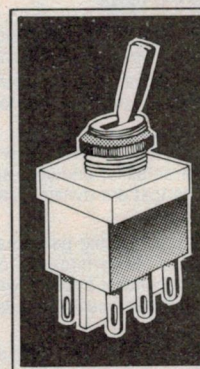
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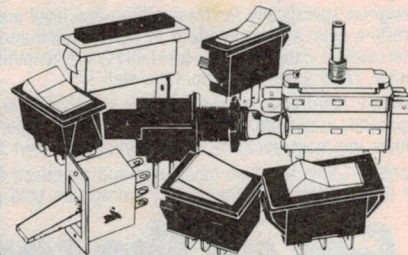


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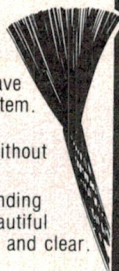
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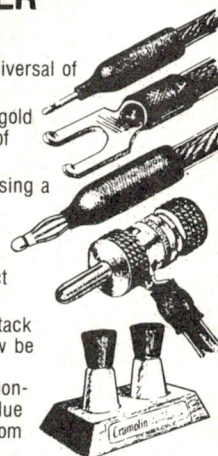
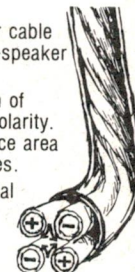
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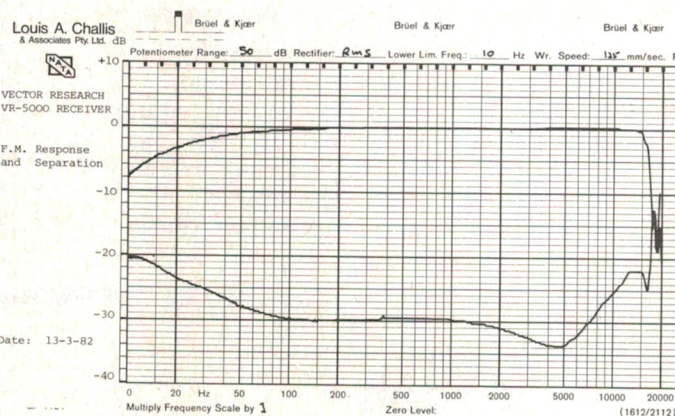
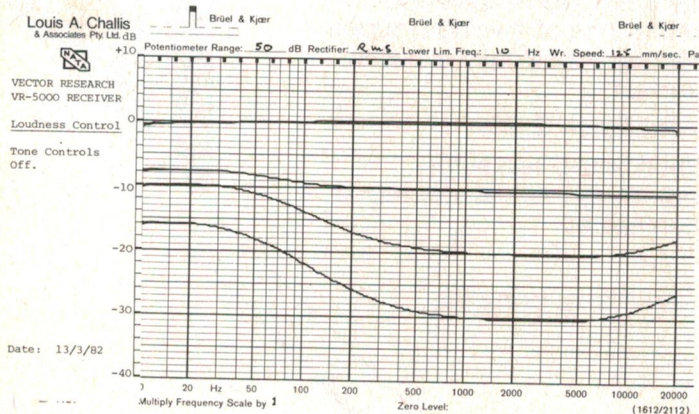
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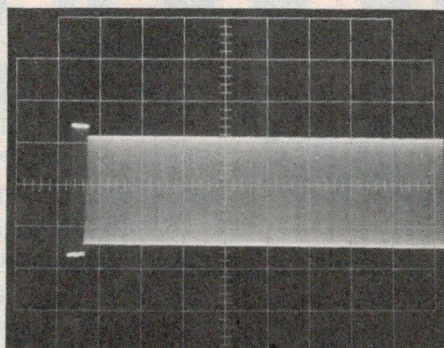
Features

Vector Research, based in California, are not yet well known in Australia. The designers have gone to a lot of trouble to provide all the features that the intending user may be looking for and a few more which he may not have thought of. The receiver is designed with an extruded aluminium front panel with a matt black finish and white silk screened designations for controls. The top third of the receiver features a rear-illuminated panel with a signal sensitivity meter at the left and a centre zero tuning meter adjacent to it so that the FM station can be visually aligned and correctly tuned. Adjacent to this is a 250 mm long slide rule dial with an FM tuning of 87 MHz to 109 MHz, whilst the AM section covers 515 kHz to 1650 kHz.

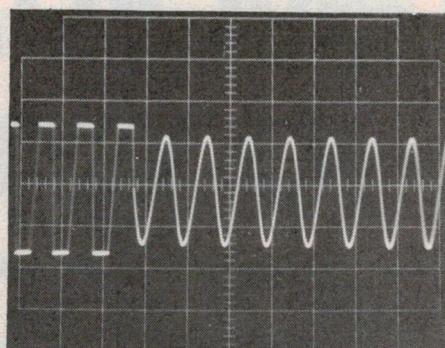
The central raised section of the receiver contains a power switch on the left hand side with two push button controls for high filter cut and FM muting on the right hand side. There are also function verification light emitting diodes to show the selection of AM/FM, FM stereo when the stereo signal is detected by the pilot tone carrier, phono and auxiliary.

The bottom row of controls includes a tip ring and sleeve socket for headphones and a speaker selection switch with OFF, A, B, and A+B. Unlike most other receivers on the market, as well as the bass and treble controls, the unit also has a mid range control. This provides modest but very usable and effective control for frequencies lying between 100 Hz and 10 kHz. The plateau region of this control covers

Transient overload recovery test (IHF-A-202). 10 dB overload re rated power into 8 ohms, both channels driven. Overload duration: 20 ms; Repetition rate: 512 ms.



50 ms/div



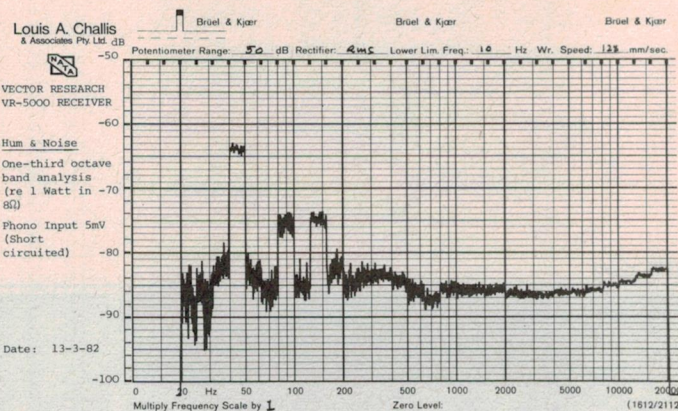
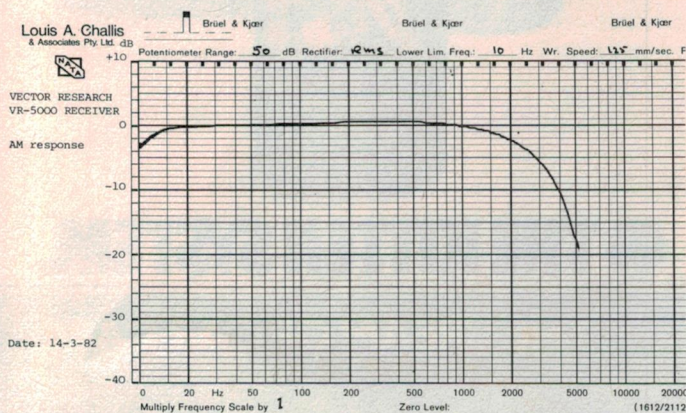
1 ms/div

the range 800 Hz to 2 kHz and provides the mid-range control in the frequency domain where it is most useful. The volume control and balance control are coaxially located centrally in the bottom panel. To the right of these are three toggle switches, one for Loudness ON/OFF, another switch is for selecting tape 1, source, or tape 2 and a third switch is for tape copy with selections of 1 to 2, out, and tape 2 to 1, for recording. The function selector provides direct switching for AM, FM MONO, FM AUTO, PHONO and Auxiliary. The 'auto' function indicates when the signal sensitivity is too low for good stereo reception to operate and then the receiver will automatically switch into the MONO mode. The tuning control operates through a counter-weighted drive system to smoothly select the required station.

The rear of the receiver has FM terminal connections of 75 ohms and 300 ohms. There is a ball-jointed moulded AM loopstick which has a

greater range of adjustment than most other loopsticks on the market but the reception will only be good on local stations. There is an aerial and earth connection for an external AM antenna which can be used in fringe areas of poor reception. Coaxial sockets are provided for moving magnet cartridges and auxiliary inputs, such as a separate radio receiver. There are two sets of input and output coaxial sockets for two separate tape recorders. Obviously, one does not need to connect both, but if one does the functions provided on the front panel are effective and facilitate tape copying without impinging on the radio or amplifier functions.

The speaker sockets are colour coded with effective spring loaded terminals which retain the bared speaker wires by means of sensibly designed taper entry moulded sockets. Above each of these sets of sockets is a speaker fuse with a 4 amp rating. These fuses are designed to blow in the event of an overload or as a result of shorting the speaker leads. ►



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KIW 311/2



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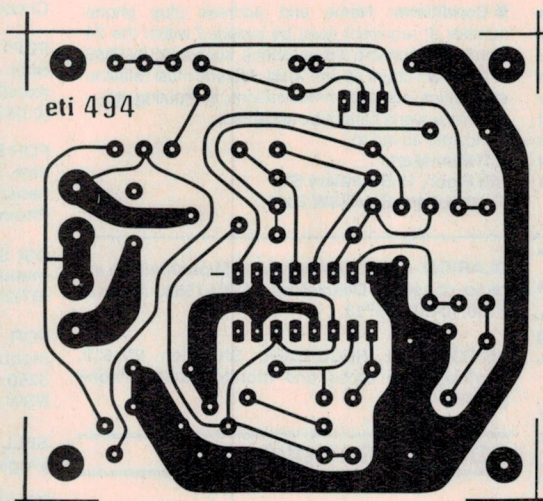
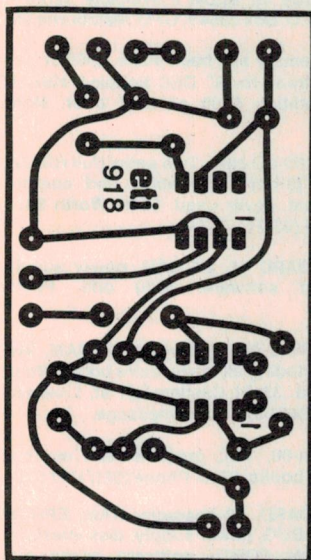
This method can be used to make negatives of ETI artwork from October 1977 on, provided the reverse of the page is printed in blue. The film used is Scotchcal 8007, which is UV sensitive and can be used under normal subdued light.

Cut a piece of film a little larger than the pc board and expose it to UV light through the magazine page. The non-emulsion side should be in contact with the page. This surface can be detected by picking the film up by one corner — it will curl towards the emulsion side. Exposures of about 20 minutes are normally necessary.

The film can now be developed by placing it emulsion side up on a table, pouring some Scotchcal 8500 developer on the surface and rubbing it with a clean tissue.

Further information on Scotchcal and pcb manufacture can be found in the September and December 1977 issues of ETI.

Please note that occasionally lack of space may prohibit the printing of blue type behind all pcbs. In this case the reader must resort to more conventional photographic techniques for pcb manufacture.

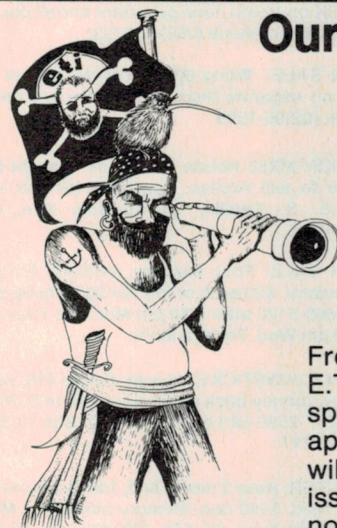


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TEKTRONIX OSCILLOSCOPE for sale. 60 MHz bandwidth, excellent condition. \$780. G. Kingsmill, 11 Plunkett St, W. Heidelberg Vic. Phone (03) 497-4291 after 7 pm.

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SORCERER 32K c/w all manuals, tapes, programs, inc. WP and Forth, cables for two tapes & printer, \$900. R. Hogan, 72 Allonby Ave, Wagga 2651. (069)22-7246.

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FOR SALE: APF Colour Computer, including colour monitor, 8K RAM, 14K ROM and joysticks. \$700. Phone (03)798-3319.

WANTED: Data sheets for MSM5523RS clock chip and MSL2312RS. S. Featherstone, 14 Pioneer St, Toowoomba Qld 4350. (076)34-3924.

SELL: System 80, Monitor, EDTASM, second tape recorder, cables and manuals. Used for master file update accounting. To be replaced by something bigger. \$750. (079)34-4225.

ZX81 MEMORY EXPANSION kits. Increases 1K to 2K for \$22 or 2K to 4K for \$40. Send order to M.E., P.O. Box 834, Traralgon Vic. 3844.

SELL: Microcomputer 48K compatible with Applesoft programs and hardware add-ons. \$790. Peripheral cards for Apple II available. Write to Choon M. Oi, P.O. Box 2354V, GPO, Melbourne Vic.

PDP11 16K memory and back plane. \$500 or best offer or part trade for 8" DEC floppies. KSR 35 excellent condition \$100 or best offer. Norm (03)534-1192.

FOR SALE: \$100 I/O card. Two serial (full RS232), nine parallel (8-bit) with cables and comprehensive manual, never used. \$300. Worth \$370. Phone Russell (03)211-7671.

OSI SUPERBOARD 11, 8K RAM, power supply, manuals and software. \$300 ono. Phone (07)205-3457.

FOR SALE: Micro-Ace computer, 2K RAM, VHF modulator. Including all leads, extra books. Value \$250, sell \$100. 33/51 Castlereagh St, Liverpool NSW 2170. (02)601-2460 take message.

SELL: System-80, 16K, green screen monitor, programs and books. \$750. Phone (02)713-7116.

OSI SUPERBOARD, 48 character RAM, EPROM expansion, DABUG, power supply, disk interface, EXMON EPROM, FORTH, software, books, bits and pieces. Offers. Ray (02)938-4767.

SELL: Commodore 8K old ROM, hardly used. \$700. Sound box fitted. David Christie, 34 Colville St, Highgate Hill Qld 4101. Phone (07)246-1474.

SUPER-80: Basic EPROM and power supply, \$100 fully connected, all IC sockets, 16K expandable to 48K on-board, tone generator and all documentation. \$339. Phone (053)32-6733.

FOR SALE: 'Wang 600' computer. Paper output but no separate display. Perfect condition. Best offer. (02)95-1599.

EPSON MX80 ribbon cartridges. Bought in bulk, have to sell excess. Send \$25 for two + small SSAE. S. Petelin, 36 Northam Ave, Bardon Qld 4065.

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APPLE JOYSTICKS for sale. Single \$18, twin \$35. 7 day, money back guarantee. Phone D. Plummer (03)267-2596 (ah) or write to P.O. Box 16, Dareton NSW 2717.

PRINTER, Base 2 model 800, 100 cps, serial, tract/frict feed. \$450 ono. Memory boards for Motorola Exorciser buss, 8K, 50, 16K \$90 ono. All in good condition. C. Stockdale, P.O. Box 871, Morwell Vic. 3840. (051)34-7836.

MICROBEE and Super-80 three tone, seven octave sound generator. Effects range from tunes to UFOs. Circuit diagrams, documentation etc, \$6. J. Burns, 6 Banksia St, Townsville Qld 4814.

ETI-660, colour, 3K RAM, audio amp, modulator, cassette leads and hex keyboard in quality metal case. Perfect working order. \$260. D. Poole, 1 Spurwood Cl, Kenthurst NSW. (02)654-1473.

SELL: Sorcerer MK1 32K RAM \$100 expansion. One disk, printer and paper. Video monitor. \$3500. (062)91-8881 after 5.30 pm.

Otherwise it's likely that the output transistors would be destroyed. The mains supply is also separately fused with a 5 amp fuse which is also accessible from the rear panel of the receiver.

The chassis is well made and solidly constructed from coated steel. The top cover and base of the unit are well ventilated to allow for heat dissipation from the amplifier. Unlike many other receivers on the market, this unit does allow the direct connection of either one set of 4 ohms speakers or two sets of 8 ohms speakers in parallel.

Inside

The inside of the receiver is sensibly designed with a fairly neat split between RF and audio stages on the two sides of the chassis. The left hand side of the chassis is divided into three sections with the preamplifier and medium level amplifier stages at the front, the main output stage on a large vertically finned heat sink in the centre and the power output stage, power supply transformer and fuses, located at the rear. The designers have utilised extensive areas of slotted perforations on both the under side chassis and in the top cover, immediately under and over the power output and power supply stages.

On the right of the unit the RF stage utilises conventional tuning gang and dual gate MOSFET RF transistors to provide reasonable sensitivity and good selectivity. Most of the wiring around the top of the printed circuits is associated with dial illumination and a small number of wires, which are in the main screened, carry signals from the rear of the RF stage through to the audio frequency and switching circuitry at the front of the unit. One unusual feature is

the design of the illumination circuits for the front of the slide rule dial. This features a fancy plastic moulding to carry the light through the clear plastic down to the front of the panel to provide a uniform light level. The designers have gone to some trouble to suppress switch-on transients with voltage dependent resistors and capacitors located across the power switch and so both switch-on and switch-off transients are positively suppressed.

Amplifier performance

The heart of the VR-5000 receiver is the amplifier which provides particularly good characteristics. The frequency response extends from 4.4 Hz to beyond 100 kHz, with the tone controls centred. The circuit provides particularly low distortion levels right across the frequency range at the rated output of 45 watts with both channels driven into 8 ohms. Under these conditions the distortion at 100 Hz is a modest .036%, at 1 kHz .047% and at 6.3 kHz .065%. At the 1 watt level these figures drop even lower being .012% at 100 Hz, .0056% at 1 kHz and .009% at 6.3 kHz. Any improvement beyond this is unwarranted in a piece of consumer electronics, although some manufacturers play the numbers game to have lower figures.

The transient intermodulation distortion is less than 0.1% whilst the hum and noise figures are a shade higher than we have come to expect from top line equipment, being -64.5 dB unweighted relative to the 1 watt level and -74 dB(A). The hum and noise levels on the phono input are very similar at -62 dB unweighted and -74 dB(A) relative to the 1 watt level. The maxi-

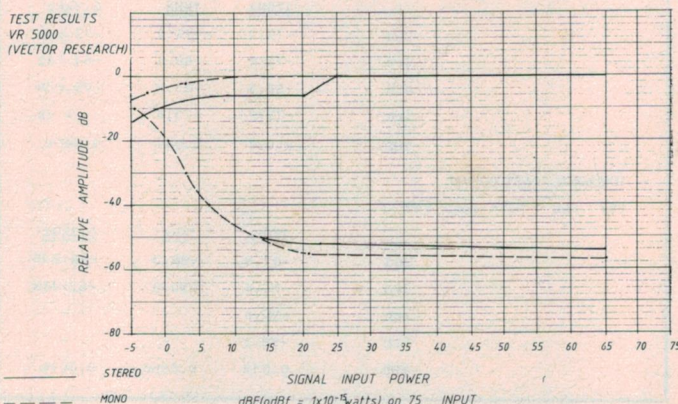
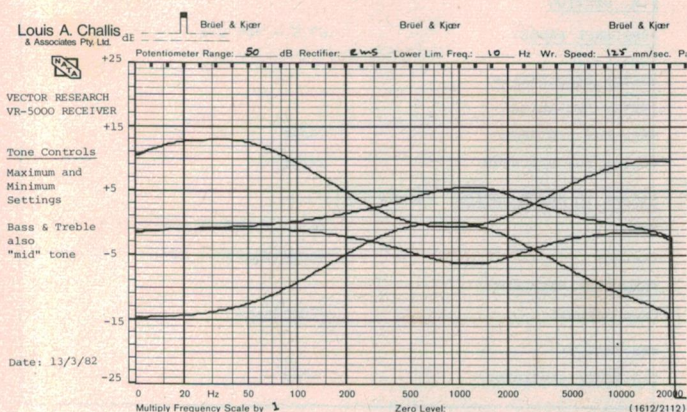
mum output at the clipping point is 66 watts which provides a dynamic head room of 1.7 dB relative to the rated 45 watts. The phono input level for 1 watt level is 390 microvolts whilst the overload point is 215 millivolts, which provides a very healthy overload margin to protect the amplifier.

The transient overload recovery test proved to be impeccable and both channels recovered within one half cycle from a full overload. The cross talk between channels is good at low frequencies being typically 40 dB up to 1000 Hz dropping down to -20 dB at 20 kHz. This performance is adequate but not superlative. The tone controls provide ± 14 dB bass boost and cut at 20 Hz, ± 10 dB of treble boost and cut at 10 kHz whilst the mid-range boost and cut is ± 6 dB at 1 kHz. The high cut filter provides a modest -10 dB of attenuation at 10 kHz extending to -15 dB at 20 kHz.

The amplifier section of the receiver is well designed and performs well considering the price of the receiver.

AM/FM tuner performance

The RF stage is not quite as well designed as the audio frequency stage. On FM the sensitivity is adequate, with 9 dB(F) input required for 26 dB signal to noise ratio on mono and 21 dB(F) required for 46 dB signal to noise ratio on stereo. The ultimate signal to noise ratio is limited by internal residual hum leakage. Because of this the maximum signal to noise ratio on FM stereo is 53.5 dB whilst the maximum signal to noise ratio on FM mono is limited to 57 dB. The frequency response, however, of the FM section is ruler flat from 100 Hz to 15 kHz and is only 3 dB down



at 20 Hz. The cross talk component is better than 30 dB down from 100 Hz to 7 kHz with a slight rise between 10 Hz and 100 Hz and a comparable slight rise between 7 kHz and 15 kHz.

The FM section has a number of other attributes and even without an aerial it pulls in half of the local FM stations well. It provides a clean and healthy signal on all the stations once the simple dipole aerial, supplied with the unit, is connected to the terminal. The distortion on FM at 50 dB(F) is less than half a percent at 1 kHz on stereo and less than 0.3% on mono. Were it not for the hum level generated back through the FM section, the performance of the FM stage would be amongst the best that we have seen.

On AM the performance is desultory, as we have grown to expect from Japanese AM tuners, with a smooth but generally modest frequency response

extending from 10 Hz to 2.5 kHz. This is adequate but not really good. The AM stage is only designed for local reception. However, when an external aerial is utilised at night time it pulls in distant stations from as far as 400 km away.

At home

I evaluated this unit at home with a wide range of cassette recorders, record players and with its own internal FM and AM receiver stages. When playing through either a set of B & W 801's or Fischer modular speakers, it provides impeccable performance with the amplifier. I played many records and tapes. Two new and very different records were Earl Klugh on "Fingerpainting" an exciting and superb original master recording from Mobile Fidelity (MFSL1-025) and the Johann and Josef Strauss "Waltzes, Polkas, Marches and Overtures" with the Berlin Philhar-

monic Orchestra conducted by Herbert von Karajan, which is a digital recording from Deutsche Gramophon (2741003). In each case the amplifier's performance proved to be superlative and the power rating of 45 watts proved to be more than adequate. I was able to appreciate the quality of the amplifier stage which is excellent.

The Vector Research VR 5000 is a basically good receiver with some outstanding features and a few weaknesses. If the designers only reduce the hum level on FM, they will then have a truly outstanding receiver that I would rate as near to 'top of the class'.

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MEASURED PERFORMANCE OF VECTOR RESEARCH VR-5000 RECEIVER

(S.N. 5011739)

FREQUENCY RESPONSE:

(-3dB re 1 Watt, 0.5V

Input to Aux)

Tone controls Centred

Left 4.4Hz to 100kHz

Left 4.5Hz to 100kHz

SENSITIVITY:

(for 1 Watt in 8)

AUX

Left

20.5mV

Right

21.0mV

TAPE

20.5mV

21.0mV

PHONO M/M

390 V

430 V

OVERLOAD M/M

215mV

220mV

INPUT IMPEDANCE:

AUX

19k

19k

TAPE

19k

19k

PHONO

52k

52k

OUTPUT IMPEDANCE:

230 milliohms (@ 1kHz)

HARMONIC DISTORTION:

(A) (At Rated power of 45 Watts

into 8 = 40.1 Volts)

| | 100Hz | 1kHz | 6.3kHz |
|------|-------|-------|---------|
| 2nd | -76.1 | -75.0 | -72.2dB |
| 3rd | -70.8 | -68.5 | -64.7dB |
| 4th | -84.9 | -83.0 | -75.6dB |
| 5th | -77.0 | -73.4 | - dB |
| THD. | 0.036 | 0.047 | 0.065% |

HARMONIC DISTORTION

(B) (At 1 Watt into 8 Ω)

| | 100Hz | 1kHz | 6.3kHz |
|-----|-------|--------|---------|
| 2nd | -81.8 | -86.3 | -84.8dB |
| 3rd | -81.4 | -90.9 | -82.4dB |
| 4th | -95.0 | - | - |
| 5th | -98.2 | - | - |
| THD | 0.012 | 0.0056 | 0.009% |

TRANSIENT INTERMODULATION DISTORTION:

< 0.1%

(3.15kHz square wave and

15kHz sine wave mixed 4:1)

NOISE & HUM LEVELS:

re 1 Watt into 8 Ω)

AUX

-64.5 dB(Lin)

-74 dB (A)

PHONO M/M

-62 dB(Lin)

-74 dB(A)

(with volume control

set for 1 Watt output with,

0.5V input (Aux)

5mV input (Phono M/M)

MAXIMUM OUTPUT POWER AT CLIPPING POINT:

(1HF-A-202)

(20mS burst repeated at 500mS intervals)

65 V P-P

= 66 Watts

Dynamic Headroom

= 1.7 dB

F.M. SECTION

FREQUENCY RANGE:

87.5 - 109 mHz

USABLE SENSITIVITY:

(40kHz deviation)

Mono for S/N 26dB

9dBf

Stereo for S/N 46dB

21dBf

SIGNAL TO NOISE RATIOS: (40kHz elevation) (

Mono

(some residual hum)

57.0dB

Stereo

(some residual hum)

53.5dB

FREQUENCY RESPONSE:

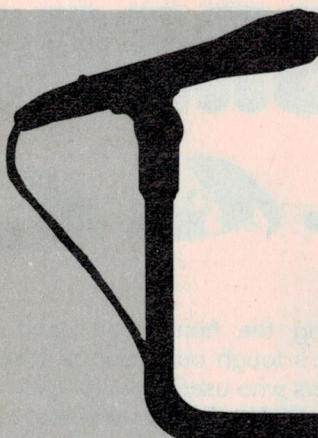
(see curves)

20Hz-16kHz

SEPARATION:

(includes generator)

30dB



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PL 88 OMNI
RE 16 RE 50
RE 20

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ROUND BASE STAND\$60.00
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SM 78
SM 10 Head-Set Mic
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565 SD
545 SD

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CK 2 20 Db Pad

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CHASSIS MALE\$4.50
BELDEN MIC CABLE (8412)\$2.85 per m

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ME 88 Shot-gun Capsule
ME 40 Cardoid Capsule
ME 20 Omni Capsule
K 3U Pre Amp
MKH 816 (P 48 UB)

SHURE

SM 81
SM 85

PZM

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6 7P

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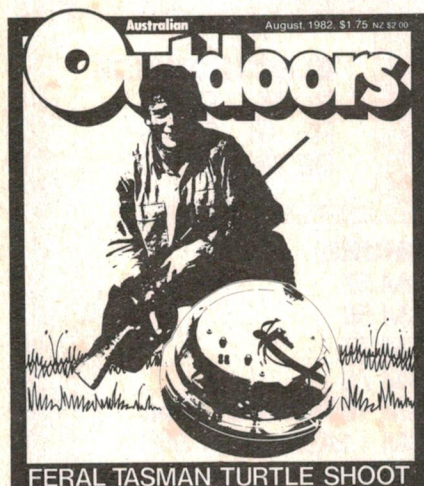
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OUTDOORS is a magazine also published by ETI's publishers and the reason why they rate a mention in Dregs is because they had a very 'special' issue recently, the cover of which is reproduced below:



The cover story on Tasman Turtles (see also ETI, April-May-June 82) was headed:

FERAL TASMAN TURTLES — the new menace!

The story went like this:

"Originally introduced to Australia by well-meaning computer buffs as robot projects, the Tasman Turtle has grown to become a threat to the Australian countryside. The feral population grew from the number of domestic Turtles being dumped in the bush because owners had lost interest or become bored with the programs provided. (After all, they couldn't play Space Invaders!)

"The feral population is now so large that Tasman Turtles have become one of the more popular game for local shooters.

"The Turtle is prized for several reasons. Because of its computer background it is a very cunning animal,

making the hunt good sport. The Turtle's tough outer skin is prized by hunters who used them for everything from potplant holders to light shades!

"The Turtle is also prized for the unique pen that it carries secreted inside its body and which it sometimes projects out to dig up Telecom wires (its main source of food) and to ring-bark electricity poles. One especially good green specimen shot recently near Wagga was found to contain a gold-plated Shaeffer pen with diamond inlay, but this is acknowledged to be a rare case. Normally, the Turtle is found with either a Bic Finepoint or Ball Pentel, particularly around the Canberra-Monaro district where the pens carry a 'Commonwealth of Australia' mark."

The special issue of Outdoors proved a complete surprise to its editor, Peter Scott (one time editor of ETI's sister publication Hi-fi Review, now defunct). The whole thing was 'dummied-up' by ETI's advertising production manager, John Gerrie and layout artist, Giitha Pilbrow!

The strangest things . . .

People talk about 'strange' things, at times. Perhaps only the physicists amongst Dregs readers will appreciate this one, but we pass it on for amusement's sake, anyway.

Conversation overheard in the Tilbrook household:

"I see they've demonstrated quark/anti-quark particle interactions."

"Well I'll be charmed!"

We leave it to readers to **imagine** what follows . . .

What's he talking about?

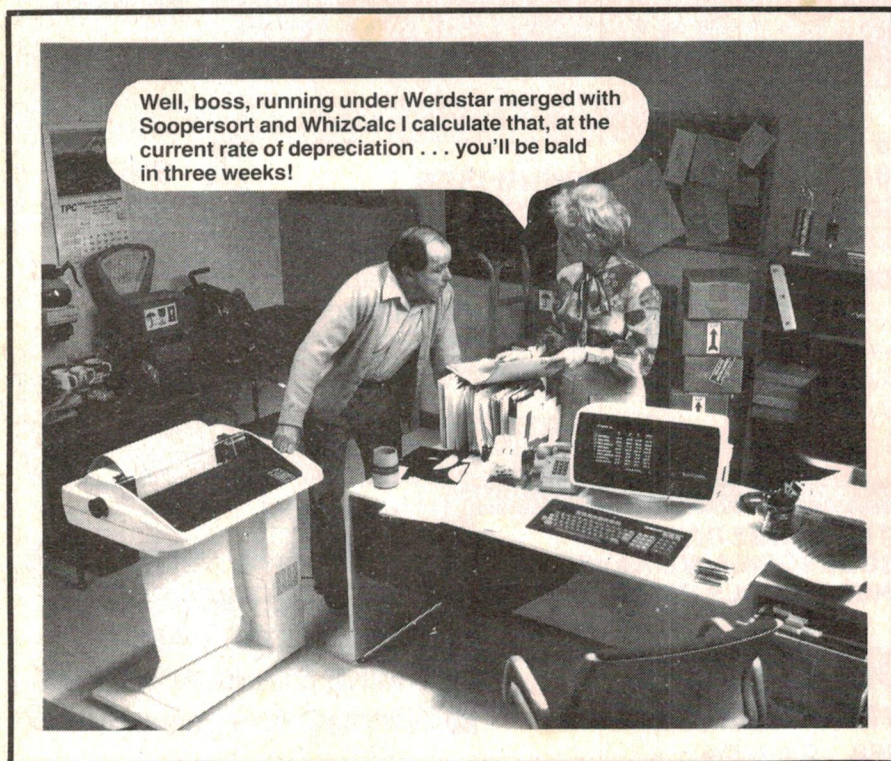
Conversation overheard in the ETI office.

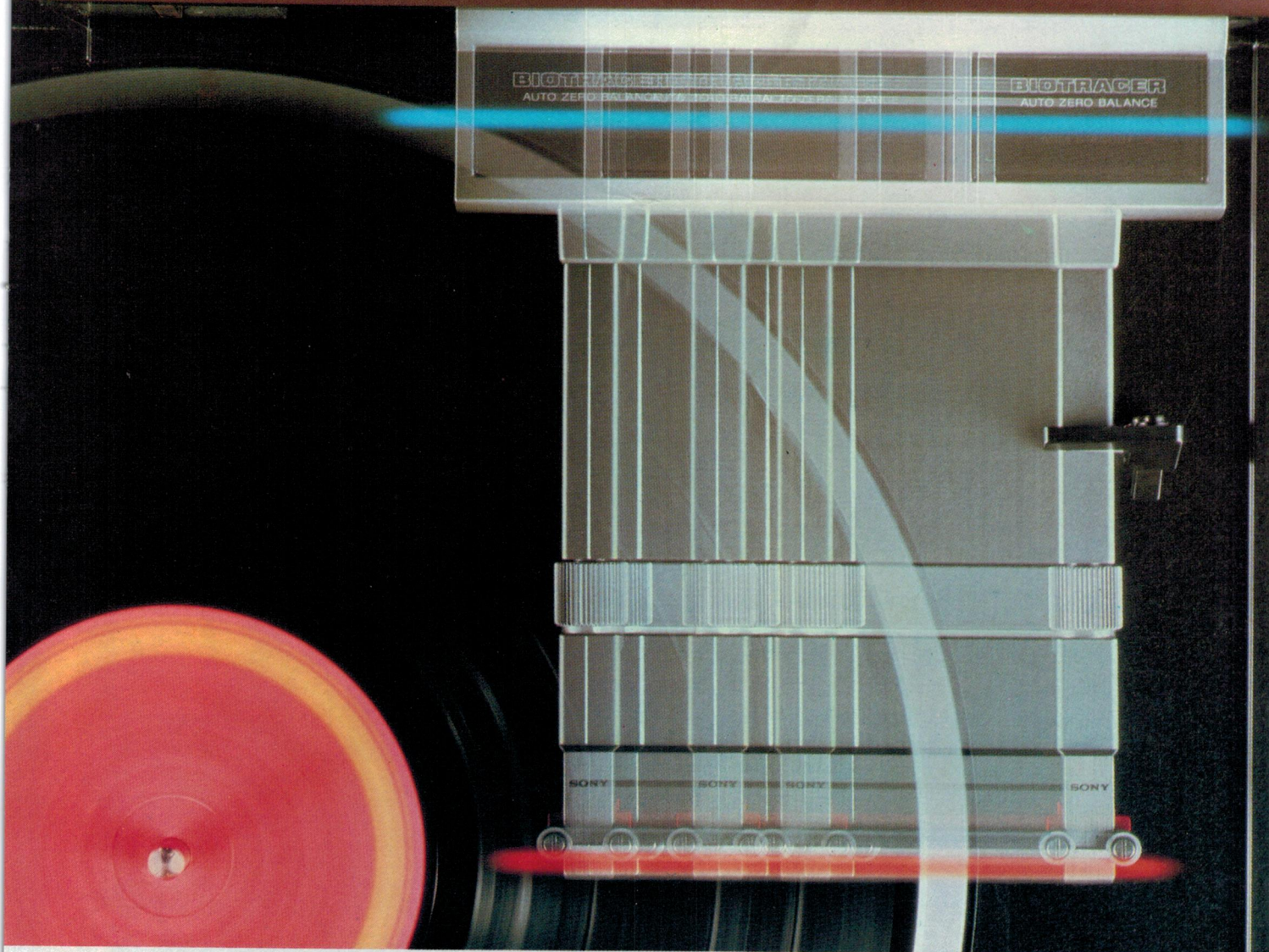
"What's a quark?"

"Well, that's a particle that . . ."

"It's the sound a duck makes when you wring its neck!"

"That's funny. I thought it was the stuff electronics people got on their teeth!"





Off on a tangent with Sony.

We didn't jump into tangential tracking turntables right off the bat. And Sony hopes you didn't either. Because while most lateral tonearms don't exactly shift gears as they travel down their path, they do run into some rough spots. A hang-up called "cogging" that inhibits totally free flowing movement, and hampers left and right stereo separation.

Sony has alleviated cogging and out of phase problems with an invention called Tangential Tracking Biotracer.

Controlled by two microcomputers and four sensors, the motion of the Biotracer tonearm is continuously fluid for precise phase alignment of the stylus.

To the average person these differences may sound slight. But if your standards are as high as Sony's, you'll understand the angle we're driving at.



SONY
THE ONE AND ONLY

SON 0116

4-101
PL259 Plug
takes Reducer



4-102
PL259 Plug b/i
Reducer RG58/U



4-103
PL259 Plug
Push-on Type



4-104
PL259 Plug
Simple RG58/U

4-105
PL259 Plug
Simple RG8/U



4-106
PL259 Crimp
Type for RG58/U



4-107
PL259 Crimp
Type for RG59/U



4-111
SO239 Socket
No Flange
Front Nut



4-112
SO239 Socket 4
Point Flange



4-113
SO239 Socket 2
Point Flange



4-116
UHF Double
Male Connector



4-117
UHF Right
Connector



4-118
UHF T
Connect
Female



UHF COLINEAR

General Specifications
These omnidirectional colinear antennas are designed for base station use in the UHF band. They are designed to operate under extreme climatic conditions. Constructed of anodized aluminum alloy, the active radiating elements are enclosed and supported within a fiberglass radome. The system is internally DC grounded to reduce static noise and internally lightning protection. The antenna incorporates a unique decoupling transformer which prevents RF energy loss through radiation by masts or transmission lines without the need for the usual decoupling radials. The low profile and light weight of the unit eliminates the need to provide heavy duty support structures.

| | |
|---------------------------|-------------------|
| Electrical Specifications | |
| Bandwidth: 2-645 | Specify frequency |
| 2-646 | 2.5% |
| 2-647 | Specify frequency |
| 2-648 | 2.5% |
| 2-649 | 477 MHz |
| 2-650 | Specify frequency |

2-645
UHF Colinear
3dB Gain
2.5% Bandwidth
Specify Frequency

2-646
UHF Colinear
3dB Gain
477MHz
Citizens' Band

2-649
UHF Colinear
9dB Gain
477MHz
Citizens' Band



SECTION 2 UHF BASE STATION ANTENNAS

Look for the components & accessories you need between these covers

Sourcing the huge range of components and accessories needed by today's electronics retailer can be a real headache. Here's the cure for that headache: Benelec's new catalogue. One of the widest ranges in Australia of active and passive components, instruments, antennas, connectors and cables. All listed with comprehensive data in the easiest to use catalogue in Australia. Benelec, formerly known as I.F.T.A. Australia, have been serving the needs of the electronics industry for over six years, with top quality products from around the world. Always at competitive prices. Always with fast delivery. Our ex-stock range includes:

- Mobile and base station transceiver antennas.
- Virtually all the popular coaxial cables, plus a variety of figure 8 styles.
- Instruments. Low cost RF test equipment and the famous Univolt range of digital and analogue multimeters.
- Power supplies ● Alarm equipment ● 27MHz transceivers ● Filters and noise suppressors ● Quartz crystals – standard and custom ● Microphones ● Speakers ● transceiver accessories ● Prepared

- leads ● Semiconductors ● Resistors ● Capacitors ● Fuses

If you're in the market for just about anything electronic, phone or write for your copy of our brand new catalogue – it could save you a fortune in time and money.

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